# Qi Compliant Wireless Power Transmitter Manager 

Check for Samples: bq500110

## FEATURES

- Intelligent Control of the Power Transfer between Base Station and Mobile Device
- Conforms to Version 1.0 of the Wireless Power Consortium (WPC) Transmitter Specifications
- Demodulates and Decodes WPC Complaint Message Packets from the Power Receiving Device Over the Same Wireless Link that Transfers Electrical Power
- Implements Closed-Loop Power Transfer PID Control by Varying Frequency of the Voltage on the Transmitting Coil
- Parasitic Metal Object Detection (PMOD)
- Operating Modes Status Indicators
- Standby
- Power Transfer
- Charge Complete
- Fault
- Overload and Over Temperature Protection


## APPLICATIONS

- WPC 1.0 Compliant Wireless Chargers for:
- Mobile and Smart Phones
- MP3 Players
- Global Positioning Devices
- Digital Cameras
- Other Wireless Power Base Stations and Transmitters in:
- Cars and Other Vehicles
- Hermetically Sealed Devices, Tools and Appliances
- Furniture Built-in Wireless Chargers
- Toy Power Supplies and Chargers


## DESCRIPTION

The bq500110 is the industry's first wireless power dedicated digital controller that integrates most of the logic functions required to control Wireless Power Transfer in a single channel WPC compliant contactless charging base station. The bq500110 is an intelligent device that periodically pings the surrounding environment for available devices to be powered; monitors all communication from the mobile device being wirelessly powered; adjusts power applied to the transmitter coil per feedback received from the powered device. The bq500110 also manages fault conditions associated with power transfer and controls the operating modes status indicator. The bq500110 is also the first wireless power controller with parasitic metal object detection that in real time analyzes efficiency of the established power transfer and protects itself and the power receiver from excessive power loss and heat associated with parasitic metal objects placed in the power transfer path.
The bq500110 comes in the area saving 48 -pin, $7 \mathrm{~mm} \times 7 \mathrm{~mm}$ QFN package and operates over temperature range from $-40^{\circ} \mathrm{C}$ to $110^{\circ} \mathrm{C}$.

ORDERING INFORMATION ${ }^{(1)}$

| OPERATING <br> TEMPERATURE <br> RANGE, $\mathbf{T}_{\mathbf{A}}$ | ORDERABLE PART NUMBER | PIN COUNT | SUPPLY | PACKAGE | TOP SIDE <br> MARKING |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $40^{\circ} \mathrm{C}$ to $110^{\circ} \mathrm{C}$ | bq500110RGZR | 48 pin | Reel of 2500 | QFN | bq500110 |
|  | bq500110RGZT | 48 pin | Reel of 250 | QFN | bq500110 |

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.


## ABSOLUTE MAXIMUM RATINGS ${ }^{(1)}$

over operating free-air temperature range (unless otherwise noted)

|  | VALUE |  | UNIT |
| :--- | :---: | :---: | :---: |
|  | MIN | MAX |  |
| Voltage applied at V33D to DGND | -0.3 | 3.8 | V |
| Voltage applied at V33A to AGND | -0.3 | 3.8 | V |
| Voltage applied to any pin ${ }^{(2)}$ | -0.3 | 3.8 | V |
| Storage temperature, $T_{\text {STG }}$ | -40 | 150 | ${ }^{\circ} \mathrm{C}$ |

(1) Stresses beyond those listed under absolute maximum ratingsmay cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditionsis not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltages referenced to GND.

## RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

|  |  | MIN | NOM | MAX |
| :--- | ---: | ---: | ---: | :---: |
| UNIT |  |  |  |  |
| V | Supply voltage during operation, V33D, V33A | 3.0 | 3.3 | 3.6 |
| $\mathrm{~T}_{\mathrm{A}}$ | Operating free-air temperature range ${ }^{(1)}$ | -40 | 125 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{J}}$ | Junction temperature ${ }^{(1)}$ |  | 125 | ${ }^{\circ} \mathrm{C}$ |

(1) When operating continuously, the bq500110's typical power consumption causes a $15^{\circ} \mathrm{C}$ temperature rise from ambient.

## ELECTRICAL CHARACTERISTICS

over operating free-air temperature range (unless otherwise noted)

|  | PARAMETER | TEST CONDITIONS | MIN | NOM | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SUPPLY CURRENT |  |  |  |  |  |  |
| $\mathrm{I}_{\text {V33 }}$ | Supply current | $V 33 \mathrm{~A}=3.3 \mathrm{~V}$ |  | 8 | 15 | mA |
| $\mathrm{I}_{\mathrm{V} 33 \mathrm{D}}$ |  | $\mathrm{V} 33 \mathrm{D}=3.3 \mathrm{~V}$ |  | 42 | 55 |  |
| $\mathrm{I}_{\text {V33D }}$ |  | $\mathrm{V} 33 \mathrm{D}=3.3 \mathrm{~V}$ while storing configuration parameters in flash memory |  | 53 | 65 |  |
| INTERNAL REGULATOR CONTROLLER INPUTS/OUTPUTS |  |  |  |  |  |  |
| V33 | 3.3-V linear regulator | Emitter of NPN transistor | 3.25 | 3.3 | 3.6 | V |
| V33FB | 3.3-V linear regulator feedback |  |  | 4 | 4.6 |  |
| $\mathrm{I}_{\text {V33FB }}$ | Series pass base drive | $\mathrm{V}_{\mathrm{IN}}=12 \mathrm{~V}$; current into V33FB pin |  | 10 |  | mA |
| Beta | Series NPN pass device |  | 40 |  |  |  |
| EXTERNALLY SUPPLIED 3.3 V POWER |  |  |  |  |  |  |
| V33D | Digital 3.3-V power | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | 3 |  | 3.6 | V |
| V33A | Analog 3.3-V power | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | 3 |  | 3.6 | V |
| V33Slew | V33 slew rate | V 33 slew rate between 2.3 V and 2.9 V , $\mathrm{V} 33 \mathrm{~A}=\mathrm{V} 33 \mathrm{D}$ | 0.25 |  |  | V/ms |
| MODULATION AMPLIFIER INPUTS EAP-A, EAN-A, EAP-B, EAN-B |  |  |  |  |  |  |
| $\mathrm{V}_{\text {CM }}$ | Common mode voltage each pin |  | -0.15 |  | 1.631 | V |
| EAP-EAN | Modulation voltage digital resolution |  |  | 1 |  | mV |
| $\mathrm{R}_{\text {EA }}$ | Input Impedance | Ground reference | 0.5 | 1.5 | 3 | $\mathrm{M} \Omega$ |
| I OFFSET | Input offset current | $1 \mathrm{k} \Omega$ source impedance | -5 |  | 5 | $\mu \mathrm{A}$ |

ANALOG INPUTS V_IN, I_IN, TEMP_IN, I_COIL, LED_MODE, PMOD_THR

| $\mathrm{V}_{\text {ADDR_OPEN }}$ | Voltage indicating open pin | LED_MODE, PMOD_THR open | 2.37 |  | V |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {ADDR_SHORT }}$ | Voltage indicating pin shorted to GND | LED_MODE, PMOD_THR shorted to ground |  | 0.36 | V |
| $\mathrm{V}_{\text {ADC_RANGE }}$ | Measurement range for voltage monitoring | Inputs: V_IN, I_IN, TEMP_IN, I_COIL | 0 | 2.5 | V |
| INL | ADC integral nonlinearity |  | -2.5 | 2.5 | mV |
| $\mathrm{I}_{\mathrm{kg}}$ | Input leakage current | 3 V applied to pin |  | 100 | nA |
| $\mathrm{R}_{\text {IN }}$ | Input impedance | Ground reference | 8 |  | $\mathrm{M} \Omega$ |
| $\mathrm{C}_{\text {IN }}$ | Input capacitance |  |  | 10 | pF |

## DIGITAL INPUTS/OUTPUTS

| $\mathrm{V}_{\mathrm{OL}}$ | Low-level output voltage | $\mathrm{I}_{\mathrm{OL}}=6 \mathrm{~mA}{ }^{(1)}, \mathrm{V} 33 \mathrm{D}=3 \mathrm{~V}$ |  | $\begin{array}{r} \text { DGND } \\ 1+0.25 \end{array}$ | V |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ | High-level output voltage | $\mathrm{I}_{\mathrm{OH}}=-6 \mathrm{~mA}{ }^{(2)}, \mathrm{V} 33 \mathrm{D}=3 \mathrm{~V}$ | $\begin{aligned} & \text { V33D } \\ & -0.6 \mathrm{~V} \end{aligned}$ |  | V |
| $\mathrm{V}_{\mathrm{IH}}$ | High-level input voltage | $V 33 \mathrm{D}=3 \mathrm{~V}$ | 2.1 | 3.6 | V |
| $\mathrm{V}_{\text {IL }}$ | Low-level input voltage | $\mathrm{V} 33 \mathrm{D}=3.5 \mathrm{~V}$ |  | 1.4 | V |
| $\mathrm{I}_{\mathrm{OH}}$ (MAX) | Output high source current |  |  | 4 | mA |
| $\mathrm{I}_{\text {OL }}$ (MAX) | Output low sink current |  |  | 4 | mA |
| SYSTEM PERFORMANCE |  |  |  |  |  |
| $\mathrm{V}_{\text {RESET }}$ | Voltage where device comes out of reset | V33D Pin | 2.3 | 2.4 | V |
| $\mathrm{t}_{\text {RESET }}$ | Pulse width needed for reset | RESET pin | 2 |  | $\mu \mathrm{s}$ |
| $\mathrm{F}_{\text {SW }}$ | Switching Frequency |  | 110 | 205 | kHz |
| $\mathrm{t}_{\text {detect }}$ | Time to detect presence of device requesting power |  |  | 0.6 | sec |
| $\mathrm{t}_{\text {retention }}$ | Retention of configuration parameters | $\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C}$ | 100 |  | Years |
| Write_Cycles | Number of nonvolatile erase/write cycles | $\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C}$ | 20 |  | K cycles |

(1) The maximum $\mathrm{I}_{\mathrm{OL}}$, for all outputs combined, should not exceed 12 mA to hold the maximum voltage drop specified.
(2) The maximum $\mathrm{I}_{\mathrm{OH}}$, for all outputs combined, should not exceed 48 mA to hold the maximum voltage drop specified.
(3) With default device calibration. PMBus calibration can be used to improve the regulation tolerance.
(4) Time from close of error ADC sample window to time when digitally calculated control effort (duty cycle) is available. This delay must be accounted for when calculating the system dynamic response. Includes EADC conversion time.

## DEVICE INFORMATION

## Functional Block Diagram




PIN FUNCTIONS

| PIN |  | I/O | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| NO. | NAME |  |  |
| 1 | AIN5 | I | Connect this pin to GND |
| 2 | AIN6 | 1 | Connect this pin to GND |
| 3 | TEMP_IN | I | Thermal protection Input |
| 4 | V_IN | 1 | Input-voltage ADC Input |
| 5 | RESET | 1 | Device reset |
| 6 | COMM1 | I | Primary communication channel |
| 7 | COMM2 | 1 | Alternate communication channel |
| 8 | COIL_DIS | 1 | Coil disable |
| 9 | DRV_SEL | 1 | Gate Driver mode select |
| 10 | PMB_SCI | I/O | Optional programming I/O. Pull up to $\mathrm{V}_{C C}$ via $5.1 \mathrm{k} \Omega$ resistor. |
| 11 | PMB_SDA | 1/O | Optional programming I/O. Pull up to $\mathrm{V}_{\mathrm{CC}}$ via $5.1 \mathrm{k} \Omega$ resistor. |
| 12 | COIL_PWM | 0 | PWM Output |
| 13 | DRV_EN | 0 | PWM Enable Output |
| 14 | LOGO1 | 0 | Optional Logic Output. Leave this pin floating. |
| 15 | BUZ_DC | O | DC Buzzer Output |
| 16 | LED2 / SS | 0 | LED Drive Output 2 / Slave Select output |

PIN FUNCTIONS (continued)

| PIN |  | I/O | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| NO. | NAME |  |  |
| 17 | LED1 / SCLK | 0 | LED Drive Output 1 / Serial Clock Output |
| 18 | DIAG_DIS | I/O | Disable Diagnostic Output. Leave this pin floating to inhibit diagnostic. |
| 19 | PMB_ALERT | I/O | Optional Programming I/O. Connect to GND. |
| 20 | PMB_CTRL | I/O | Optional programming I/O. Pull up to $\mathrm{V}_{\mathrm{CC}}$ via $5.1 \mathrm{k} \Omega$ resistor. |
| 21 | SCI-TX | I/O | Optional Programming I/O. Leave floating. |
| 22 | SCI-RX | I/O | Optional Programming I/O. Leave floating. |
| 23 | WDO | 0 | External Watchdog Output |
| 24 | BUZ_AC | 0 | AC Buzzer Output |
| 25 | MOSI | I/O | Master Out Slave In |
| 26 | LED3/MISO/NO_SUP | I/O | LED Drive Output 3 / Master In Slave Out / Select stand alone operation (no supervisor) |
| 27 | TCK | I/O | Optional Programming I/O. Leave floating. |
| 28 | TDO | I/O | Optional Programming I/O. Leave floating. |
| 29 | TDI | I/O | Optional programming I/O. Pull up to $\mathrm{V}_{\mathrm{CC}}$ via $5.1 \mathrm{k} \Omega$ resistor. |
| 30 | TMS | I/O | Optional programming I/O. Pull up to $\mathrm{V}_{\mathrm{CC}}$ via $5.1 \mathrm{k} \Omega$ resistor. |
| 31 | TRST | I/O | Optional programming I/O. Pull to GND via $10 \mathrm{k} \Omega$ resistor. |
| 32 | DGND | - | Digital GND |
| 33 | V33D | - | Digital Core 3.3V Supply |
| 34 | V33A | - | Analog 3.3V Supply |
| 35 | BPCAP | - | 1.8V Bypass Capacitor Connect Pin |
| 36 | AGND | - | Analog GND |
| 37 | AN1 | 1 | Reserved Analog Input. Connect this pin to GND. |
| 38 | AN2 | 1 | Reserved Analog Input. Connect this pin to GND. |
| 39 | AN3 | I | Reserved Analog Input. Connect this pin to GND. |
| 40 | AN4 | 1 | Reserved Analog Input. Connect this pin to GND. |
| 41 | V33FB | 1 | 3.3V Linear-Regulator Feedback Input. Leave this pin floating. |
| 42 | I_COIL | I | Coil Current Input |
| 43 | PMOD_THR | I | Input to Program Parasitic Metal Object Detection Threshold |
| 44 | LED_MODE | 1 | Input to Select LED Mode |
| 45 | AIN7 | 1 | Reserved Analog Input. Connect this pin to GND. |
| 46 | I_IN | 1 | Transmitter Input Current |
| 47 | AGN2 | - | Analog GND 2. |
| 48 | EXTREF | I | External Reference Voltage Input. Connect this Input to GND. |

## TYPICAL CHARACTERISTICS



Figure 1.

RECTIFIER LOADING
vs
OUTPUT POWER


Figure 2.

## FUNCTIONAL OVERVIEW

The typical Wireless Power Transfer System consists of the primary and the secondary coils that are positioned against each other in the way to maximize mutual coupling of their electromagnetic fields. Both coils have ferrite shields as parts of their structure to even further maximize field coupling. The primary coil is exited with the switching waveform of the transmitter power driver that gets its power from AC-DC wall adapter. The secondary coil is connected to the rectifier that can either directly interface the battery or can have an electronic charger or post-regulator connected to its output. The capacitors in series with the coils are tuned to create resonance in the system. The system being in resonance facilitates better energy transfer compared to the inductive transfer. Power transfer in the resonant system can also be easier controlled with the variable frequency control approach.
The rectifier output voltage is monitored by the secondary side microcontroller that generates signals to control modulation circuit to pass coded information from the secondary side to the primary side. The coded information is organized into information packets that have Preamble bytes, Header bytes, message bytes and the Checksum bytes. Per WPC standard information packets can be related to Identification, Configuration, Control Error, Rectified Power, Charge Status and End of Power Transfer. For detailed information on WPC standard visit Wireless Power Consortium web site at http://www.wirelesspowerconsortium.com/.
There are two ways the coupled electromagnetic field can be manipulated to achieve information transfer from the secondary side to the primary side. With resistive modulation approach shown in Figure 3, the communication resistor periodically loads the rectifier output changing system $Q$ factor and as a result value of the voltage on the primary side coil. With capacitive modulation approach shown in Figure 4 a pair of communication capacitors are periodically connected to the receiver coil network. These extra capacitance application changes slightly the resonance frequency of the system and its response on the current operating frequency, which in turn leads to coil voltage variation on the primary side.
With both modulation techniques primary side coil voltage variations are detected with demodulation circuit and further interpreted in bq500110 to restore the content of the information packets and adjust controls to the transmitter schematic.


Figure 3. Resistive Modulation Circuit


Figure 4. Capacitive Modulation Circuit
The bq500110 is the industry's first wireless power dedicated transmitter controller that simplifies integration of wireless power technology in consumer electronics, such as digital cameras, smart phones, MP3 players and global positioning systems, along with infrastructure applications such as furniture and cars.
The bq500110 is a specialized digital power microcontroller that controls WPC A1, single coil, transmitter functions such as analog ping, digital ping, variable frequency output power control, parasitic metal object detection, protection against transmitter coil over-current, over temperature protection of the transmitter top surface, and indication of the transmitter states of operation.
The bq500110 digital inputs receive and interpret signals from the analog demodulator circuit that provides first stage of demodulation for digital commands sent by the WPC compliant Receiver. Based on the received commands the controller provides control signals to the transmitter coil half-bridge power-driver. The controller analog inputs monitor input DC voltage, Input Current, the Coil Current and the Thermal Protection input. These analog inputs support monitoring and protective functions of the controller.

The bq500110 directly controls two LEDs to indicate the controller standby, power transfer, PMOD warning, PMOD-stop and system fault states.

## Option Select Pins

Two pins in bq500110 are allocated to program the LED mode and the PMOD mode of the device. At power-up, a bias current is applied to pins LED_MODE and PMOD_THR and the resulting voltage measured in order to identify the value of the attached programming resistor. The values of the operating parameters set by these pins are determined using Option Select Bins. For LED_MODE the selected bin determines the LED behavior based on LED Modes; for the PMOD_THR the selected bin sets a threshold used for parasitic metal object detection (see Metal Object Detection (MOD) section).


Figure 5. Option Programming

Table 1. Option Select Bins

| BIN NUMBER | RESISTANCE <br> $(\mathbf{k} \Omega \mathbf{)}$ | LED OPTION | PMD <br> THRESHOLD <br> $\mathbf{( m W )}$ |
| :---: | :---: | :---: | :---: |
| 0 | GND | 0 | 1400 |
| 1 | 42.2 | 1 | 1500 |
| 2 | 48.7 | 2 | 1600 |
| 3 | 56.2 | 3 | 1700 |
| 4 | 64.9 | 4 | 1800 |
| 5 | 75.0 | 5 | 1900 |
| 6 | 86.6 | 6 | 2000 |
| 7 | 100 | 7 | 2100 |
| 8 | 115 | 8 | 2200 |
| 10 | 133 | 9 | 2300 |
| 11 | 154 | 10 | 2400 |
| 12 | 205 | 11 | 2500 |
| 13 | open | 12 | 2600 |
| 13 | 13 | OFF |  |

## LED Modes

The bq500110 can directly control up to three LED outputs. They are driven based on one of twelve selectable modes. Using the resistor of the 44 pin to GND select one of the desired LED Indication scheme presented in Table 2.

Table 2. LED Modes


Table 2. LED Modes (continued)


## Thermal Protection

The bq500110 can provide thermal protection to the transmitter. An external NTC resistor can be placed in the most thermally challenged area, which usually is the center of the transmitting coil, and connected between $\mathrm{V}_{\mathrm{CC}}$ and the dedicated pin 3 . The threshold on the pin 3 is set 1.25 V . The NTC resistor and the resistor from the pin 3 to GND create temperature sensitive divider. User has full flexibility choosing the NTC resistor and the value of the resistor from the pin 3 to GND to set the desired temperature when the system shuts down.

$$
\begin{equation*}
\mathrm{R}_{\text {TEMP_IN }}=0.6097 \times \mathrm{R}_{\text {NTC }}\left(\mathrm{T}_{\text {MAX }}\right) \tag{1}
\end{equation*}
$$

The system will attempt to restore normal operation after approximately five minutes being in the suspended mode due to tripping the over-temperature threshold, or if the receiver is removed.

## Audible Notification on Initiation of Power Transfer

The bq500110 is capable of activating two types of buzzers to indicate that power transfer has begun. Pin 15 outputs a high logic signal for 0.5 s which is suitable to activate DC type buzzers with built in tone generation, other types of sound generators, or custom indication systems. Pin 24 outputs a $0.2 \mathrm{~s}, 4000 \mathrm{~Hz}$ square wave signal suitable for inexpensive AC type ceramic buzzers.

## Gate Driver Modes

The inner PID (proportional-integral-derivative) loop feeds the variable frequency driver, which produces a digital signal of $50 \%$ duty cycle with variable frequency. In operation, the inner PID loop calculates the necessary frequency, which is then generated by the variable frequency driver. The variable frequency is then fed into a MOSFET power train that excites the serial resonance transmitter coil.
The bq500110 can operate with several types of MOSFET gate drivers to accommodate various power train topologies. The DRV_SEL input, pin 9, selects between two modes of drive. When pin 9 is pulled to GND, the DRV_EN output, pin 13, will be driven high while the COIL_PWM output sends a square waveform to the gate driver. The most typical and suggested solution is to use a synchronous buck driver like the TPS28225 that drives n -channel upper and lower power MOSFETs with a safe dead-time.
An alternative solution that may utilize a combination of $p$-channel and $n$-channel MOSFETs can be used when input DRV_SEL input, pin 9, is pulled high to $\mathrm{V}_{\mathrm{Cc}}$. In this case the outputs COIL_PWM and DRV_EN, both output the square waveforms to discrete gate drivers. The dead-time is provided by pulse duration difference between the two waveforms.

## Coil Disable Signal

As the part of the WPC 1.0 compliance communication protocol, the bq500110 has the coil damping control signal that is provided on the output COIL_DIS, pin 8 . The damping signal activates the MOSFET that loads the output of the half-bridge with the $100 \Omega$ resistor.

## Power-On Reset

The bq500110 has an integrated power-on reset (POR) circuit that monitors the supply voltage. At power-up, the POR circuit detects the V33D rise. When V33D is greater than VRESET, the device initiates an internal startup sequence. At the end of the startup sequence, the device begins normal operation.

## External Reset

The device can be forced into a reset state by an external circuit connected to the RESET pin. A logic low voltage on this pin holds the device in reset. To avoid an erroneous trigger caused by noise, a $10 \mathrm{k} \Omega$ pull up resistor to 3.3 V is recommended.

## Non-Volatile Memory Error Correction Coding

The device uses Error Correcting Code (ECC) to improve data integrity and provide high reliability storage of Data Flash contents. ECC uses dedicated hardware to generate extra check bits for the user data as it is written into the Flash memory. This adds an additional six bits to each 32 -bit memory word stored into the Flash array. These extra check bits, along with the hardware ECC algorithm, allow for any single bit error to be detected and corrected when the Data Flash is read. Note that the Data Flash configuration has been factory programmed and is not generally available for customization.

## Parasitic Metal Object Detection (PMOD)

As a safety feature, the bq500110 can be configured to detect presence of a parasitic metal object placed in the vicinity of the magnetic field. The BQ500100 uses the Rectified Power Packet information and the measured transmitter input-power to calculate parasitic losses in the system. When an excessive power loss is detected the device will lit the red LED to warn about this undesirable condition, If during a twenty second warning time the parasitic metal object was not removed, the controller will disable power transfer. After being in halt for five minutes bq500110 will attempt normal operation. If the object that caused excessive power dissipation is still present, the sequence will be repeated over and over again. If the metal object is removed during this twenty second warning time, the normal operation will be restored momentarily.
To facilitate parasitic loss function, the bq500110 monitors the input voltage and the input current supplied to the coil power drive circuit.

The PMOD_THR pin is used to set the threshold at which the MOD is activated. The MOD operation can be disabled by selecting the highest bin( leaving the pin is left floating).
The threshold is set by Equation 2:
Threshold $=1400 \mathrm{~mW}+$ Bin_Number $\times 100 \mathrm{~mW}$
Note: The WPC Specification V1.0 does not define the requirements and thresholds for MOD feature, thus metal object detection may perform differently with different products. Therefore make your own decision when setting the threshold. In most desktop wireless charger applications setting the PMOD threshold to 1.5 W shown to give good results in stopping power transfer and preventing small metal objects like coins, pharmaceutical wraps, etc. getting hot when placed in the path of wireless power transfer. Figure 1 depicts PMOD performance measured on bq500110 EVM. .

## APPLICATION INFORMATION

Typical application diagrams for the WPC 1.0 compliant transmitter are shown on the following pages.



Changes from Original (November 2010) to Revision A

- Updated the data sheet for product release ..... 1


## PACKAGING INFORMATION

| Orderable Device | Status <br> (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan <br> (2) | Lead finish/ Ball material <br> (6) | MSL Peak Temp <br> (3) | Op Temp ( ${ }^{\circ} \mathrm{C}$ ) | Device Marking <br> (4/5) | Samples |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BQ500110RGZR | NRND | VQFN | RGZ | 48 | 2500 | RoHS \& Green | NIPDAU | Level-3-260C-168 HR | -40 to 110 | BQ500110 |  |
| BQ500110RGZT | NRND | VQFN | RGZ | 48 | 250 | RoHS \& Green | NIPDAU | Level-3-260C-168 HR | -40 to 110 | BQ500110 |  |

${ }^{(1)}$ The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but Tl does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.
${ }^{(2)}$ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed $0.1 \%$ by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free"
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.
${ }^{(3)}$ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature
${ }^{(4)}$ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
${ }^{(5)}$ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
${ }^{(6)}$ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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## TAPE AND REEL INFORMATION



TAPE DIMENSIONS


| A0 | Dimension designed to accommodate the component width |
| :--- | :--- |
| B0 | Dimension designed to accommodate the component length |
| K0 | Dimension designed to accommodate the component thickness |
| W | Overall width of the carrier tape |
| P1 | Pitch between successive cavity centers |

Reel Width (W1)
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

*All dimensions are nominal

| Device | Package <br> Type | Package <br> Drawing | Pins | SPQ | Reel <br> Diameter <br> $(\mathbf{m m})$ | Reel <br> Width <br> W1 $(\mathbf{m m})$ | A0 <br> $(\mathbf{m m})$ | B0 <br> $(\mathbf{m m})$ | K0 <br> $(\mathbf{m m})$ | P1 <br> $(\mathbf{m m})$ | W <br> $(\mathbf{m m})$ | Pin1 <br> Quadrant |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BQ500110RGZR | VQFN | RGZ | 48 | 2500 | 330.0 | 16.4 | 7.3 | 7.3 | 1.5 | 12.0 | 16.0 | Q2 |
| BQ500110RGZR | VQFN | RGZ | 48 | 2500 | 330.0 | 16.4 | 7.3 | 7.3 | 1.1 | 12.0 | 16.0 | Q2 |
| BQ500110RGZT | VQFN | RGZ | 48 | 250 | 180.0 | 16.4 | 7.3 | 7.3 | 1.5 | 12.0 | 16.0 | Q2 |


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BQ500110RGZR | VQFN | RGZ | 48 | 2500 | 356.0 | 356.0 | 35.0 |
| BQ500110RGZR | VQFN | RGZ | 48 | 2500 | 367.0 | 367.0 | 38.0 |
| BQ500110RGZT | VQFN | RGZ | 48 | 250 | 210.0 | 185.0 | 35.0 |



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.


NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.


NOTES: (continued)
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN <br> VQFN - 1 mm max height 



NOTES: (continued)
6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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