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# FAN7392

## 大电流、高侧与低侧、栅极驱动 IC

### 特性

- 浮动通道可实现高达 +600V 的自举运行
- 3A/3A 的典型源电流 / 灌电流驱动能力
- 共模 dv/dt 噪声消除电路
- 兼容 3.3V 逻辑输入电平
- 单独逻辑供电 ( $V_{DD}$ )，范围为 3.3V 至 20V
- $V_{CC}$  和  $V_{BS}$  欠压锁定
- 逐周期边沿触发关闭逻辑
- 适用于两个通道的匹配传播延迟
- 输出与输入信号同相
- 采用 14-DIP 和 16-SOP (宽) 封装

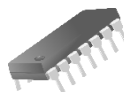
### 应用

- 高速功率 MOSFET 和 IGBT 栅极驱动器
- 服务器电源
- 不间断电源 (UPS)
- 电信系统电源
- 分布式电源
- 电机驱动变频器

### 说明

FAN7392 是单片高侧和低侧栅极驱动 IC，可以驱动工作电压最高达 +600V 的高速 MOSFET 和 IGBT。它具有缓冲输出级，且所有 NMOS 晶体管设计为具有高脉冲电流驱动能力和最低交叠导通。飞兆的高压工艺和共模噪声消除技术可使高侧驱动器在高 dv/dt 噪声环境下稳定运行。先进的电平转换电路，能使高侧栅极驱动器的工作电压在  $V_{BS}=15V$  时  $V_S$  达到 -9.8V (典型值)。逻辑输入与兼容标准 CMOS 或 LSTTL 输出，低至 3.3V 逻辑。UVLO 电路可防止  $V_{CC}$  和  $V_{BS}$  低于指定的阈值电压时发生故障。大电流和低输出压降的特性，使得该器件适合半桥和全桥转换器，如开关电源和大功率 DC-DC 转换器应用。

14-PDIP



16-SOP



### 订购信息

器件编号	工作温度范围	封装	包装方法
FAN7392N	-40°C 至 +125°C	14-PDIP	塑料管
FAN7392M		16-SOP	塑料管
FAN7392MX			卷带和卷盘

典型应用电路图

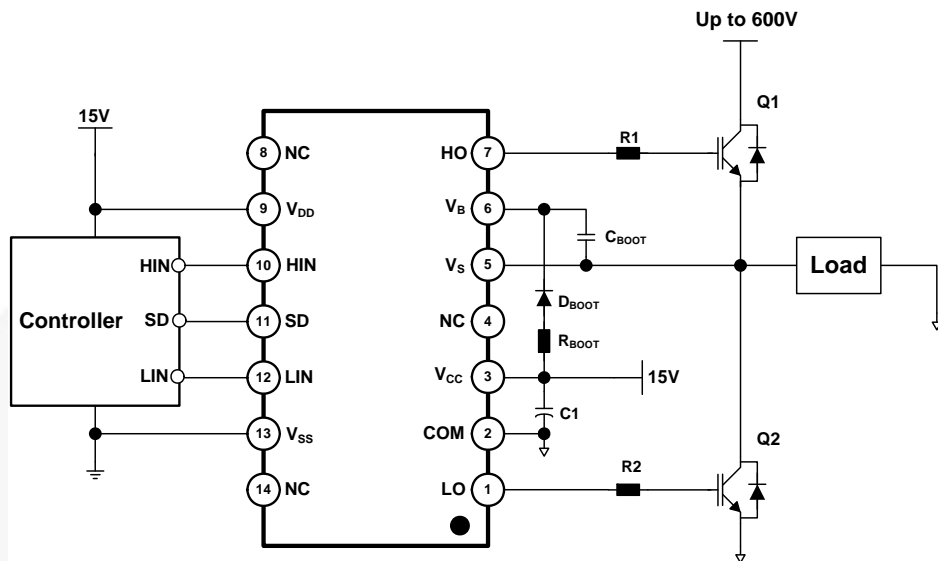


图 1. 典型应用电路（参考 14-DIP）

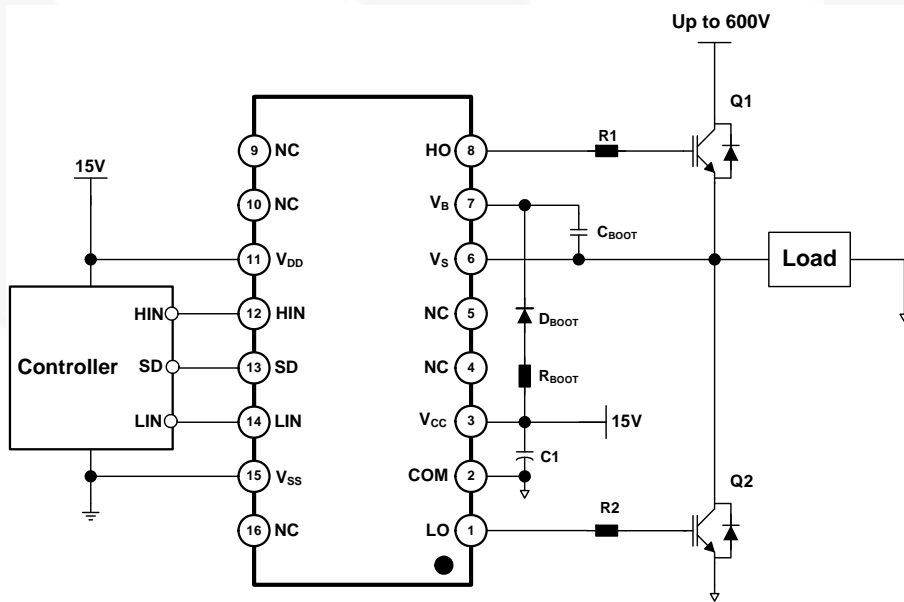


图 2. 典型应用电路（参考 16-SOP）

内部框图

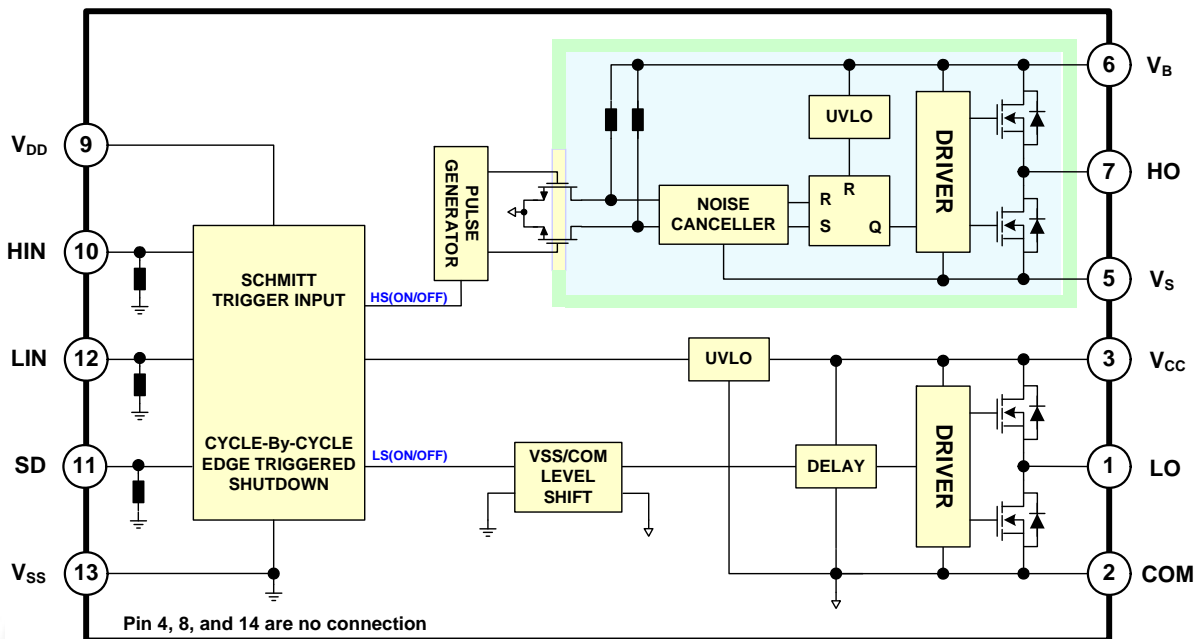


图 3. 功能框图 (参考 14 引脚)

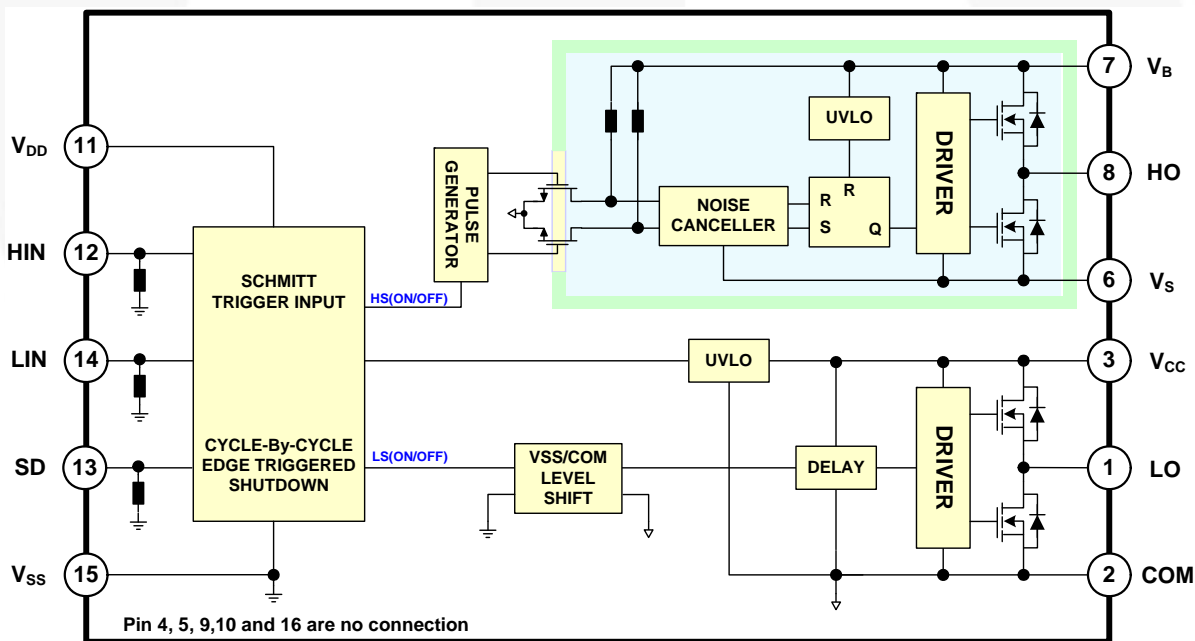


图 4. 功能框图 (参考 16-SOP)

## 引脚布局

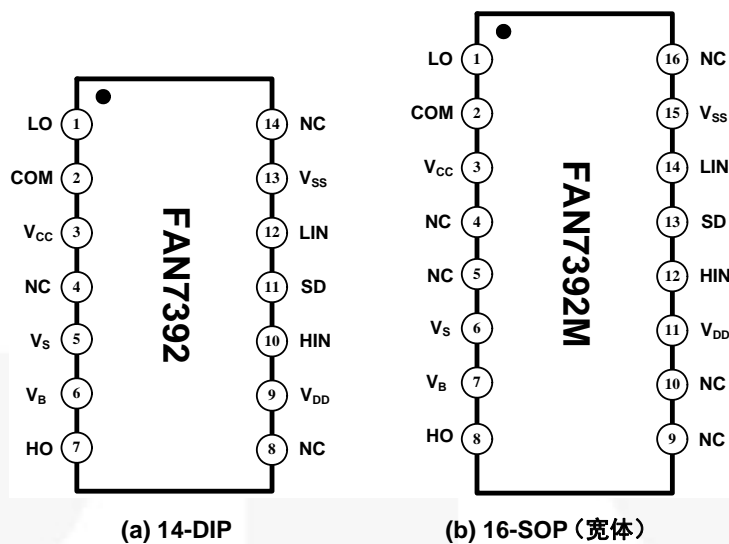


图 5. 引脚布局 (俯视图)

## 引脚定义

14 引脚	16 引脚	名称	说明
1	1	LO	低侧栅极输出
2	2	COM	低侧返回
3	3	V <sub>CC</sub>	低侧电源电压
5	6	V <sub>S</sub>	高侧浮动电源电压返回
6	7	V <sub>B</sub>	高侧浮动电源
7	8	HO	高侧驱动输出
9	11	V <sub>DD</sub>	逻辑电源电压
10	12	HIN	高侧栅极驱动器输出的逻辑输入
11	13	SD	关闭功能逻辑输入
12	14	LIN	低侧栅极驱动器输出的逻辑输入
13	15	V <sub>SS</sub>	逻辑地
4,8,14	4, 5, 9, 10, 16	NC	未连接

## 绝对最大额定值

应力超过绝对最大额定值，可能会损坏器件。在超出推荐的工作条件的情况下，该器件可能无法正常工作，所以不建议让器件在这些条件下长期工作。此外，长期工作在高压推荐的工作条件下工作，会影响器件的可靠性。绝对最大额定值仅是应力规格值。除非另有说明， $T_A = 25^\circ\text{C}$ 。

符号	特性	最小值	最大值	单位
$V_B$	高侧浮动电源电压	-0.3	625.0	V
$V_S$	高侧浮动偏置电压	$V_B-25.0$	$V_B+0.3$	V
$V_{HO}$	高侧浮动输出电压	$V_S-0.3$	$V_B+0.3$	V
$V_{CC}$	低侧电源电压	-0.3	25.0	V
$V_{LO}$	低侧浮动输出电压	-0.3	$V_{CC}+0.3$	V
$V_{DD}$	逻辑电源电压	-0.3	$V_{SS}+25.0$	V
$V_{SS}$	逻辑电源偏置电压	$V_{CC}-25.0$	$V_{CC}+0.3$	V
$V_{IN}$	逻辑输入电压 (HIN、LIN 和 SD)	$V_{SS}-0.3$	$V_{DD}+0.3$	V
$dV_S/dt$	允许的偏置电压变化速率		$\pm 50$	V/ns
$P_D$	功耗 (1, 2, 3)	14-PDIP	1.6	W
		16-SOP	1.3	
$\theta_{JA}$	热阻	14-PDIP	75	$^\circ\text{C/W}$
		16-SOP	95	
$T_J$	最大结温		+150	$^\circ\text{C}$
$T_{STG}$	存储温度	-55	+150	$^\circ\text{C}$

### 注意：

1. 安装到 76.2 x 114.3 x 1.6mm PCB 板 (FR-4 环氧玻璃材料)。
2. 参考下列标准：
  - JESD51-2: 集成电路热测试方法环境条件 – 自然通风和
  - JESD51-3: 含铅表面贴装封装的低有效导热系数测试板
3. 任何情况下都不得超过功耗 ( $P_D$ )。

## 推荐工作条件

推荐的操作条件表明了器件的真实工作条件。指定推荐的工作条件，以确保器件的最佳性能达到数据表中的规格。飞兆不建议超出额定或依照绝对最大额定值进行设计。

符号	参数	最小值	最大值	单位
$V_B$	高侧浮动电源电压	$V_S+10$	$V_S+20$	V
$V_S$	高侧浮动电源偏置电压	$6-V_{CC}$	600	V
$V_{HO}$	高侧输出电压	$V_S$	$V_B$	V
$V_{CC}$	低侧电源电压	10	20	V
$V_{LO}$	低侧输出电压	0	$V_{CC}$	V
$V_{DD}$	逻辑电源电压	$V_{SS}+3$	$V_{SS}+20$	V
$V_{SS}$	逻辑电源偏置电压	-5	5	V
$V_{IN}$	逻辑输入电压	$V_{SS}$	$V_{DD}$	V
$T_A$	工作环境温度	-40	+125	$^\circ\text{C}$

## 电气特性

除非另有说明,  $V_{BIAS}(V_{CC}, V_{BS}, V_{DD})=15.0V$ ,  $V_{SS}=COM=0V$ ,  $T_A=25^{\circ}C$ 。  $V_{IH}$ 、 $V_{IL}$  和  $I_{IN}$  参数以  $V_{SS}$  为参考点, 并适用于相应的输入引脚: HIN、LIN 和 SD。  $V_O$  和  $I_O$  参数以  $V_S$  和 COM 点, 并适用于相应的输出引脚: HO 和 LO。

符号	特性	测试条件	最小值	典型值	最大值	单位
<b>低侧电源部分</b>						
$I_{QCC}$	$V_{CC}$ 静态电源电流	$V_{IN}=0V$ 或 $V_{DD}$		40	80	$\mu A$
$I_{QDD}$	$V_{DD}$ 静态电源电流	$V_{IN}=0V$ 或 $V_{DD}$			10	$\mu A$
$I_{PCC}$	$V_{CC}$ 工作电源电流	$f_{IN}=20kHz$ (均方根), $V_{IN}=15V_{PP}$		430		$\mu A$
$I_{PDD}$	$V_{DD}$ 工作电源电流	$f_{IN}=20kHz$ (均方根), $V_{IN}=15V_{PP}$		300		$\mu A$
$I_{SD}$	关闭电源电流	$S_D=V_{DD}$		120		$\mu A$
$V_{CCUV+}$	$V_{CC}$ 电源欠压正向阈值电压	$V_{IN}=0V$ , $V_{CC}$ = 扫描	7.7	8.8	9.9	V
$V_{CCUV-}$	$V_{CC}$ 电源欠压负向阈值电压	$V_{IN}=0V$ , $V_{CC}$ = 扫描	7.3	8.4	9.5	V
$V_{CCUVH}$	$V_{CC}$ 电源欠压锁定滞回电压回差	$V_{IN}=0V$ , $V_{CC}$ = 扫描		0.4		V
<b>自举电源部分</b>						
$I_{QBS}$	$V_{BS}$ 静态电源电流	$V_{IN}=0V$ 或 $V_{DD}$		60	130	$\mu A$
$I_{PBS}$	$V_{BS}$ 工作电源电流	$f_{IN}=20kHz$ (均方根值)		500		$\mu A$
$V_{BSUV+}$	$V_{BS}$ 电源欠压正向阈值电压	$V_{IN}=0V$ , $V_{BS}$ = 扫描	7.7	8.8	9.9	V
$V_{BSUV-}$	$V_{BS}$ 电源欠压负向阈值电压	$V_{IN}=0V$ , $V_{BS}$ = 扫描	7.3	8.4	9.5	V
$V_{BSUVH}$	$V_{BS}$ 电源欠压锁定滞回电压回差	$V_{IN}=0V$ , $V_{BS}$ = 扫描		0.4		V
$I_{LK}$	偏置漏电流	$V_B=V_S=600V$			50	$\mu A$
<b>输入逻辑部分 (HIN、LIN 和 SD)</b>						
$V_{IH}$	逻辑“1”输入阈值电压	$V_{DD}=3V$	2.4			V
		$V_{DD}=15V$	9.5			V
$V_{IL}$	逻辑“0”输入阈值电压	$V_{DD}=3V$			0.8	V
		$V_{DD}=15V$			4.5	V
$I_{IN+}$	逻辑输入高偏置电流	$V_{IN}=V_{DD}$		20	40	$\mu A$
$I_{IN-}$	逻辑输入低电平偏置电流	$V_{IN}=0V$			3	$\mu A$
$R_{IN}$	逻辑输入下拉电阻		375	750		$K\Omega$
<b>栅极驱动器输出部分</b>						
$V_{OH}$	高电平输出电压 ( $V_{BIAS} - V_O$ )	空载 ( $I_O=0A$ )			1.5	V
$V_{OL}$	低电平输出电压	空载 ( $I_O=0A$ )			200	mV
$I_{O+}$	输出高电平、短路脉冲电流 <sup>(4)</sup>	$V_O=0V$ , $PW \leq 10\mu s$	2.5	3.0		A
$I_{O-}$	输出低电平、短路脉冲电流 <sup>(4)</sup>	$V_O=15V$ , $PW \leq 10\mu s$	2.5	3.0		A
$V_{SS}/COM$	$V_{SS}-COM/COM-V_{SS}$ 电压承受		-5.0		5.0	V
$-V_S$	IN 信号传播到 HO 时允许的 $V_S$ 引脚负电压			-9.8	-7.0	V

**动态电气特性**

除非另有说明,  $V_{BIAS}(V_{CC}, V_{BS}, V_{DD})=15.0V$ ,  $V_{SS}=COM=0V$ ,  $C_{LOAD}=1000pF$ ,  $T_A=25^{\circ}C$ 。

符号	参数	工作条件	最小值	典型值	最大值	单位
$t_{on}$	导通传播延迟时间	$V_S=0V$		130	180	ns
$t_{off}$	关断传播延迟时间	$V_S=0V$		150	200	ns
$t_{sd}$	关闭传播延迟时间 <sup>(4)</sup>			130	180	ns
$t_r$	导通上升时间			25	50	ns
$t_f$	关断下降时间			20	45	ns
MT	延迟匹配, HO 和 LO 开启 / 关断				35	ns

**注:**

4. 这些参数由设计保证。



典型特性

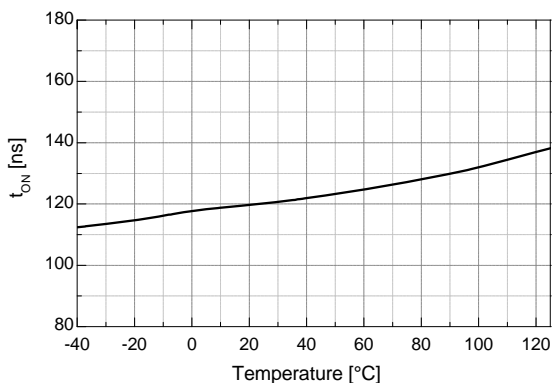


图 6. 导通传播延时与温度的关系

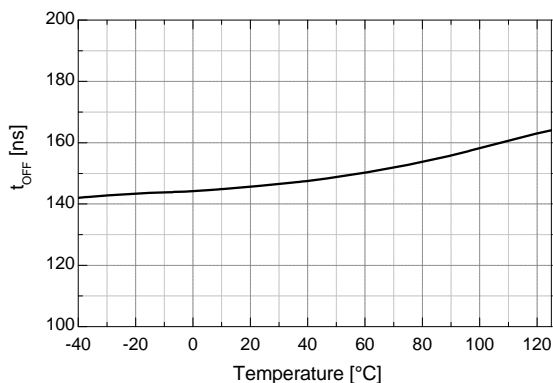


图 7. 关断传播延迟与温度的关系

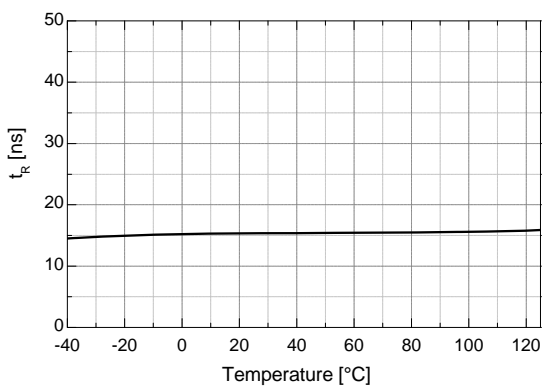


图 8. 导通上升时间与温度的关系

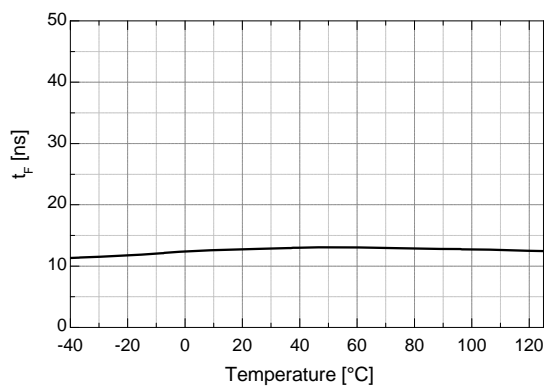


图 9. 关断下降时间与温度的关系

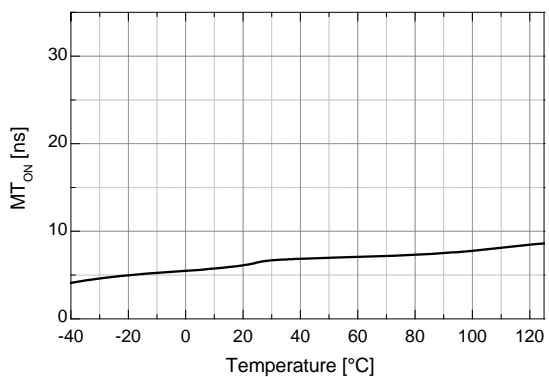


图 10. 开启延迟匹配与温度的关系

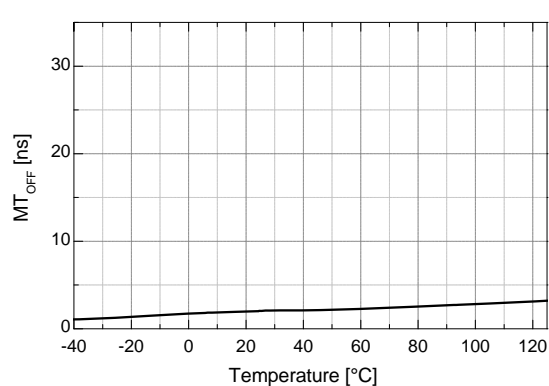


图 11. 关断延迟匹配与温度的关系

典型特性 (续)

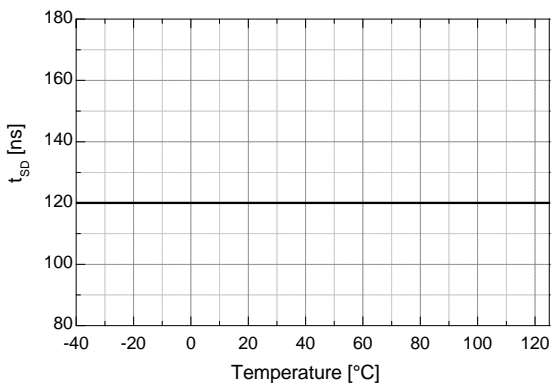


图 12. 关闭传播延迟与温度的关系

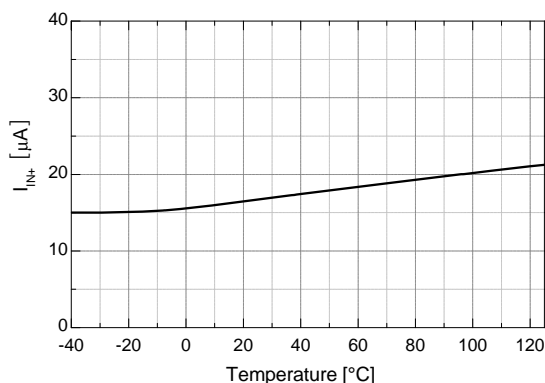


图 13. 逻辑输入高电平偏置电流与温度的关系

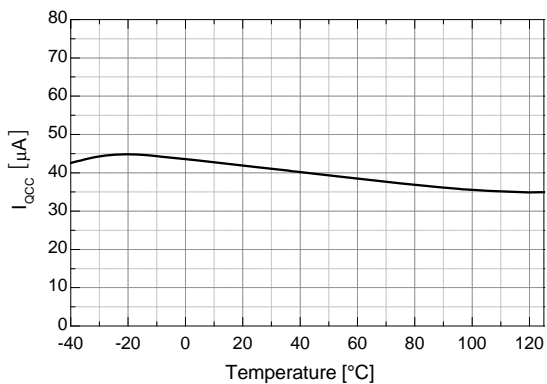


图 14. V<sub>CC</sub> 静态电源电流与温度的关系

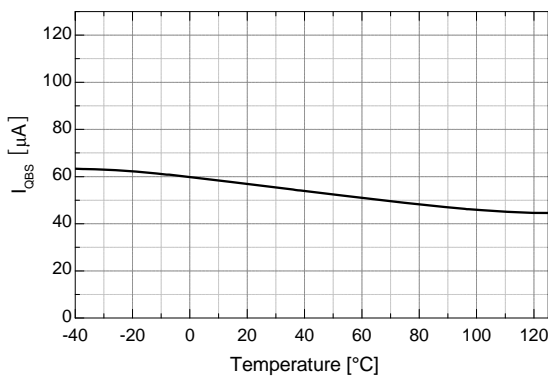


图 15. V<sub>BS</sub> 静态电源电流与温度的关系

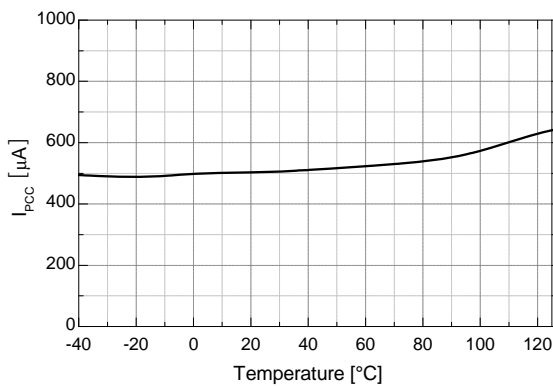


图 16. V<sub>CC</sub> 工作电源电流与温度的关系

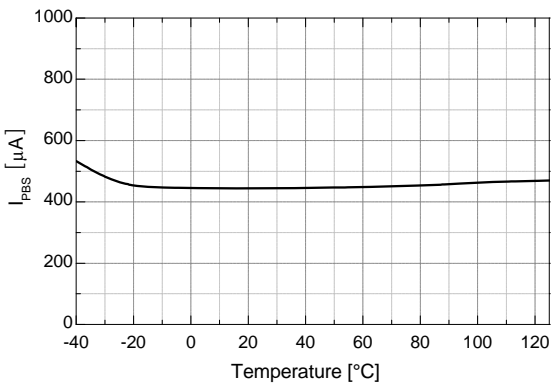


图 17. V<sub>BS</sub> 工作电源电流与温度的关系

典型特性 (续)

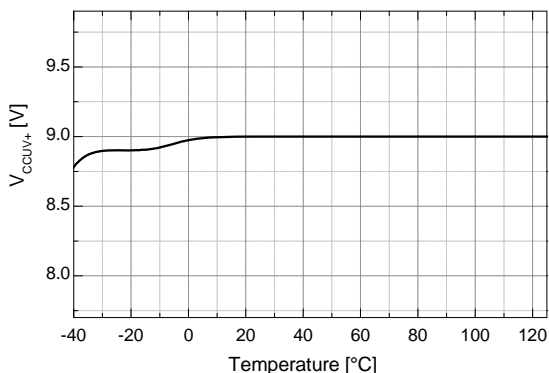


图 18.  $V_{CC}$  UVLO+ 与温度的关系

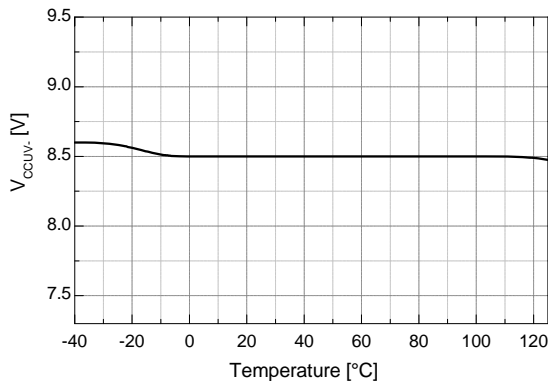


图 19.  $V_{CC}$  UVLO- 与温度的关系

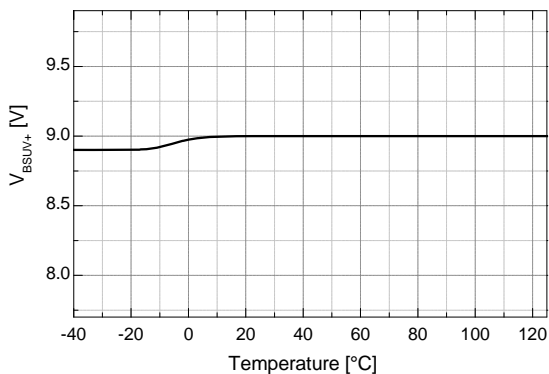


图 20.  $V_{BS}$  UVLO+ 与温度的关系

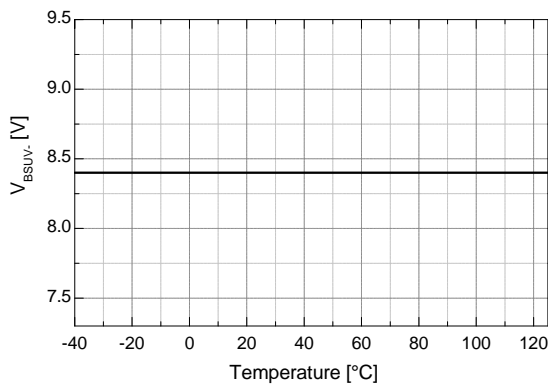


图 21.  $V_{BS}$  UVLO- 与温度的关系

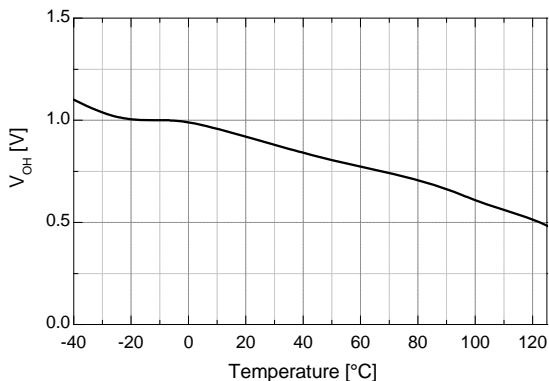


图 22. 高电平输出电压与温度的关系

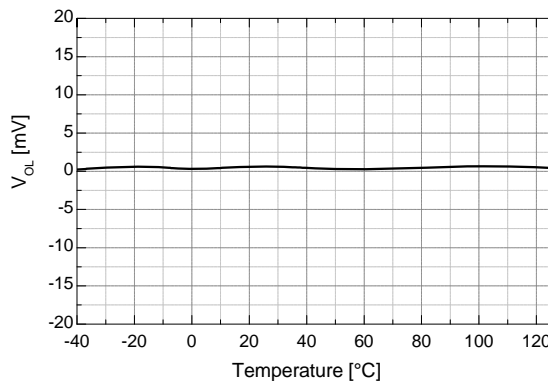


图 23. 低电平输出电压与温度的关系

典型特性 (续)

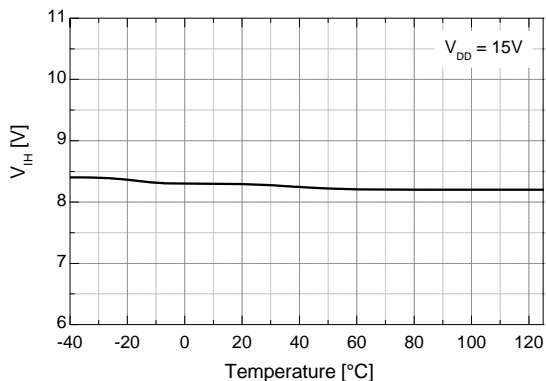


图 24. 逻辑高输入电压与温度的关系

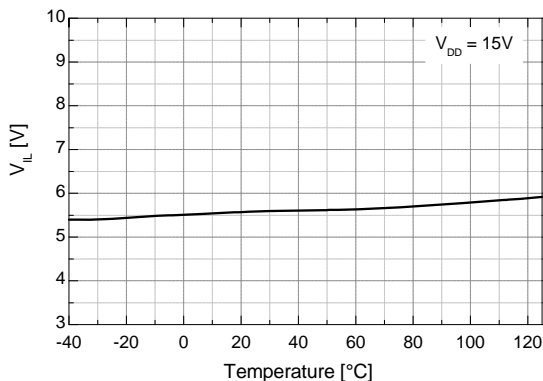


图 25. 逻辑低输入电压与温度的关系

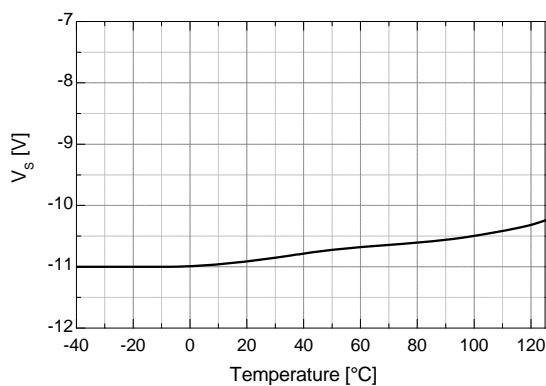


图 26. 允许的负 Vs 电压与温度的关系

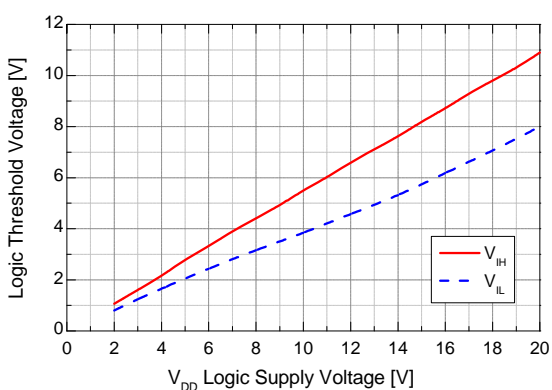


图 27. 输入逻辑 (HIN 和 LIN) 阈值电压与 V<sub>DD</sub> 电源电压的关系

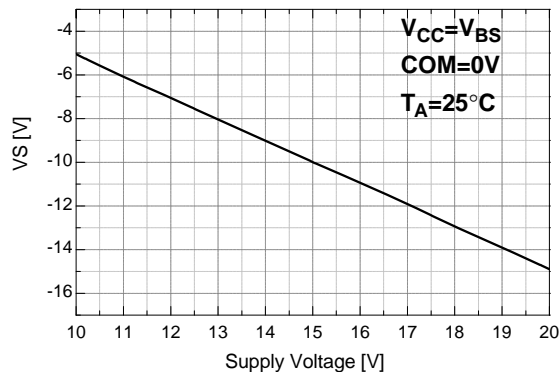


图 28. HIN 信号传播至高侧允许的 Vs 负电压与电源电压的关系

### 开关时间定义

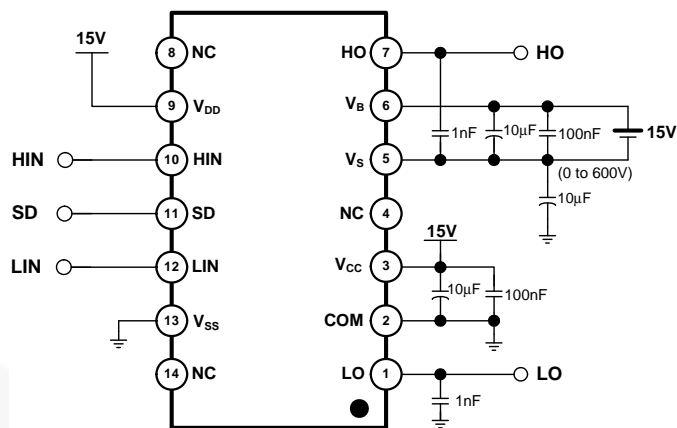


图 29. 开关时间测试电路 (参考 14-DIP)

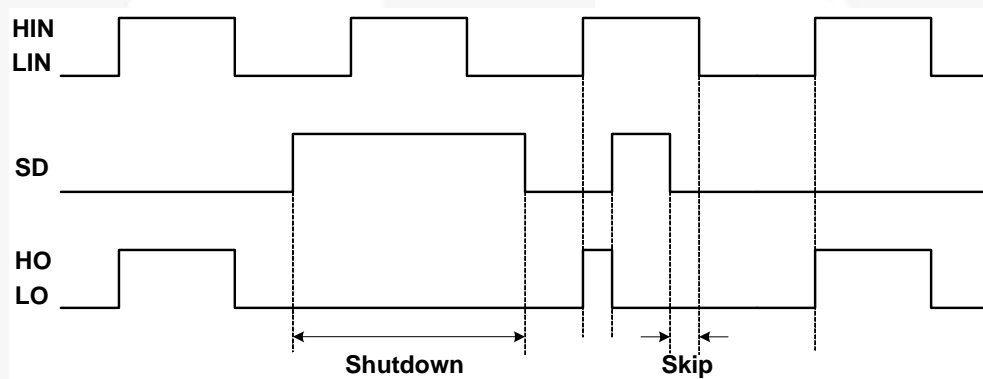


图 30. 输入 / 输出时序图

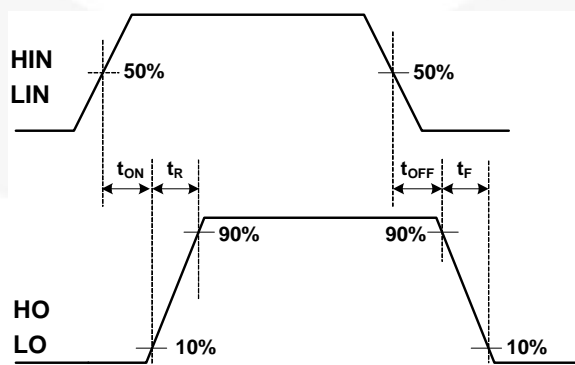


图 31. 开关时间波形定义

开关时间定义 (续)

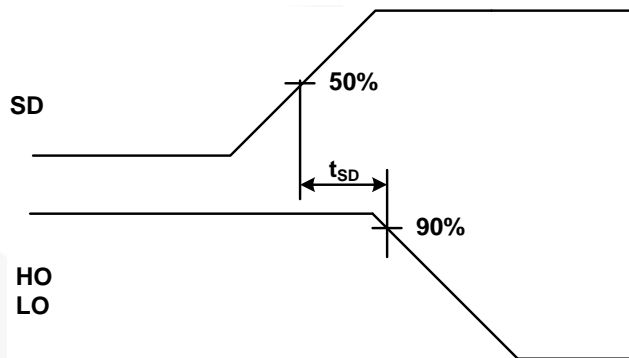


图 32. 关闭波形定义

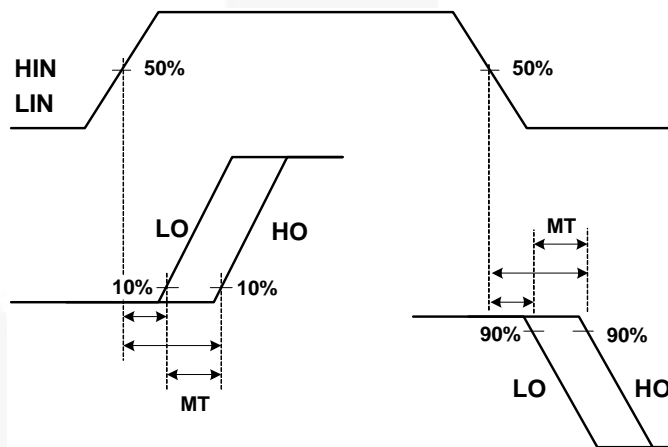


图 33. 开关匹配波形定义

## 应用信息

### 瞬变 $V_S$ 负电压

自举式电路具有简单和低成本的优点，但是，它也有一些局限。此电路的最大难题是，在半桥应用中，高侧的开关器件关断时在其发射极存在负电压。

如果高侧开关 Q1 关断，同时负载电流流向电感负载；从高侧开关 Q1 至二极管 D2（与同一逆变器桥臂的低侧开关并联）出现电流转移。然后，负电压出现在高侧开关器件的发射极，在续流二极管开始箝位前，负载电流突然流向低侧续流二极管 D2，如图 34 所示。

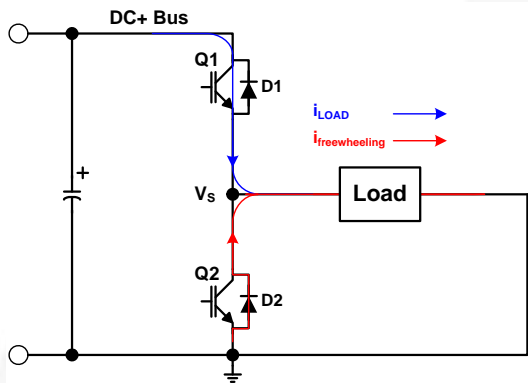


图 34. 半桥应用电路

此负电压给栅极驱动器输出时带来麻烦。很可能产生自举电容过压的情况，输入信号丢失以及闭锁问题，因为它直接影响栅极驱动器的电源  $V_S$  引脚，如图 35 所示。该下冲电压称为“瞬变  $V_S$  负电压”。

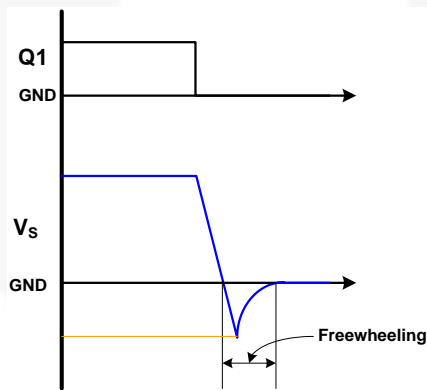


图 35. Q1 关断时的  $V_S$  波形

图 36 和图 37 显示高侧开关 Q1 和低侧续流二极管 D3（同一逆变器桥臂）之间的负载电流换流。逆变电路中从芯片引脚绑定到 PCB 走线的寄生电感对于每个 IGBT 的就是  $L_C$  和  $L_E$ 。当高侧开关 Q1 和低侧开关 Q4 导通后， $V_{S1}$  节点电压稍低于 DC+ 电压，压差与电路的功率开关以及寄生电感有关（负载电流从 Q1 和 Q4 流出，如图 36 所示）。若高侧开关 Q1 关断且 Q4 保持导通状态，则负载电流流入低侧续流二极管 D3（感性负载连接  $V_{S1}$ ，如图 37 所示）。电流从地（连接栅极驱动器的 COM 引脚）流入负载，高侧开关器件的发射极为负电压。

这种情况下，栅极驱动器的 COM 引脚电位高于  $V_S$  引脚，因为续流二极管 D3 有压降，寄生元件  $L_{C3}$ 、 $L_{E3}$  也有压降。

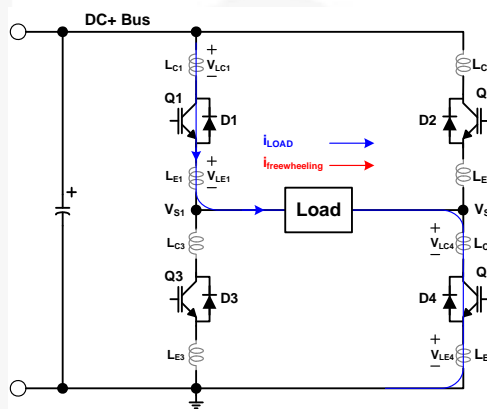


图 36. Q1 和 Q4 导通

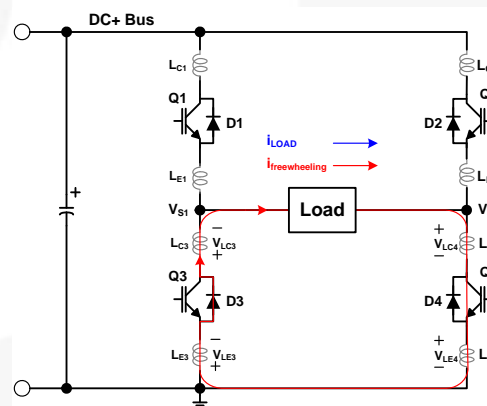


图 37. Q1 关断和 D3 导通

FAN7392 的瞬态  $V_S$  负电压曲线如图 38 所示。

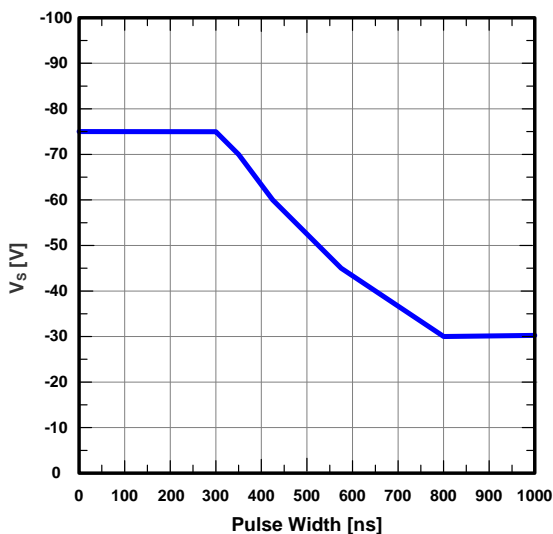


图 38. 瞬态  $V_S$  负电压特性

虽然 FAN7392 已表现出可处理这些瞬态  $V_S$  负电压条件的能力，仍然强烈建议电路设计人员谨慎进行 PCB 布局，尽量限制负  $V_S$ ，最大程度降低寄生元件的数值以及所用元器件的数量。负  $V_S$  电压幅度与寄生电感以及开关器件的关断速度 ( $di/dt$ ) 成比例。

## 一般准则

### 印刷电路板布局

最大程度降低元件寄生特性的布局建议如下：

- 开关之间的走线没有回路或偏差。
- 避免互连链路。它会显著增加电感。
- 降低封装体距离 PCB 板的高度，以减少引脚电感效应。
- 考虑所有功率开关的配合放置，以减少走线长度。
- 若要尽可能减少噪声耦合，接地层不应放置在高侧浮动侧的下方，或位于高侧浮动侧附近。
- 若要减少 EM 耦合并改善功率开关的导通/关断性能，必须尽可能缩短栅极驱动环路。

## 元器件放置

元器件的位置与选型建议如下：

- 在  $V_{DD}$  和  $V_{SS}$  引脚之间放置一个旁路电容。1 $\mu$ F 电容适用于大部分应用。该元件应尽可能靠近引脚放置，以便降低寄生元素。
- $V_{CC}$  和 COM 之间的旁路电容同时为低侧驱动器和自举电容的再充电供电。建议该电容值至少是自举电容的十倍以上。
- 在量化自举电阻和初次自举充电时的电流时，必须考虑自举电阻  $R_{BOOT}$ 。若需要将电阻与自举二极管串联连接，请验证  $V_B$  不低于 COM（地）。建议使用典型 5 ~ 10 $\Omega$ ，可增加  $V_{BS}$  时间常数。如果自举电阻和二极管的压降过高，或电路拓扑不允许足够的充电时间，则可以使用快恢复或超快恢复二极管。
- 自举电容  $C_{BOOT}$  使用一个低 ESR 电容，比如陶瓷电容。

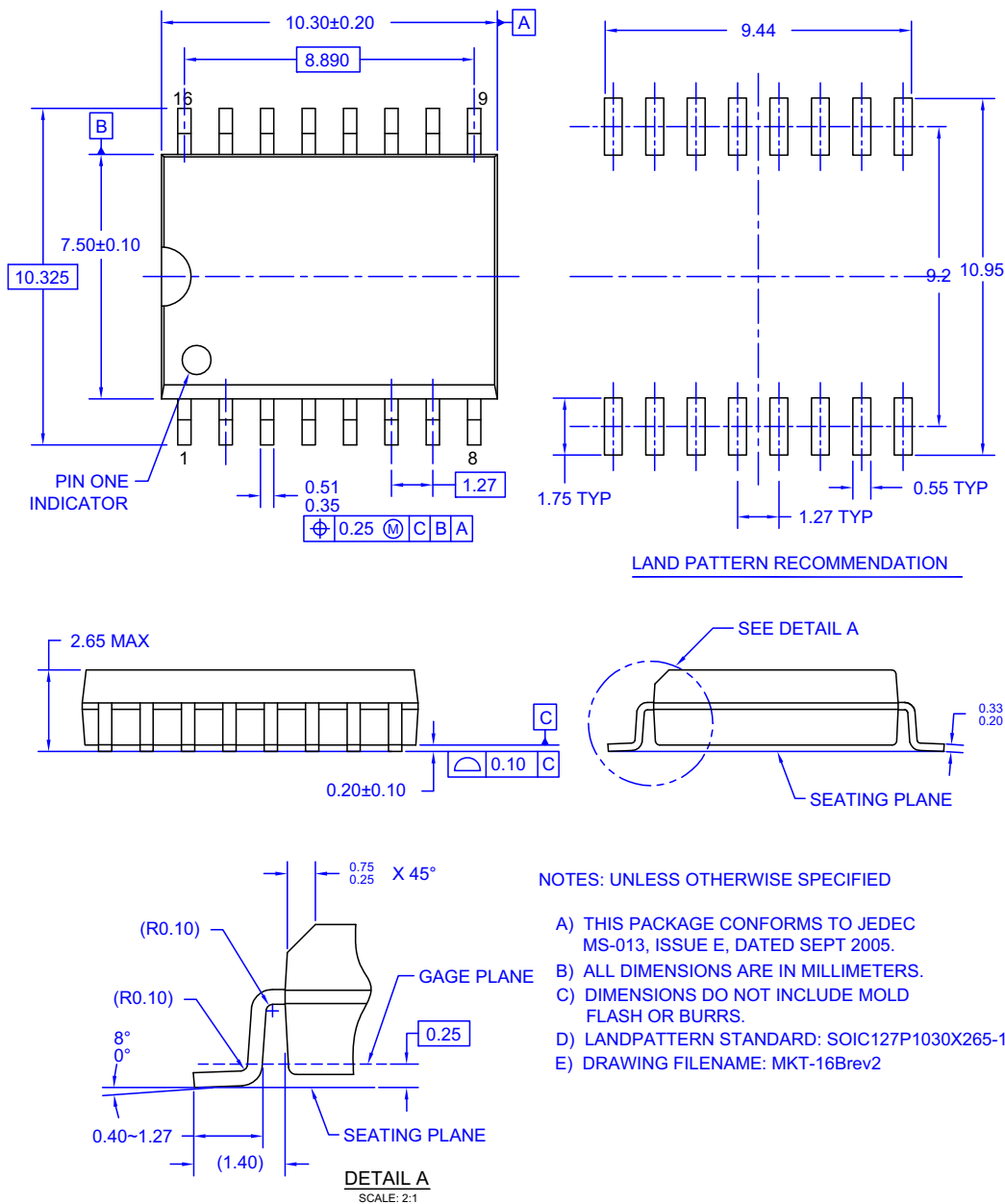
强烈建议元器件如下放置：

- 与浮动电压引脚 ( $V_B$  和  $V_S$ ) 相连的元器件放置在器件以及 FAN7392 的对应高压部分附近。该封装中的 NC（未连接）引脚使高压引脚和低压引脚之间的距离最大化（见图 5）。
- 旁路电容和栅极电阻的布局和布线应尽可能靠近栅极驱动 IC。
- 自举二极管  $D_{BOOT}$  应尽量靠近自举电容  $C_{BOOT}$  放置。
- 自举二极管必须使用较低的正向压降，为了快速恢复，开关时间必须尽可能短，如超快恢复二极管。





物理尺寸 (续)



M16BREV2

图 40.16 引脚小尺寸封装 (SOP)

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