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FL103 LED 照明用 原边反馈 (PSR) PWM 控制器

特性

- 低待机功率: < 30 mW
- 高压启动
- 极少的外部元件数量
- 恒压 (CV) 和恒流 (CC) 控制 (不带次级反馈电路)
- 绿色模式: 线性降低 PWM 频率
- 50 kHz 和 33 kHz 的固定 PWM 频率 (采用抖频来解决电磁干扰问题)
- 恒压模式下的峰值电流模式控制
- 逐周期限流
- V_{DD} 过压保护 (OVP)
- V_{DD} 欠压锁定 (UVLO)
- 可调欠压检测器
- 栅极输出最大电压箝位在 15 V
- 热关闭保护 (TSD)
- 采用 8 引脚 SOIC 封装
- 应用电压范围: $80 V_{AC} \sim 308 V_{AC}$

应用

- LED 照明
- 电池充电器适用于移动电话、无线电话、PDA、数码相机、电动工具

说明

这种第三代原边反馈 (PSR) 和高度集成的 PWM 控制器提供了各种增强 LED 照明性能的功能。

专有拓扑结构 TRUECURRENT® 能够为 LED 照明应用提供精确的恒流调节和简化电路。与常规设计或线性变压器相比, 其具有低成本和更小型化的 LED 照明。

为了尽量降低待机功耗, 专有绿色模式功能提供了关断时间调制, 可在轻负载条件下线性降低 PWM 频率。绿色模式有助于电源达到节电要求。

通过使用 FL103, 可以用极少的外部元件和最低的成本来实现 LED 照明。



图1.8 引脚 SOIC

订购信息

器件编号	工作温度范围	顶标	封装	包装方法
FL103M	-40° C至+125° C	FL103	8 引脚小尺寸封装 (SOIC-8)	卷带和卷盘

应用框图

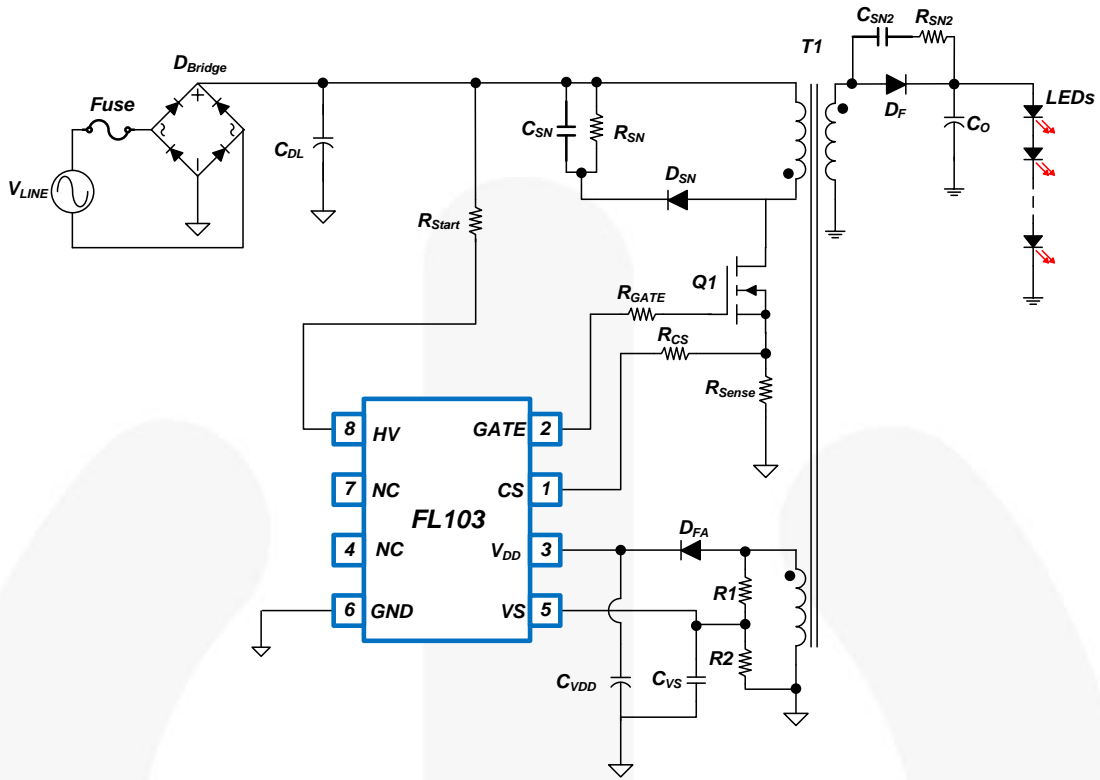


图2.典型应用

框图

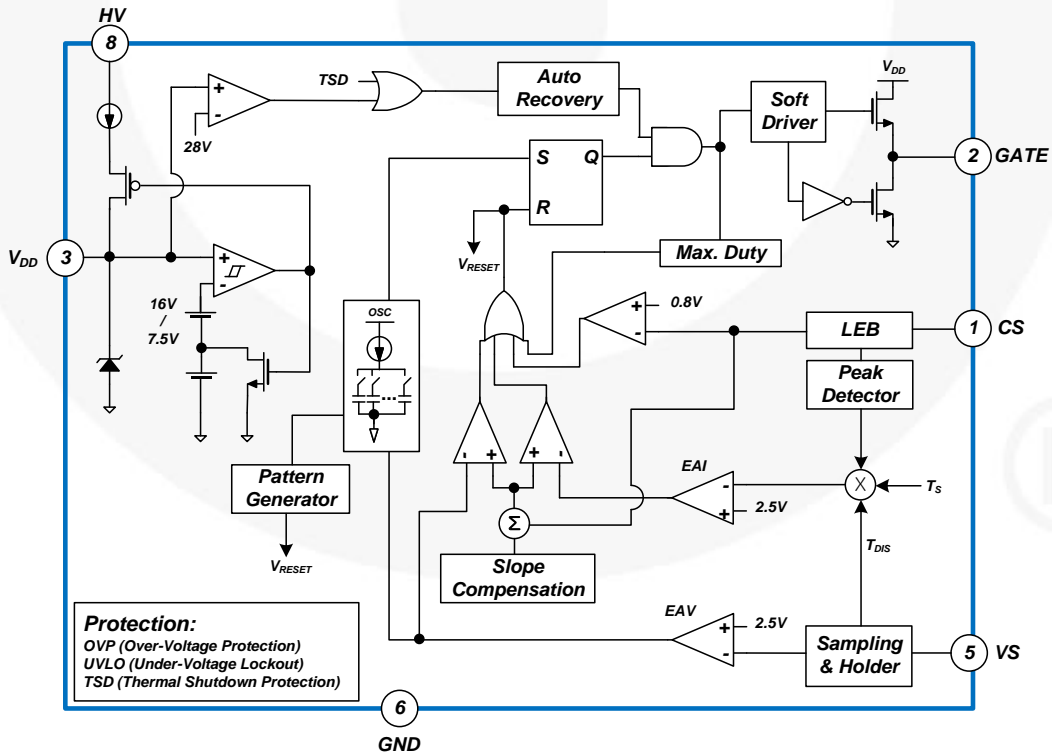
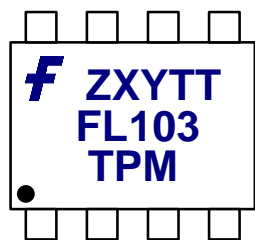


图3.内部框图

标识信息



F: 飞兆公司标志
 Z: 工厂编码
 X: 一位数字年份代码
 Y: 一位数字周代码
 TT: 两位数字模具运行代码
 T: 封装类型 (M=SOP)
 P: Y = 绿色封装
 M: 制造 流程编码

图4.顶标

引脚布局

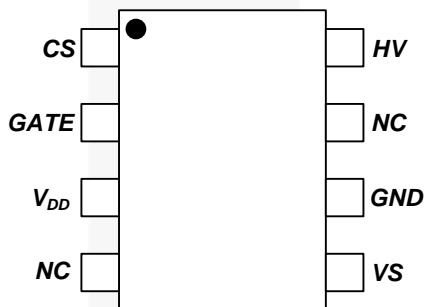


图5.引脚布局

引脚定义

引脚号	名称	说明
1	CS	电流检测。 该引脚连接电流检测电阻以检测 MOSFET 电流，实现恒压模式下的峰值电流模式控制，该引脚还提供恒流模式下的输出电流调节。
2	栅极	PWM 信号输出。 此引脚采用内部图腾柱输出驱动器，用于驱动功率 MOSFET。该引脚内部箝位到低于 15 V。
3	V _{DD}	电源。 集成电路工作电流和 MOSFET 驱动电流通过此引脚提供。该引脚连接至外部 V _{DD} 电容，典型值为 10 μF。启动和关断的阈值电压分别为 16 V 和 7.5 V。工作电流低于 5 mA。
4	NC	未连接。 该引脚连接至 GND 或无连接。不连接任何电压源。
5	VS	电压检测。 该引脚根据辅助绕组电压检测输出电压信息和放电时间。
6	GND	接地
7	NC	未连接
8	HV	高压。 该引脚连接至直流支撑电容，实现高压启动。该引脚连接至典型值为 100 kΩ 的外部启动电阻。

绝对最大额定值

应力超过绝对最大额定值，可能会损坏器件。在超出推荐的工作条件的情况下，该器件可能无法正常工作，所以不建议让器件在这些条件下长期工作。此外，过度暴露在高于推荐的工作条件下，会影响器件的可靠性。绝对最大额定值仅是应力规格值。

符号	参数	最小值	最大值	单位
V _{HV}	HV引脚输入电压		500	V
V _{VDD}	直流电源电压 ⁽¹⁾		30	V
V _{VS}	VS 引脚输入电压	-0.3	7.0	V
V _{CS}	CS 引脚输入电压	-0.3	7.0	V
P _D	功率耗散 (T _A < 50° C)		660	mW
θ _{JA}	热阻 (结到空气)		+150	° C/W
θ _{JC}	热阻 (结到外壳)		39	° C/W
T _J	结温	-40	+150	° C
T _{STG}	存储温度范围	-55	+150	° C
T _L	引脚温度 (波峰焊接或 IR, 10 秒)		+260	° C
ESD ⁽²⁾	静电放电能力	人体放电模式 (HV 引脚除外), JEDEC-JESD22_A114	4.50	kV
		组件充电模式 (HV 引脚除外), JEDEC- ESD22_C101	1.25	

注：

- 测得的所有电压，除差模电压之外，都参照GND引脚。
- 全部引脚：HBM =1500 V, CDM =750 V.

推荐工作条件

推荐的操作条件表定义了器件的真实工作条件。指定推荐的工作条件，以确保器件的最佳性能达到数据表中的规格。飞兆半导体建议不要超过推荐工作条件，也不能按照绝对最大额定值进行设计。

符号	参数	最小值	典型值	最大值	单位
V _{DD}	连续工作电压			25	V
T _A	工作环境温度	-40		+125	° C

电气特性

除非另有说明, $V_{DD} = 15\text{ V}$ 且 $T_A = 25^\circ\text{ C}$ 。

符号	参数	工作条件	最小值	典型值	最大值	单位	
V_{DD}部分							
V _{DD-ON}	导通阈值电压		15	16	17	V	
V _{DD-OFF}	关断阈值电压		7.0	7.5	8.0	V	
I _{DD-OP}	工作电流			3.2	5.0	mA	
I _{DD-GREEN}	绿色模式工作电源电流			0.95	1.20	mA	
V _{DD-OVP}	V _{DD} 过压保护值		27	28	29	V	
t _{D-VDDOVP}	V _{DD} OVP 延迟时间		90	200	350	μs	
高压 (HV) 部分							
V _{HV-MIN}	HV 引脚的最小启动电压				50	V	
I _{HV}	引脚 HV 可补充的电源电流	V _{DL} =100 V	1.5	2.0	5.0	mA	
I _{HV-LC}	启动后漏电流	HV=500 V, V _{DD} =V _{DD-OFF} +1 V		0.5	3.0	μA	
振荡器部分							
f _{OSC}	正常频率	中心频率	> V _O * 0.5	47	50	53	kHz
		抖频范围		±1.5	±2.0	±2.5	
	保护频率 ⁽³⁾	中心频率	< V _O * 0.5		33		
		抖频范围			±1.3		
V _{F-JUM-53}	跳频点	50 kHz → 33 kHz, VS	1.05	1.25	1.55	V	
V _{F-JUM-35}		33 kHz → 50 kHz, VS	1.28	1.50	1.75	V	
f _{OSC-N-MIN}	空载时的最小频率		300	450	600	Hz	
f _{OSC-CM-MIN}	CCM 模式下的最小频率		7	12	17	kHz	
f _{DV}	频率变化与 V _{DD} 偏差的关系	V _{DD} =10 V to 25 V		1	2	%	
f _{DT}	频率变化与温度偏差的关系	T _A = -40° C 至 +105° C			15	%	
电压检测 (V_S) 部分							
V _R	误差放大器的参考电压		2.475	2.500	2.525	V	
V _N	绿色模式开启 EAV 上的电压	f _{OSC} =2 kHz		2.5		V	
V _G	绿色模式结束 EAV ⁽³⁾ 上的电压	f _{OSC} =1 kHz		0.5		V	
V _{BIAS-COMV}	V _{COMV} 控制的自适应偏置电压	R _{VS} =20 kΩ		1.4		V	
I _{tc}	IC 偏置电流		7.3	10.0	12.7	μA	
I _{VS-B0}	欠压检测电流 ⁽³⁾			175		μA	
I _{VS-MIN}	最小 VS 电流 ⁽³⁾	90 V _{AC} , 重载		227		μA	
I _{VS-MAX}	最大 VS 电流 ⁽³⁾	264 V _{AC} , 空载		721		μA	
t _{DIS-MIN}	最小放电时间	正常运行 ⁽³⁾		0.65		μs	
		保护区	f _{OSC} =33 kHz	2.0	2.6		4.0

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电气特性 (续)

除非另有说明, $V_{DD} = 15\text{ V}$ 且 $T_A = 25^\circ\text{ C}$ 。

符号	参数	工作条件	最小值	典型值	最大值	单位
电流检测 (CS) 部分						
t_{PD}	GATE 输出传播延迟			90	200	ns
t_{MIN-N}	空载时的最小导通时间	$V_{COMR} = 1\text{ V}$	800	975	1150	ns
V_{TH}	限流的阈值电压		0.75	0.80	0.85	V
V_{TL}	VS 引脚阈值电压低于 0.5V			0.25		V
栅极部分						
DCY_{MAX}	最大占空比		60	75	85	%
V_{OL}	输出低电平	$V_{DD} = 20\text{V}$, 栅极灌电流为 10mA			1.5	V
V_{OH}	输出高电平	$V_{DD} = 8\text{V}$, 栅极源电流为 1mA	5			V
t_r	上升时间	$C_L = 1\text{ nF}$		200	250	ns
t_f	下降时间	$C_L = 1\text{ nF}$		60	100	ns
V_{CLAMP}	输出箝位电压	$V_{DD} = 25\text{ V}$		15	18	V
热关断 (TSD) 部分						
TSD	热关断温度 ⁽³⁾		+140			$^\circ\text{ C}$
TSD _{HYS}	热关断滞回 ⁽³⁾			+15		$^\circ\text{ C}$

注:

3. 该参数由设计保证; 未经 100% 产品测试。

典型性能特征

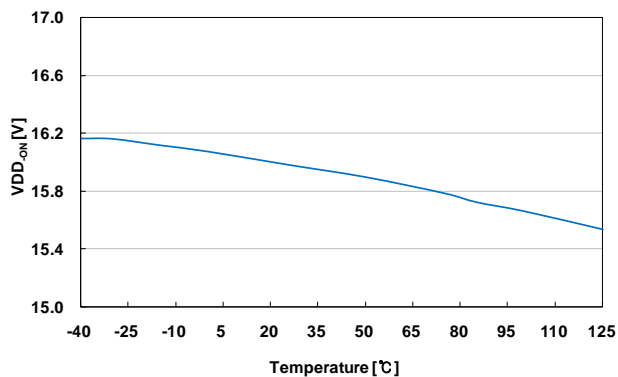


图6. V_{DD-ON}与温度的关系

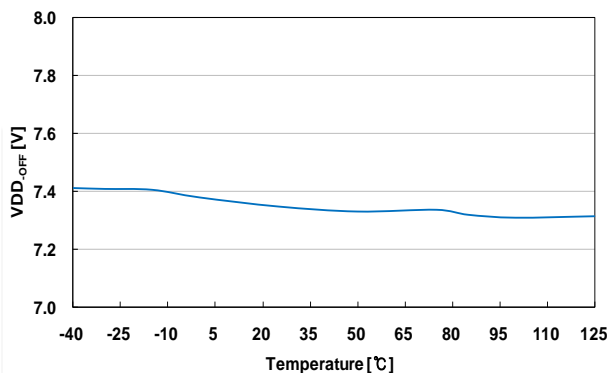


图7. V_{DD-OFF}与温度的关系

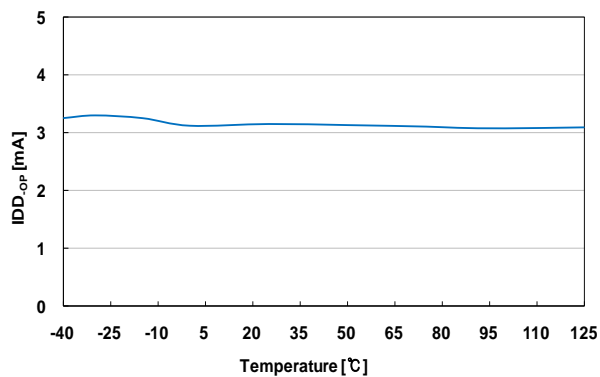


图8. I_{DD-OP}与温度的关系

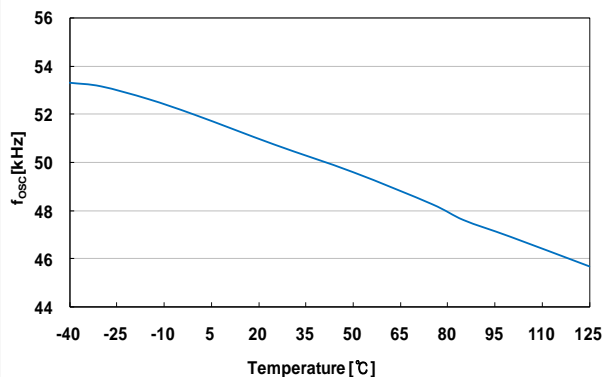


图9. f_{OSC}与温度的关系

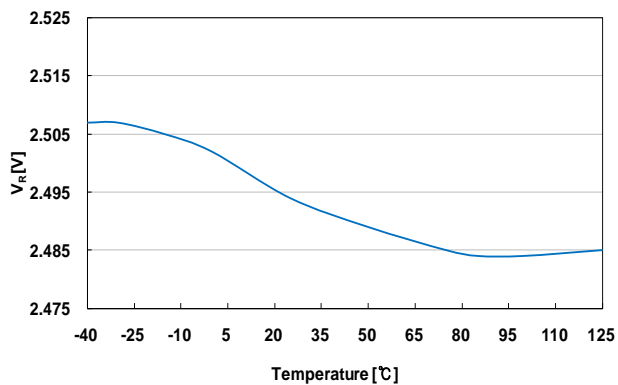


图10. V_R与温度的关系

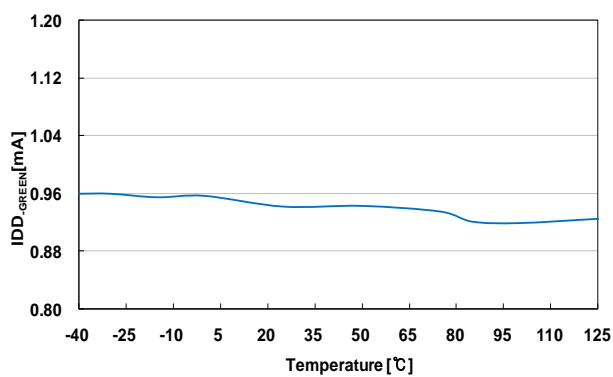


图11. I_{DD-GREEN}与温度的关系

典型性能特征 (续)

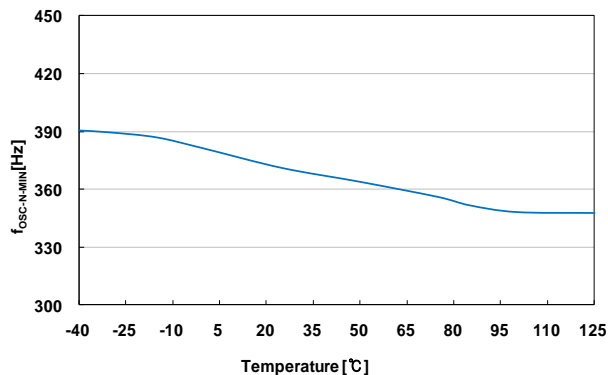


图 12. $f_{OSC-N-MIN}$ 与温度的关系

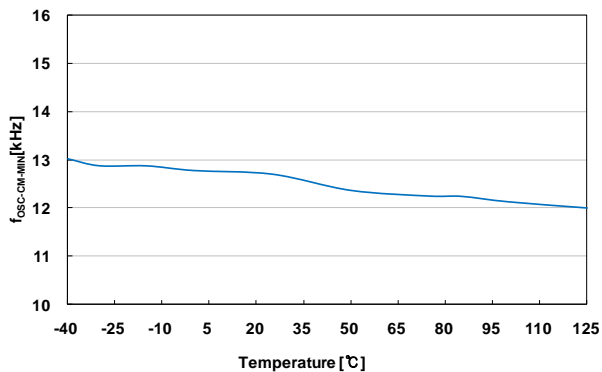


图 13. $f_{OSC-CN-MIN}$ 与温度的关系

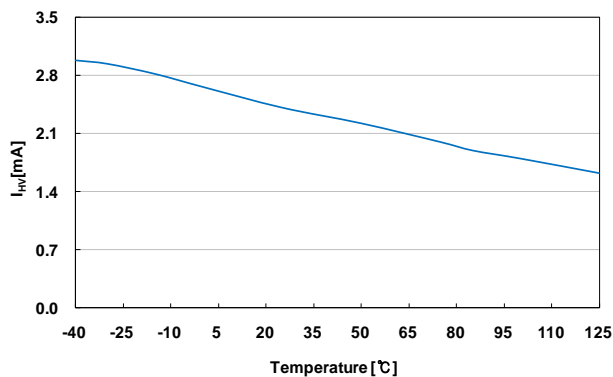


图 14. I_{HV} 与温度的关系

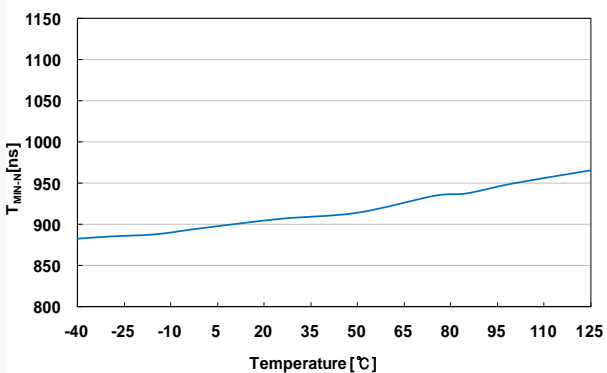


图 15. t_{MIN-N} 与温度的关系

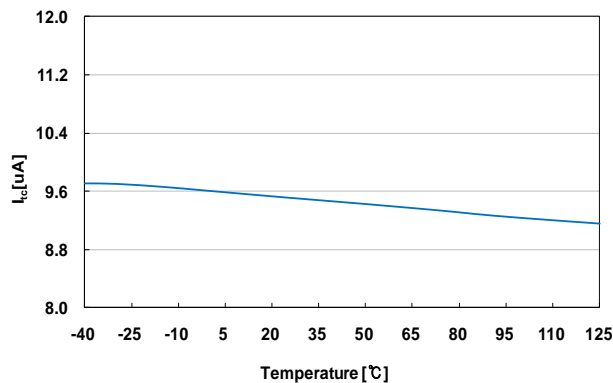


图 16. $I_{\tau c}$ 与温度的关系

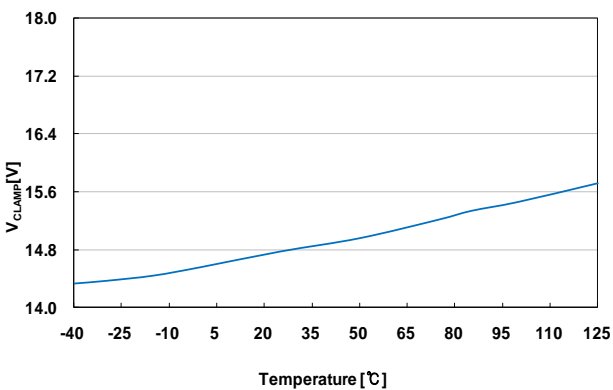


图 17. V_{CLAMP} 与温度的关系

典型性能特征 (续)

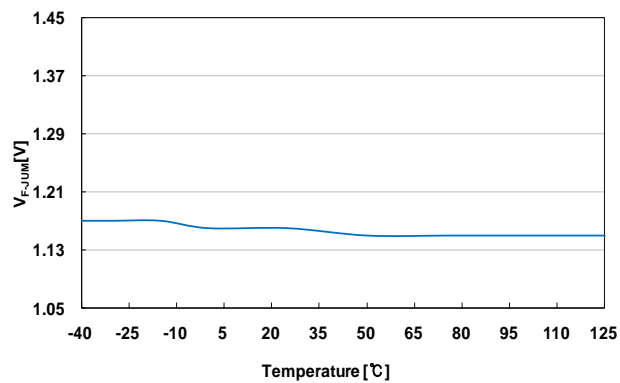


图 18. V_{F-JUM}与温度的关系

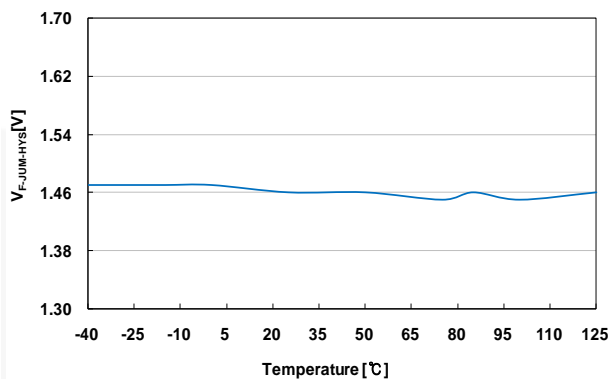


图 19. V_{F-JUM-HYS}与温度的关系

功能说明

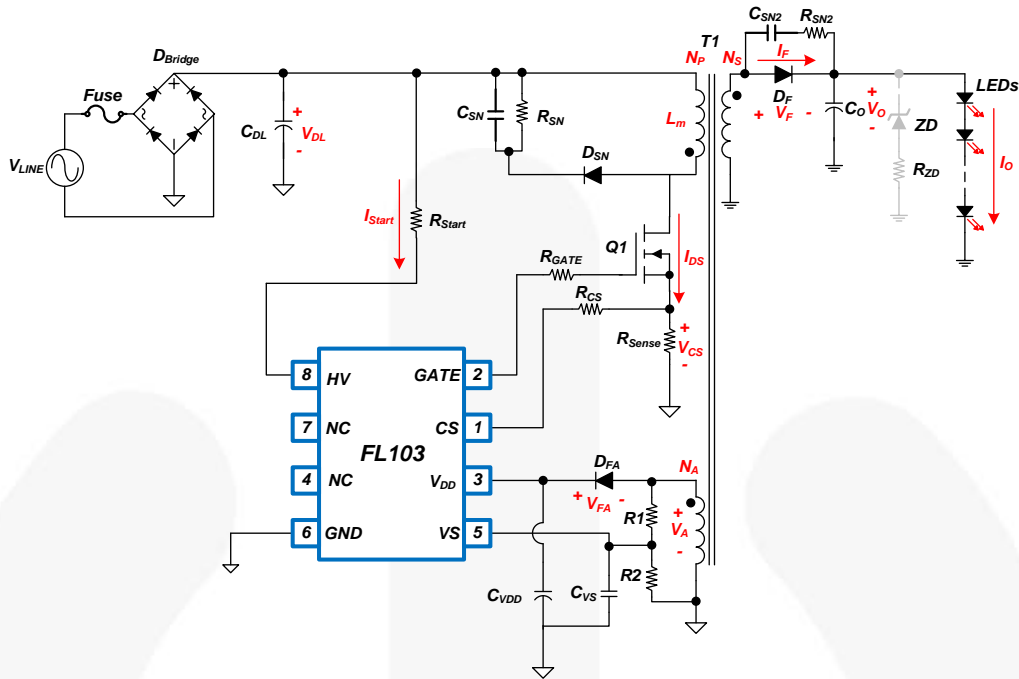


图20. LED 照明用 PSR 反激式转换器基本电路

图20显示原边反馈反激式转换器的基本电路图，图21显示典型波形。一般而言，原边反馈更适合在不连续导通模式 (DCM) 下工作，因为它可实现更佳输出调节。

DCM 反激式转换器的工作原理如下所示：

第一阶段

在 MOSFET 导通期间 (t_{ON})，输入电压 (V_{DD}) 加在初级端电感 (L_m) 两端。随后，MOSFET 电流 (I_{DS}) 由零至峰值 (I_{PK}) 线性上升。在此期间，从输入电源获取能量，并存储在电感中。

第二阶段

当 MOSFET (Q1) 关断时，存储在电感中的能量迫使整流二极管 (D_F) 导通。当二极管导通时，输出电压 (V_O) 以及二极管正向压降 (V_F) 施加到次级端电感两端，二极管电流 (I_F) 由峰值 ($I_{PK} \times N_P/N_S$) 至零线性下降。电感电流放电时间 (t_{DIS}) 结束时，存储在电感中的所有能量都被传输至输出。

第三阶段

当二极管电流达到零时，变压器辅助绕组电压 (V_A) 开始因初级端电感 (L_m) 与 MOSFET (Q1) 等效电容之间的谐振而振荡。

恒压调节

在电感电流放电期间 (t_{DIS})，输出电压 (V_O) 和二极管正向压降 (V_F) 之和反射到辅助绕组端，为 $(V_O + V_F) \times N_A/N_S$ 。由于二极管正向压降 (V_F) 随着电流的减小而减小，辅助绕组电压 (V_A) 在二极管导通时间 (t_{DIS}) 结束时最能准确反映输出电压 (V_O)，此时二极管电流 (I_F) 减小至零。通过在二极管导通时间 (t_{DIS}) 结束时对绕组电压进行采样，可以获得输出电压 (V_O) 信息。用于输出电压调节 (EAV) 的内部误差放大器将采样得到的电压与内部精确参考值进行比较，生成误差电压 (V_{ERR})，该值可确定 MOSFET (Q1) 在恒压模式下的占空比。

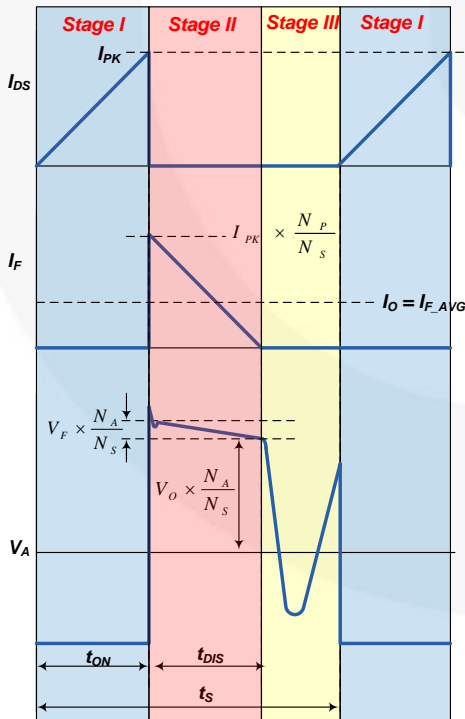


图21. DCM 反激式转换器的波形

恒流调节

输出电流 (I_o) 可以通过峰值漏极电流 (I_{pk}) 和电感电流放电时间 (t_{dis}) 来计算, 这是因为稳定状态时输出电流 (I_o) 与二极管电流 (I_{EAV}) 的平均值相等。输出电流估计量 (I_o 估计量) 通过峰值检测电路确定漏极电流的峰值, 并通过电感放电时间 (t_{dis}) 和开关周期 (t_s) 来计算输出电流 (I_o)。该输出信息同精确的内部参考相比较从而产生一个误差电压 (V_{com}), 它可以确定恒流模式中 MOSFET (Q1) 的占空比。凭借飞兆公司的创新技术 TRUECURRENT®, 恒流输出可以实现精确控制。

电压和电流误差放大器

在两个误差电压 V_{comV} 和 V_{comI} 中, 较小的电压确定占空比。因此, 在恒压控制模式中, V_{comV} 确定占空比, 而 V_{comI} 饱和至高电平。在恒流控制模式期间, V_{comI} 确定占空比, 而 V_{comV} 饱和至高电平。

工作电流

工作电流通常为 3.2 mA。工作电流较低时, 会提高效率, 并且减少对 V_{DD} 电容 (C_{VDD}) 要求。FL103 进入绿色模式后, 工作电流下降至 0.95 mA, 有助于电源满足节能要求。

绿色模式运行

FL103 将电压调节误差放大器输出 (V_{comV}) 用作输出负载指示器, 并调节 PWM 频率, 如图22所示。开关频率随着负载的减少而减小。在重载条件下, 开关频率固定为 50 kHz。一旦 V_{comV} 降至 2.5 V 以下, PWM 频率就会从 50 kHz 线性下降。FL103 进入绿色模式时, PWM 频率会降至最小频率 370 kHz, 从而实现节能, 有助于满足国际节能要求。

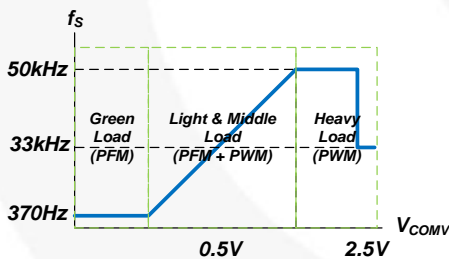


图22. 随输出负载变化的开关频率

逐脉冲限流

当 MOSFET (Q1) 电流检测电阻 (R_{sense}) 两端的电流检测电压 (V_{cs}) 超过内部阈值 0.8 V 时, 在本次剩下的开关周期中 MOSFET (Q1) 被关断。在正常运行中, 逐脉冲限流不会触发, 因为峰值电流由控制环路限制。

前沿消隐 (LEB)

每次功率 MOSFET (Q1) 导通时, 感测电阻 (R_{sense}) 上都会出现一个导通尖峰信号。为了避免过早终止开关脉冲,

抖频

EMI 的减少可通过抖频实现, 它将能量分布在比 EMI 测试设备测得的带宽还要宽的频率范围内。FL103 具有内部抖频电路, 能够在 47 kHz 和 53 kHz 之间改变开关频率。

高压启动

图23显示启动框图。HV 引脚连接至线路输入或直流支撑电容 (C_{DC})。启动期间, 内部启动电路被启用。同时, 线路输入提供电流 (I_{start}), 为 V_{DD} 电容 (C_{VDD}) 充电。当 V_{DD} 电压达到 V_{DD-ON} (16 V) 且 V_{DC} 足够高以避免欠压时, 内部启动电路被禁用, 阻止 I_{start} 流入 HV 引脚。一旦 IC 启动, C_{VDD} 是 PWM 开始开关前唯一提供 IC 消耗电流的能源。因此, C_{VDD} 必须足够大, 以在能量从辅助绕组输入之前阻止 V_{DD-OFF} (7.5 V)。为了避免输入电源浪涌, 在 C_{DC} 和 HV 之间连接 R_{start} , 推荐值为 100 k Ω 。

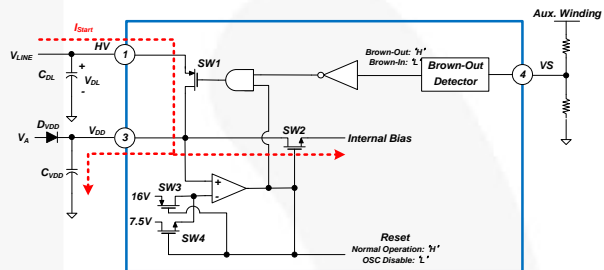


图23. 启动框图

保护

FL103 具有几个自我保护功能, 过压保护、热关断保护、欠压保护和逐脉冲限流。

V_{DD} 欠压锁定 (UVLO)

内部导通和关断阈值分别固定为 16 V 和 7.5 V。在启动期间, V_{DD} 电容 (C_{VDD}) 必须被充电至 16 V。 V_{DD} 电容 (C_{VDD}) 继续提供 V_{DD} 直至主变压器的辅助绕组能提供能量。在启动过程中, V_{DD} 不能低于 7.5 V。在启动过程中, UVLO 滞回窗口确保 V_{DD} 电容 (C_{VDD}) 足够为 V_{DD} 供电。

V_{DD} 过压保护 (OVP)

OVP 功能防止过压状况下的损害。如果在开环反馈状况下 V_{DD} 电压超过 28 V, 则触发 OVP 功能并且禁用 PWM 开关。OVP 设置有保护延迟时间 (典型值 200 μ s), 防止开关噪声引起误触发。

热关断保护 (TSD)

当结温超过 140° C 时, 内置温度检测电路会关断 PWM 输出。有 15° C 的滞回。

内置前沿消隐时间。不必使用传统的 RC 滤波。在消隐时间内, 限流比较器被禁用并且不能关断栅极驱动器。

栅极输出

FL103 输出级为快速图腾柱栅极驱动器。为了最大程度降低热损、增加效率并提高可靠性, 避免出现交叉导通。输出驱动器被一个内部 15 V 的齐纳二极管箝位, 防止功率 MOSFET 晶体管遭受不想要的过压栅极信号。

内置斜率补偿

电流检测电阻两端的检测电压用于电流模式控制和逐脉冲限流。内部斜率补偿能够提高稳定性，并阻止因峰值电流模式控制产生的谐波振荡。FL103 具有内置于每个开关周期中的同步的，正斜率斜坡。

抗噪性

电流检测或控制信号产生的噪声可能导致明显的脉宽抖动，尤其是在连续导通模式下。尽管斜坡补偿有助于缓解这些问题，但仍应该特别谨慎小心。应该遵循良好的布置和布局实践。建议避免长 PCB 引线 and 元件引线，将补偿和滤波元件放置在 FL103 附近，并且增加功率 MOSFET 栅极电阻。

工作区

图24显示工作区。FL103 在恒流模式下具有两个开关频率 (f_s)。一个是 50 kHz。这种情况下，FL103 可以以最佳状态运行，提供 LED 照明。输出电压范围介于正常输出电压 (V_o^N) 和 50% 的正常输出电压 (V_o^N) 之间。另一个是 33 kHz。当输出电压由于负载增加和 LED 数量减少而降低时，输出电压 (V_o) 降低至正常电压 (V_o^N) 的 50% 以下。那时， V_{ov} 跌至欠压锁定保护值附近，并触发保护功能。为了避免 33 kHz 的频率， V_o^N 应该设计有足够的余量。

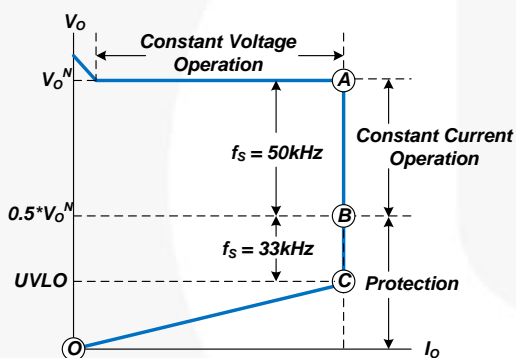


图24. 工作区




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