MCV14A
Data Sheet

14-Pin, 8-Bit Flash Microcontroller

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MCV14A

14-Pin, 8-Bit Flash Microcontroller

High-Performance RISC CPU:

- Only 33 Single-Word Instructions
- All Single-Cycle Instructions except for Program Branches which are Two-Cycle
- Two-Level Deep Hardware Stack
- Direct, Indirect and Relative Addressing modes for Data and Instructions
- Operating Speed:
  - DC – 20 MHz crystal oscillator
  - DC – 200 ns instruction cycle
- On-chip Flash Program Memory
  - 1024 x 12
- General Purpose Registers (SRAM)
  - 67 x 8
- Flash Data Memory
  - 64 x 8

Special Microcontroller Features:

- 8 MHz Precision Internal Oscillator
  - Factory calibrated to ±1%
- In-Circuit Serial Programming™ (ICSP™)
- In-Circuit Debugging (ICD) Support
- Power-On Reset (POR)
- Device Reset Timer (DRT)
- Watchdog Timer (WDT) with Dedicated On-Chip RC Oscillator for Reliable Operation
- Programmable Code Protection
- Multiplexed $\overline{\text{MCLR}}$ Input Pin
- Internal Weak Pull-ups on I/O Pins
- Power-Saving Sleep mode
- Wake-Up from Sleep on Pin Change
- Selectable Oscillator Options:
  - INTRC: 4 MHz or 8 MHz precision Internal RC oscillator
  - EXTRC: External low-cost RC oscillator
  - XT: Standard crystal/resonator
  - HS: High-speed crystal/resonator
  - LP: Power-saving, low-frequency crystal
  - EC: High-speed external clock input

Low-Power Features/CMOS Technology:

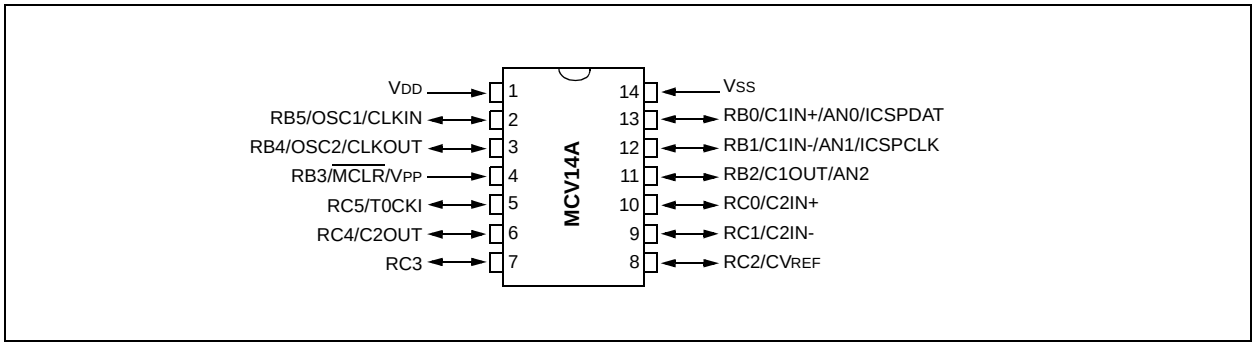
- Standby Current:
  - 100 nA @ 2.0V, typical
- Operating Current:
  - 15 μA @ 32 kHz, 2.0V, typical
  - 170 μA @ 4 MHz, 2.0V, typical
- Watchdog Timer Current:
  - 1 μA @ 2.0V, typical
  - 7 μA @ 5.0V, typical
- High Endurance Program and Flash Data Memory Cells
  - 100,000 write Program Memory endurance
  - 1,000,000 write Flash Data Memory endurance
  - Program and Flash Data retention: >40 years
- Fully Static Design
- Wide Operating Voltage Range: 2.0V to 5.5V
  - Wide temperature range
  - Industrial: -40°C to +85°C

Peripheral Features:

- 12 I/O Pins
  - 11 I/O pins with individual direction control
  - 1 input-only pin
  - High current sink/source for direct LED drive
  - Wake-up on change
  - Weak pull-ups
- 8-Bit Real-Time Clock/Counter (TMR0) with 8-Bit Programmable Prescaler
- Two Analog Comparators
  - Comparator inputs and output accessible externally
  - One comparator with 0.6V fixed on-chip absolute voltage reference (V_{REF})
  - One comparator with programmable on-chip voltage reference (V_{REF})
- Analog-to-Digital (A/D) Converter
  - 8-bit resolution
  - 3-channel external programmable inputs
  - 1-channel internal input to internal absolute 0.6 voltage reference

Device	Program Memory	Data Memory		I/O	Comparators	Timers 8-bit	8-bit A/D Channels
	Flash (words)	SRAM (bytes)	Flash (bytes)				
MCV14A	1024	67	64	12	2	1	3

FIGURE 1: 14-PIN PDIP AND SOIC DIAGRAM



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1.0 GENERAL DESCRIPTION

The MCV14A device from Microchip Technology is low-cost, high-performance, 8-bit, fully-static, Flash-based CMOS microcontrollers. It employs a RISC architecture with only 33 single-word/single-cycle instructions. All instructions are single cycle (200 μs) except for program branches, which take two cycles. The MCV14A device delivers performance an order of magnitude higher than their competitors in the same price category. The 12-bit wide instructions are highly symmetrical, resulting in a typical 2:1 code compression over other 8-bit microcontrollers in its class. The easy-to-use and easy to remember instruction set reduces development time significantly.

The MCV14A product is equipped with special features that reduce system cost and power requirements. The Power-on Reset (POR) and Device Reset Timer (DRT) eliminate the need for external Reset circuitry. There are four oscillator configurations to choose from, including INTRC Internal Oscillator mode and the power-saving LP (Low-Power) Oscillator mode. Power-Saving Sleep mode, Watchdog Timer and code protection features improve system cost, power and reliability.

The MCV14A device is available in the cost-effective Flash programmable version, which is suitable for production in any volume. The customer can take full advantage of Microchip's price leadership in Flash programmable microcontrollers, while benefiting from the Flash programmable flexibility.

The MCV14A product is supported by a full-featured macro assembler, a software simulator, an in-circuit emulator, a 'C' compiler, a low-cost development programmer and a full featured programmer. All the tools are supported on PC and compatible machines.

1.1 Applications

The MCV14A device fits in applications ranging from personal care appliances and security systems to low-power remote transmitters/receivers. The Flash technology makes customizing application programs (transmitter codes, appliance settings, receiver frequencies, etc.) extremely fast and convenient. The small footprint packages, for through hole or surface mounting, make these microcontrollers perfect for applications with space limitations. Low cost, low power, high performance, ease of use and I/O flexibility make the MCV14A device very versatile even in areas where no microcontroller use has been considered before (e.g., timer functions, logic and PLDs in larger systems and coprocessor applications).

TABLE 1-1: FEATURES AND MEMORY OF MCV14A

		MCV14A
Clock	Maximum Frequency of Operation (MHz)	20
Memory	Flash Program Memory	1024
	SRAM Data Memory (bytes)	67
	Flash Data Memory	64
Peripherals	Timer Module(s)	TMR0
	Wake-up from Sleep on Pin Change	Yes
Features	I/O Pins	11
	Input Pins	1
	Internal Pull-ups	Yes
	In-Circuit Serial Programming™	Yes
	Number of Instructions	33
	Packages	14-pin PDIP and SOIC

The MCV14A device has Power-on Reset, selectable Watchdog Timer, selectable code-protect, high I/O current capability and precision internal oscillator.
The MCV14A device uses serial programming with data pin RB0 and clock pin RB1.

NOTES:

2.0 ARCHITECTURAL OVERVIEW

The high performance of the MCV14A device can be attributed to a number of architectural features commonly found in RISC microprocessors. To begin with, the MCV14A device uses a Harvard architecture in which program and data are accessed on separate buses. This improves bandwidth over traditional von Neumann architectures where program and data are fetched on the same bus. Separating program and data memory further allows instructions to be sized differently than the 8-bit wide data word. Instruction opcodes are 12 bits wide, making it possible to have all single-word instructions. A 12-bit wide program memory access bus fetches a 12-bit instruction in a single cycle. A two-stage pipeline overlaps fetch and execution of instructions. Consequently, all instructions (33) execute in a single cycle (200 ns @ 20 MHz, 1 μs @ 4 MHz) except for program branches.

Table 2-1 below lists memory supported by the MCV14A device.

TABLE 2-1: MCV14A MEMORY

Device	Program Memory	Data Memory	
	Flash (words)	SRAM (bytes)	Flash (bytes)
MCV14A	1024	67	64

The MCV14A device can directly or indirectly address its register files and data memory. All Special Function Registers (SFR), including the PC, are mapped in the data memory. The MCV14A device has a highly orthogonal (symmetrical) instruction set that makes it possible to carry out any operation, on any register, using any Addressing mode. This symmetrical nature and lack of “special optimal situations” make programming with the MCV14A device simple, yet efficient. In addition, the learning curve is reduced significantly.

The MCV14A device contains an 8-bit ALU and working register. The ALU is a general purpose arithmetic unit. It performs arithmetic and Boolean functions between data in the working register and any register file.

The ALU is 8 bits wide and capable of addition, subtraction, shift and logical operations. Unless otherwise mentioned, arithmetic operations are two’s complement in nature. In two-operand instructions, one operand is typically the W (working) register. The other operand is either a file register or an immediate constant. In single operand instructions, the operand is either the W register or a file register.

The W register is an 8-bit working register used for ALU operations. It is not an addressable register.

Depending on the instruction executed, the ALU may affect the values of the Carry (C), Digit Carry (DC) and Zero (Z) bits in the STATUS register. The C and DC bits operate as a borrow and digit borrow out bit, respectively, in subtraction. See the SUBWF and ADDWF instructions for examples.

A simplified block diagram is shown in Figure 2-2, with the corresponding device pins described in Table 2-2.

FIGURE 2-1: MCV14A BLOCK DIAGRAM

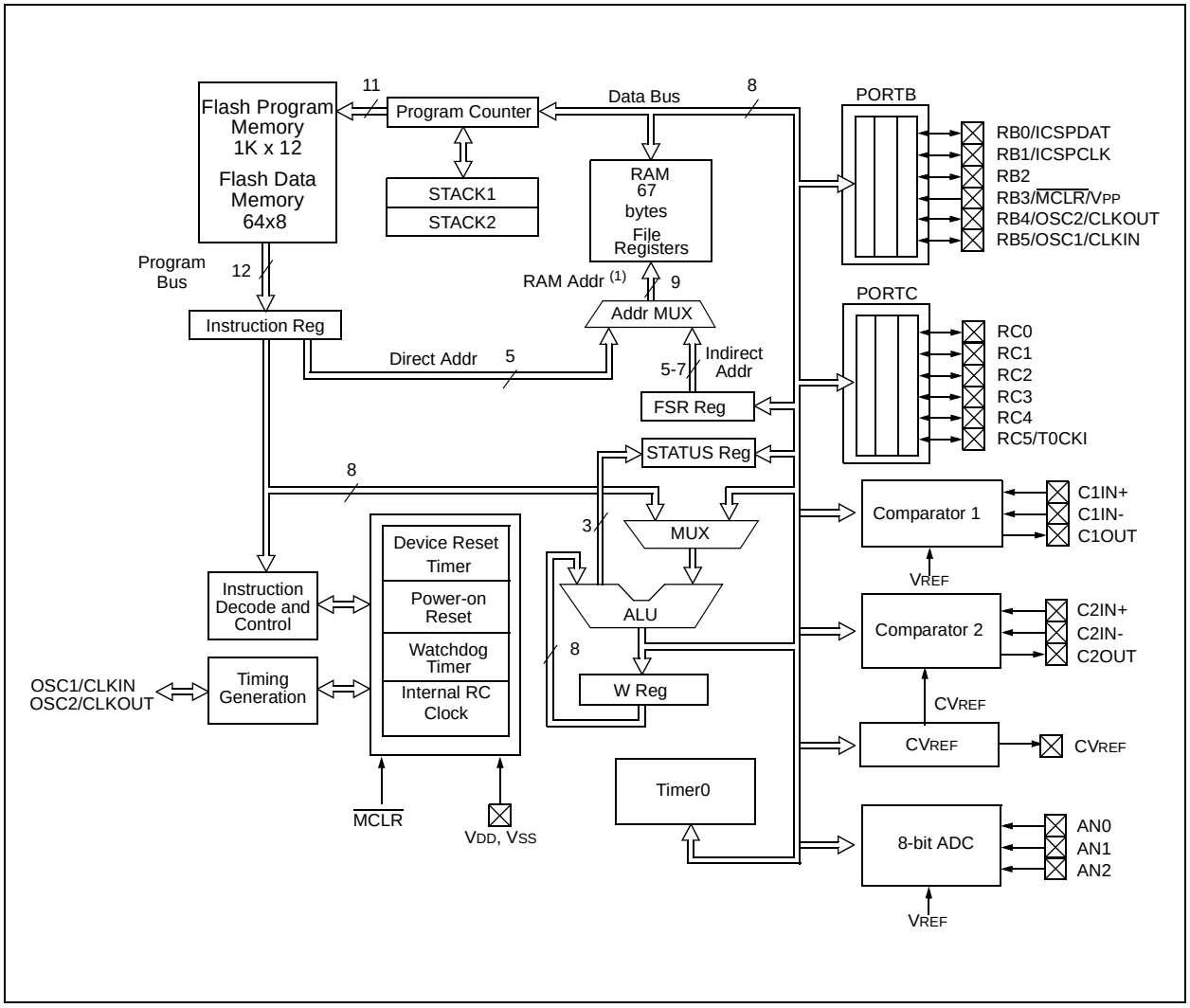


TABLE 2-2: MCV14A PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RB0//C1IN+/AN0/ ICSPDAT	RB0	TTL	CMOS	Bidirectional I/O pin. Can be software programmed for internal weak pull-up and wake-up from Sleep on pin change.
	C1IN+	AN	—	Comparator 1 input.
	AN0	AN	—	ADC channel input.
	ICSPDAT	ST	CMOS	ICSP™ mode Schmitt Trigger.
RB1/C1IN-/AN1/ ICSPCLK	RB1	TTL	CMOS	Bidirectional I/O pin. Can be software programmed for internal weak pull-up and wake-up from Sleep on pin change.
	C1IN-	AN	—	Comparator 1 input.
	AN1	AN	—	ADC channel input.
	ICSPCLK	ST	CMOS	ICSP mode Schmitt Trigger.
RB2/C1OUT/AN2	RB2	TTL	CMOS	Bidirectional I/O pin.
	C1OUT	—	CMOS	Comparator 1 output.
	AN2	AN	—	ADC channel input.
RB3/$\overline{\text{MCLR}}$/V_{PP}	RB3	TTL	—	Input pin. Can be software programmed for internal weak pull-up and wake-up from Sleep on pin change.
	$\overline{\text{MCLR}}$	ST	—	Master Clear (Reset). When configured as $\overline{\text{MCLR}}$, this pin is an active-low Reset to the device. Voltage on $\overline{\text{MCLR}}$/V_{PP} must not exceed V_{DD} during normal device operation or the device will enter Programming mode. Weak pull-up always on if configured as $\overline{\text{MCLR}}$.
	V_{PP}	HV	—	Programming voltage input.
RB4/OSC2/CLKOUT	RB4	TTL	CMOS	Bidirectional I/O pin.
	OSC2	—	XTAL	Oscillator crystal output. Connections to crystal or resonator in Crystal Oscillator mode (XT, HS and LP modes only, PORTB in other modes).
	CLKOUT	—	CMOS	EXTRC/INTRC CLKOUT pin (F_{OSC}/4).
RB5/OSC1/CLKIN	RB5	TTL	CMOS	Bidirectional I/O pin.
	OSC1	XTAL	—	Oscillator crystal input.
	CLKIN	ST	—	External clock source input.
RC0/C2IN+	RC0	TTL	CMOS	Bidirectional I/O port.
	C2IN+	AN	—	Comparator 2 input.
RC1/C2IN-	RC1	TTL	CMOS	Bidirectional I/O port.
	C2IN-	AN	—	Comparator 2 input.
RC2/CV_{REF}	RC2	TTL	CMOS	Bidirectional I/O port.
	CV_{REF}	—	AN	Programmable Voltage Reference output.
RC3	RC3	TTL	CMOS	Bidirectional I/O port.
RC4/C2OUT	RC4	TTL	CMOS	Bidirectional I/O port.
	C2OUT	—	CMOS	Comparator 2 output.
RC5/T0CKI	RC5	TTL	CMOS	Bidirectional I/O port.
	T0CKI	ST	—	Timer0 Schmitt Trigger input pin.
V_{DD}	V_{DD}	—	P	Positive supply for logic and I/O pins.
V_{SS}	V_{SS}	—	P	Ground reference for logic and I/O pins.

Legend: I = Input, O = Output, I/O = Input/Output, P = Power, — = Not used, TTL = TTL input, ST = Schmitt Trigger input, HV = High Voltage

2.1 Clocking Scheme/Instruction Cycle

The clock input (OSC1/CLKIN pin) is internally divided by four to generate four non-overlapping quadrature clocks, namely Q1, Q2, Q3 and Q4. Internally, the PC is incremented every Q1 and the instruction is fetched from program memory and latched into the instruction register in Q4. It is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow is shown in Figure 2-2 and Example 2-1.

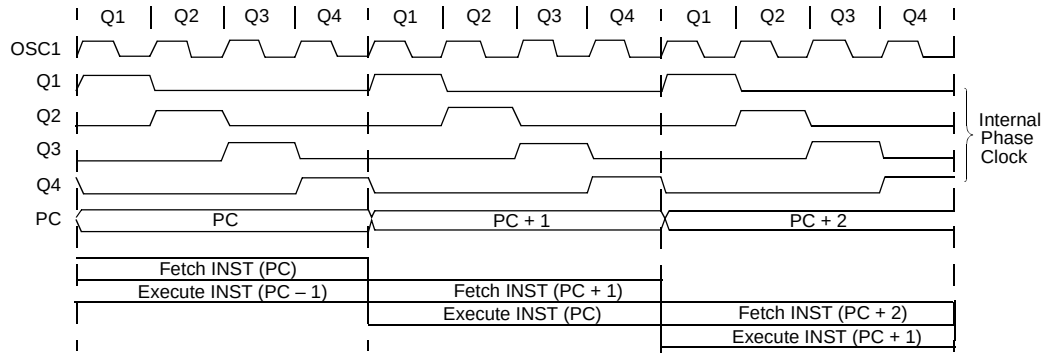
2.2 Instruction Flow/Pipelining

An instruction cycle consists of four Q cycles (Q1, Q2, Q3 and Q4). The instruction fetch and execute are pipelined such that fetch takes one instruction cycle, while decode and execute take another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the PC to change (e.g., GOTO), then two cycles are required to complete the instruction (Example 2-1).

A fetch cycle begins with the PC incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the Instruction Register (IR) in cycle Q1. This instruction is then decoded and executed during the Q2, Q3 and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

FIGURE 2-2: CLOCK/INSTRUCTION CYCLE



EXAMPLE 2-1: INSTRUCTION PIPELINE FLOW

```
1. MOVLW 03H
2. MOVWF PORTB
3. CALL  SUB_1
4. BSF   PORTB, BIT1
```

Fetch 1	Execute 1				
	Fetch 2	Execute 2			
		Fetch 3	Execute 3		
			Fetch 4	Flush	
				Fetch SUB_1	Execute SUB_1

All instructions are single cycle, except for any program branches. These take two cycles, since the fetch instruction is “flushed” from the pipeline, while the new instruction is being fetched and then executed.

3.0 MEMORY ORGANIZATION

The MCV14A memories are organized into program memory and data memory (SRAM).The self-writable portion of the program memory called Flash data memory is located at addresses at 400h-43Fh. All Program mode commands that work on the normal Flash memory work on the Flash data memory. This includes bulk erase, row/column/cycling toggles, Load and Read data commands (Refer to **Section 4.0 "Flash Data Memory"** for more details). For devices with more than 512 bytes of program memory, a paging scheme is used. Program memory pages are accessed using one STATUS register bit. For the MCV14A, with data memory register files of more than 32 registers, a banking scheme is used. Data memory banks are accessed using the File Select Register (FSR).

3.1 Program Memory Organization for the MCV14A

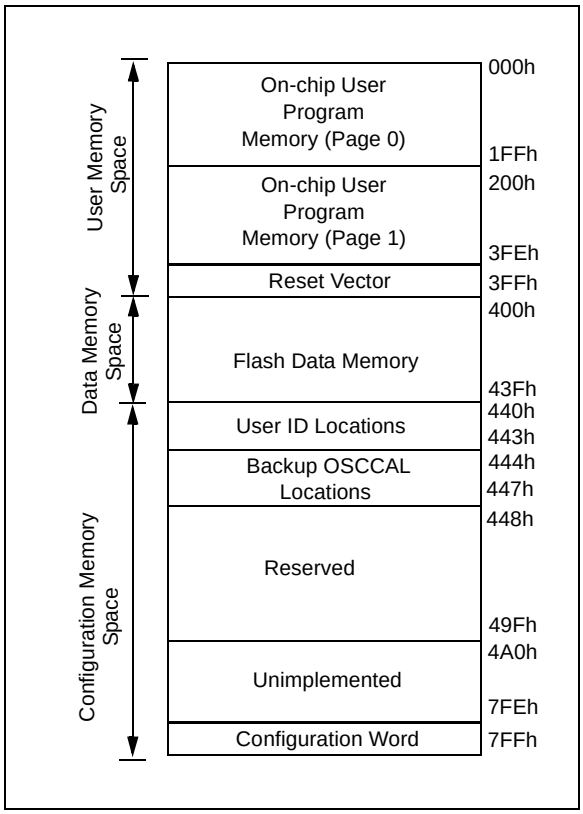
The MCV14A device has an 11-bit Program Counter (PC) capable of addressing a 2K x 12 program memory space. Program memory is partitioned into user memory, data memory and configuration memory spaces.

The user memory space is the on-chip user program memory. As shown in Figure 3-1, it extends from 0x000 to 0x3FF and partitions into pages, including Reset vector at address 0x3FF.

The data memory space is the Flash data memory block and is located at addresses PC = 400h-43Fh. All Program mode commands that work on the normal Flash memory work on the Flash data memory block. This includes bulk erase, Load and Read data commands.

The Configuration Memory Space extends from 0x440 to 0x7FF. Locations from 0x448 through 0x49F are reserved. The User I.D. locations extend from 0x440 through 0x443. The Backup OSCCAL locations extend from 0x444 through 0x447. The Configuration Word is physically located at 0x7FF.

FIGURE 3-1: MEMORY MAP



3.2 Data Memory (SRAM and FSRs)

Data memory is composed of registers or bytes of SRAM. Therefore, data memory for a device is specified by its register file. The register file is divided into two functional groups: Special Function Registers (SFR) and General Purpose Registers (GPR).

The Special Function Registers are registers used by the CPU and peripheral functions for controlling desired operations of the MCV14A. See Figure 3-2 for details.

The MCV14A register file is composed of 13 Special Function Registers and 41 General Purpose Registers

3.2.1 GENERAL PURPOSE REGISTER FILE

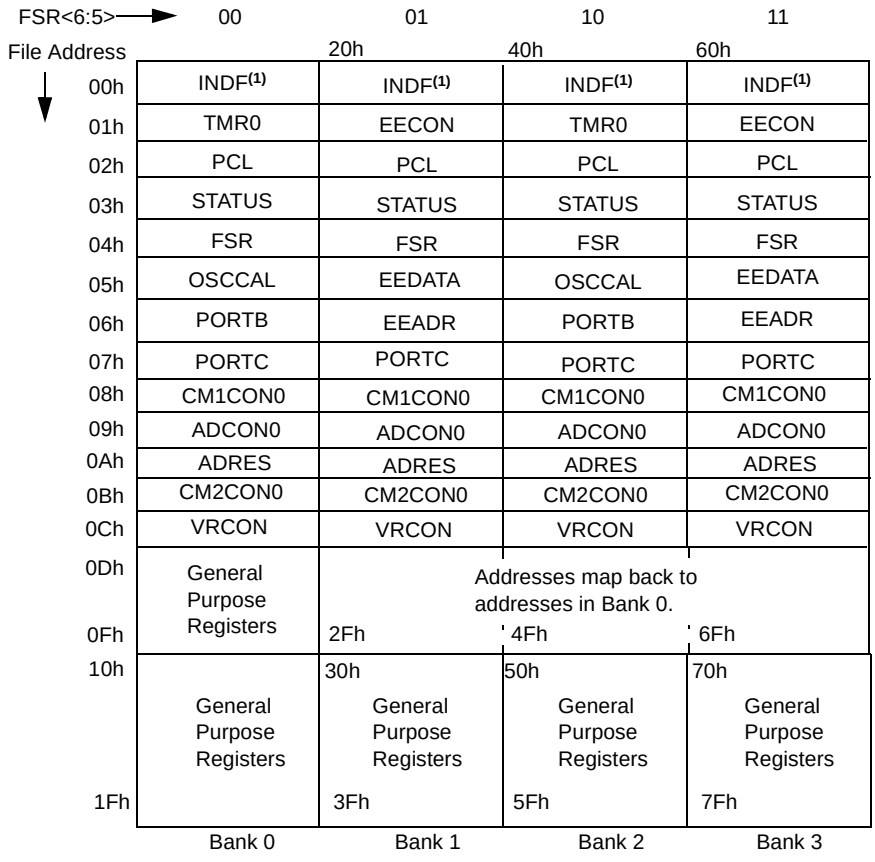
The General Purpose Register file is accessed, either directly or indirectly, through the File Select Register (FSR). See **Section 3.8 "Indirect Data Addressing: INDF and FSR Registers"**.

3.2.2 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral functions to control the operation of the device (Table 3-1).

The Special Function Registers can be classified into two sets. The Special Function Registers associated with the "core" functions are described in this section. Those related to the operation of the peripheral features are described in the section for each peripheral feature.

FIGURE 3-2: REGISTER FILE MAP



Note 1: Not a physical register. See **Section 3.8 "Indirect Data Addressing: INDF and FSR Registers"**.

3.2.3 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral functions to control the operation of the device (Table 3-1).

The Special Function Registers can be classified into two sets. The Special Function Registers associated with the “core” functions are described in this section. Those related to the operation of the peripheral features are described in the section for each peripheral feature.

TABLE 3-1: SPECIAL FUNCTION REGISTER (SFR) SUMMARY

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Page #
N/A	TRIS	—	—	I/O Control Register (PORTB, PORTC)						--11 1111	25
N/A	OPTION	Contains control bits to configure Timer0 and Timer0/WDT prescaler								1111 1111	17
00h	INDF	Uses contents of FSR to Address Data Memory (not a physical register)								xxxx xxxx	20
01h/41h	TMR0	Timer0 Module Register								xxxx xxxx	29
02h(1)	PCL	Low order 8 bits of PC								1111 1111	19
03h	STATUS	RBWUF	CWUF	PA0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	16
04h	FSR	Indirect Data Memory Address Pointer								100x xxxx	20
05h/45h	OSCCAL	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	—	1111 111-	18
06h/46h	PORTB	—	—	RB5	RB4	RB3	RB2	RB1	RB0	--xx xxxx	25
07h	PORTC	—	—	RC5	RC4	RC3	RC2	RC1	RC0	--xx xxxx	26
08h	CM1CON0	C1OUT	$\overline{C1OUTEN}$	C1POL	$\overline{C1T0CS}$	C1ON	C1NREF	C1PREF	$\overline{C1WU}$	1111 1111	53
09h	ADCON0	ANS1	ANS0	ADCS1	ADCS0	CHS1	CHS0	GO/$\overline{DONE}$	ADON	1111 1100	51
0Ah	ADRES	ADC Conversion Result								xxxx xxxx	52
0Bh	CM2CON0	C2OUT	$\overline{C2OUTEN}$	C2POL	C2PREF2	C2ON	C2NREF	C2PREF1	$\overline{C2WU}$	1111 1111	54
0Ch	VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	000- 0000	59
21h/61h	EECON	—	—	—	FREE	WRERR	WREN	WR	RD	---0 0000	22
25h/65h	EEDATA	SELF READ/WRITE DATA								xxxx xxxx	21
26h/66h	EEADR	—	—	SELF READ/WRITE ADDRESS						--xx xxxx	21

Legend: x = unknown, u = unchanged, – = unimplemented, read as '0' (if applicable). Shaded cells = unimplemented or unused

Note 1: The upper byte of the Program Counter is not directly accessible. See **Section 3.6 “Program Counter”** for an explanation of how to access these bits.

3.3 STATUS Register

This register contains the arithmetic status of the ALU, the Reset status and the page preselect bit.

The STATUS register can be the destination for any instruction, as with any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the $\overline{TO}$ and $\overline{PD}$ bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, `CLRF STATUS,` will clear the upper three bits and set the Z bit. This leaves the STATUS register as `000u u1uu` (where u = unchanged).

Therefore, it is recommended that only `BCF`, `BSF` and `MOVWF` instructions be used to alter the STATUS register. These instructions do not affect the Z, DC or C bits from the STATUS register.

REGISTER 3-1: STATUS: STATUS REGISTER

R/W-0	R/W-0	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
RBWUF	CWUF	PA0	$\overline{TO}$	$\overline{PD}$	Z	DC	C
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **RBWUF**: Wake-up from Sleep on Pin Change bit
`1` = Reset due to wake-up from Sleep on pin change
`0` = After power-up or other Reset

bit 6 **CWUF:** Wake-up from Sleep on Comparator Change bit
`1` = Reset due to wake-up from Sleep on comparator change
`0` = After power-up or other Reset

bit 5 **PA0**: Program Page Preselect bit
`1` = Page 1 (000h-1FFh)
`0` = Page 0 (200h-3FFh)

bit 4 **$\overline{TO}$**: Time-out bit
`1` = After power-up, `CLRWDT` instruction, or `SLEEP` instruction
`0` = A WDT time-out occurred

bit 3 **$\overline{PD}$**: Power-down bit
`1` = After power-up or by the `CLRWDT` instruction
`0` = By execution of the `SLEEP` instruction

bit 2 **Z**: Zero bit
`1` = The result of an arithmetic or logic operation is zero
`0` = The result of an arithmetic or logic operation is not zero

bit 1 **DC**: Digit carry/$\overline{\text{borrow}}$ bit (for `ADDWF` and `SUBWF` instructions)
`ADDWF:`
`1` = A carry from the 4th low-order bit of the result occurred
`0` = A carry from the 4th low-order bit of the result did not occur
`SUBWF:`
`1` = A borrow from the 4th low-order bit of the result did not occur
`0` = A borrow from the 4th low-order bit of the result occurred

bit 0 **C**: Carry/$\overline{\text{borrow}}$ bit (for `ADDWF`, `SUBWF` and `RRF`, `RLF` instructions)

`ADDWF:`	`SUBWF:`	`RRF` or `RLF:`
`1` = A carry occurred	`1` = A borrow did not occur	Load bit with LSb or MSb, respectively
`0` = A carry did not occur	`0` = A borrow occurred	

3.4 OPTION Register

The OPTION register is a 8-bit wide, write-only register, which contains various control bits to configure the Timer0/WDT prescaler and Timer0.

By executing the OPTION instruction, the contents of the W register will be transferred to the OPTION register. A Reset sets the OPTION<7:0> bits.

Note:	If TRIS bit is set to '0', the wake-up on change and pull-up functions are disabled for that pin (i.e., note that TRIS overrides Option control of $\overline{\text{RBPU}}$ and $\overline{\text{RBWU}}$).

REGISTER 3-2: OPTION: OPTION REGISTER

W-1	W-1	W-1	W-1	W-1	W-1	W-1	W-1
$\overline{\text{RBWU}}$	$\overline{\text{RBPU}}$	T0CS(1)	T0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **$\overline{\text{RBWU}}$:** Enable Wake-up On Pin Change bit
1 = Disabled
0 = Enabled

bit 6 **$\overline{\text{RBPU}}$:** Enable Weak Pull-ups bit
1 = Disabled
0 = Enabled

bit 5 **T0CS:** Timer0 Clock Source Select bit(1)
1 = Transition on T0CKI pin
0 = Internal instruction cycle clock (CLKOUT)

bit 4 **T0SE:** Timer0 Source Edge Select bit
1 = Increment on high-to-low transition on T0CKI pin
0 = Increment on low-to-high transition on T0CKI pin

bit 3 **PSA:** Prescaler Assignment bit
1 = Prescaler assigned to the WDT
0 = Prescaler assigned to Timer0

bit 2-0 **PS<2:0>:** Prescaler Rate Select bits

Bit Value	Timer0 Rate	WDT Rate
000	1 : 2	1 : 1
001	1 : 4	1 : 2
010	1 : 8	1 : 4
011	1 : 16	1 : 8
100	1 : 32	1 : 16
101	1 : 64	1 : 32
110	1 : 128	1 : 64
111	1 : 256	1 : 128

Note 1: If the T0CS bit is set to '1', it will override the TRIS function on the T0CKI pin.

3.5 OSCCAL Register

The Oscillator Calibration (OSCCAL) register is used to calibrate the 8 MHz internal oscillator macro. It contains 7 bits of calibration that uses a two's complement scheme for controlling the oscillator speed. See Register 3-3 for details.

REGISTER 3-3: OSCCAL: OSCILLATOR CALIBRATION REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	U-0
CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	—
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-1 **CAL<6:0>**: Oscillator Calibration bits

0111111 = Maximum frequency
•
•
•
0000001
0000000 = Center frequency
1111111
•
•
•
1000000 = Minimum frequency

bit 0 **Unimplemented**: Read as '0'

3.6 Program Counter

As a program instruction is executed, the Program Counter (PC) will contain the address of the next program instruction to be executed. The PC value is increased by one every instruction cycle, unless an instruction changes the PC.

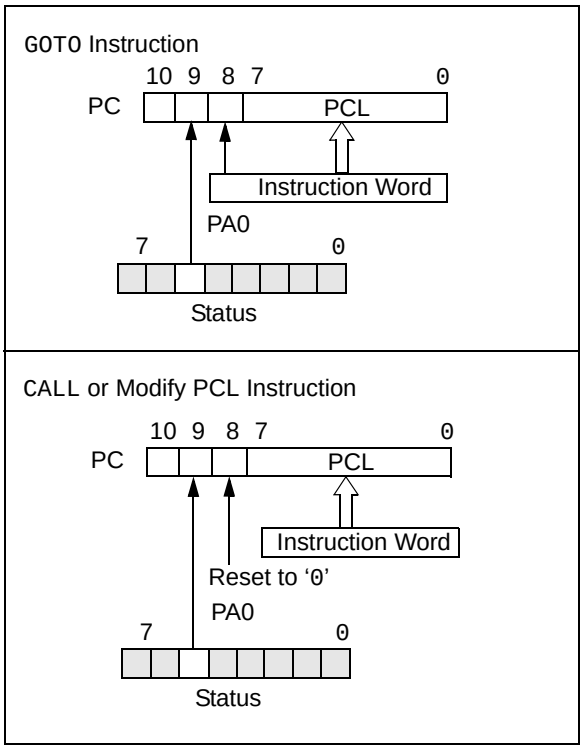
For a GOTO instruction, bits 8:0 of the PC are provided by the GOTO instruction word. The Program Counter (PCL) is mapped to PC<7:0>. Bit 5 of the STATUS register provides page information to bit 9 of the PC (Figure 3-3).

For a CALL instruction, or any instruction where the PCL is the destination, bits 7:0 of the PC again are provided by the instruction word. However, PC<8> does not come from the instruction word, but is always cleared (Figure 3-3).

Instructions where the PCL is the destination, or modify PCL instructions, include MOVWF PC, ADDWF PC and BSF PC,5.

Note:	Because PC<8> is cleared in the CALL instruction or any modify PCL instruction, all subroutine calls or computed jumps are limited to the first 256 locations of any program memory page (512 words long).

FIGURE 3-3: LOADING OF PC BRANCH INSTRUCTIONS



3.6.1 EFFECTS OF RESET

The PC is set upon a Reset, which means that the PC addresses the last location in the last page (i.e., the oscillator calibration instruction). After executing MOVLW XX, the PC will roll over to location 00h and begin executing user code.

The STATUS register page preselect bits are cleared upon a Reset, which means that page 0 is pre-selected.

Therefore, upon a Reset, a GOTO instruction will automatically cause the program to jump to page 0 until the value of the page bits is altered.

3.7 Stack

The MCV14A device has a 2-deep, 12-bit wide hardware PUSH/POP stack.

A CALL instruction will PUSH the current value of Stack 1 into Stack 2 and then PUSH the current PC value, incremented by one, into Stack Level 1. If more than two sequential CALLs are executed, only the most recent two return addresses are stored.

A RETLW instruction will POP the contents of Stack Level 1 into the PC and then copy Stack Level 2 contents into Stack Level 1. If more than two sequential RETLWs are executed, the stack will be filled with the address previously stored in Stack Level 2. Note that the W register will be loaded with the literal value specified in the instruction. This is particularly useful for the implementation of data look-up tables within the program memory.

Note 1:	There are no Status bits to indicate Stack Overflows or Stack Underflow conditions.
2:	There are no instruction mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL and RETLW instructions.

3.8 Indirect Data Addressing: INDF and FSR Registers

The INDF Register is not a physical register. Addressing INDF actually addresses the register whose address is contained in the FSR Register (FSR is a *pointer*). This is indirect addressing.

Reading INDF itself indirectly (FSR = 0) will produce 00h. Writing to the INDF Register indirectly results in a no-operation (although Status bits may be affected).

The FSR is 8-bit wide register. It is used in conjunction with the INDF Register to indirectly address the data memory area.

The FSR<4:0> bits are used to select data memory addresses 00h to 1Fh.

FSR<6:5> are the bank select bits and are used to select the bank to be addressed (00 = Bank 0, 01 = Bank 1, 10 = Bank 2, 11 = Bank 3).

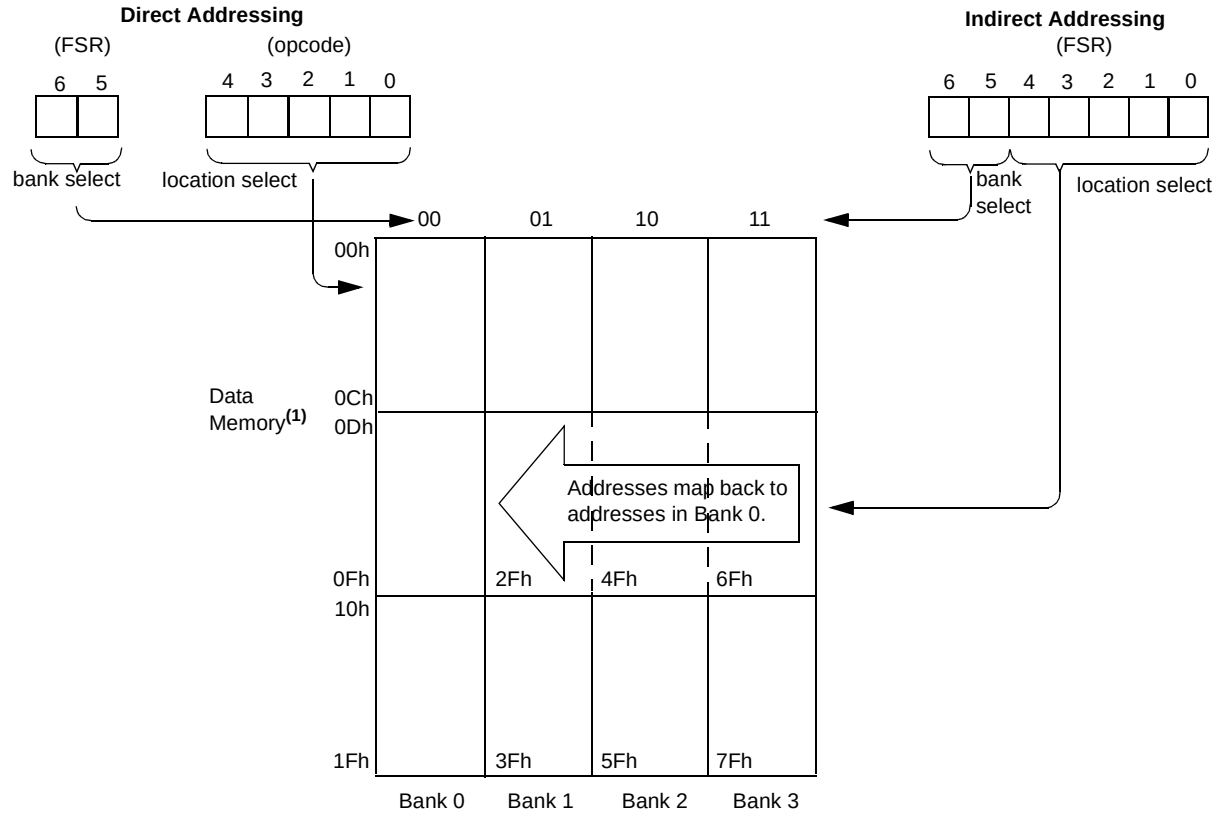
FSR<7> is unimplemented and read as '1'.

A simple program to clear RAM locations 10h-1Fh using indirect addressing is shown in Example 3-1.

EXAMPLE 3-1: HOW TO CLEAR RAM USING INDIRECT ADDRESSING

```
        MOVLW   0x10    ;initialize pointer
        MOVWF   FSR     ;to RAM
NEXT    CLRF    INDF    ;clear INDF
                        ;register
        INCF    FSR,F   ;inc pointer
        BTFSC   FSR,4   ;all done?
        GOTO    NEXT    ;NO, clear next
CONTINUE
            :           ;YES, continue
            :
```

FIGURE 3-4: DIRECT/INDIRECT ADDRESSING



Note 1: For register map detail see Figure 3-2.

4.0 FLASH DATA MEMORY

The data memory is the Flash data memory block, which attaches to the user Flash program memory. It is located at addresses 0x400-0x43F, as shown in Figure 5-1.

This Flash data memory block consists of 8 rows and has self-write capability of up to 64 bytes. This memory block is not directly mapped in the register file space. Instead, it is indirectly addressed through the Special Function Registers. There are three SFRs used to read and write this memory:

- EEDATA (Register 4-1)
- EEADR (Register 4-2)
- EECON (Register 4-3)

EEDATA holds the 8-bit data for read/write, and EEADR holds the address of the EEDATA location being accessed. The effective program counter is EEADR + 400h with only the lower 8 bits of each word being readable or writable.

- EEADR = 00h, PC = 400h
- EEADR = 01h, PC = 401h

The Flash data memory allows byte read and write, and during the operations of read and write cycles, the CPU stalls.

The timing for all self-writes and erases is controlled by the internal timing block of the program memory (see **Section 11.0 "Electrical Characteristics"**, Table 11-11). The write/erase voltages are generated by an on-chip charge pump rated to operate over the voltage range of the device for byte or word operations.

When the device is code-protected, the CPU may continue to read and write the Flash data memory and read the program memory. When code-protected, the device programmer can no longer access data or program memory.

REGISTER 4-1: EEDATA: FLASH DATA REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
EEDATA7	EEDATA6	EEDATA5	EEDATA4	EEDATA3	EEDATA2	EEDATA1	EEDATA0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 **EEDATA<7:0>**: 8-bits of data to be read from/written to data Flash

REGISTER 4-2: EEADR: FLASH ADDRESS REGISTER

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	EEADR5	EEADR4	EEADR3	EEADR2	EEADR1	EEADR0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Do not use

bit 5-0 **EEADR<5:0>**: 6-bits of data to be read from/written to data Flash

REGISTER 4-3: EECON: FLASH CONTROL REGISTER

U-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	—	FREE	WRERR	WREN	WR	RD
bit 7							bit 0

Legend:			
S = Bit can only be set			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-5 **Unimplemented:** Do not use

bit 4 **FREE:** Flash Data Memory Row Erase Enable Bit
- 1 = Program memory row being pointed to by EEADR will be erased on the next write cycle. No write will be performed. This bit is cleared at the completion of the erase operation.
- 0 = Perform write only

bit 3 **WRERR:** Write Error Flag bit
- 1 = A write operation terminated prematurely (by device Reset)
- 0 = Write operation completed successfully

bit 2 **WREN:** Write Enable bit
- 1 = Allows write cycle to Flash data memory
- 0 = Inhibits write cycle to Flash data memory

bit 1 **WR:** Write Control bit
- 1 = Initiate a erase or write cycle
- 0 = Write/Erase cycle is complete

bit 0 **RD:** Read Control bit
- 1 = Initiate a read of Flash data memory
- 0 = Do not read Flash data memory

4.1 Reading Data Memory

To read a memory location, the user must write the address to be read into the EEADR register and then set the RD bit in the EECON register. The data will be available in the next instruction cycle.

EXAMPLE 4-1: FLASH DATA MEMORY READ

```
 BSF       FSR,5       ;SWITCH TO BANK 1
 MOVLW     EE_ADR_READ ;LOAD ADDRESS TO READ
 BSF       EECON,RD    ;INITITATE THE READ
INSTRUCTION
                       ;IS DECODED
 MOVF      EEDATA,W    ;GET NEW DATA
```

Note: Only a BSF command will work to enable the Flash data memory read documented in Example 4-1. No other sequence of commands will work, no exceptions.

4.2 Erasing a Data Memory Row

In order to write new data to the Flash data memory, the program memory row that is being addressed by EEADR<5:0> must be erased.

To prevent a spurious row erasure, a specific sequence must be executed to initiate the erase to the program memory. The sequence is as follows:

- Set the FREE bit (enable Flash data memory row erase)
- Set the WREN bit (enable writes to the Flash data memory array)
- Set the WR bit (initiates the row erase of the Flash data memory array)

If the WREN bit is not set in the instruction cycle after the FREE bit is set, the FREE bit will be cleared in hardware.

If the WR bit is not set in the instruction cycle after the WREN bit is set, the WREN bit will be cleared in hardware.

Both of these sequences is to prevent an accidental erase of the Flash data memory.

EXAMPLE 4-2: ERASE DATA MEMORY ROW

```
BSF    FSR,5         ;SWITCH TO BANK 1
MOVLW  EE_ADR_ERASE  ;LOAD ADDRESS TO ERASE
MOVWF  EEADR         ;LOAD ADDRESS TO SFR
BSF    EECON,FREE    ;SELECT ERASE
BSF    EECON,WREN    ;ENABL FLASH PROG'ING
BSF    EECON,WR      ;INITITATE ERASE

XXX                  ;NEXT INSTRUCTION
```

Note: The FREE bit may be set by any command normally used by the core. However, the WREN and WR bits can only be set using a series of BSF commands, as documented in Example 4-2. No other sequence of commands will work, no exceptions.

4.3 Writing a Data Memory Word

To write a memory location, the user must write the address to be written to into the EEADR register. He must then load the data to be written into the EEDATA register. Once the data and address have been loaded, a specific sequence must be executed to initiate the write to the program memory. The sequence is as follows:

- Set the WREN bit (enable writes to the Flash data memory array)
- Set the WR bit (initiates the write to the Flash data memory array)

If the WR bit is not set in the instruction cycle after the WREN bit is set, the WREN bit will be cleared in hardware.

This sequence is to prevent an accidental write to the Flash memory.

EXAMPLE 4-3: DATA MEMORY WRITE

```
    BSF    FSR,5            ;SWITCH TO BANK 1
    MOVLW  EE_ADR_WRITE     ;LOAD ADDRESS TO
                            ;WRITE
    MOVWF  EEADR            ;INTO EEADR
                            ;REGISTER
    MOVLW  EE_DATA_TO_WRITE;LOAD DATA TO
    MOVWF  EEDATA           ;INTO EEDATA
                            ;REGISTER
    BSF    EECON,WREN       ;ENABLE WRITES
    BSF    EECON,WR         ;START WRITE
                            ;SEQUENCE
    NOP                     ;WAIT AS READ
                            ;INSTRUCTION
                            ;IS DECODED
    NOP                     ;INSTRUCTION
IGNORED
```

Note 1: Only a series of BSF commands will work to enable the memory write sequence documented in Example 4-3. No other sequence of commands will work, no exceptions.

2: For reads, erases and writes to the Flash data memory, there is no need to insert a NOP into the user code as is done on mid-range devices. The instruction immediately following the "BSF EECON,WR/RD" will be fetched and executed properly.

4.4 DATA MEMORY OPERATION DURING CODE-PROTECT

Data memory can be code-protected by programming the $\overline{\text{CPDF}}$ bit in the Configuration Word (Register 7-1) to '0'.

5.0 I/O PORT

As with any other register, the I/O register(s) can be written and read under program control. However, read instructions (e.g., MOVF PORTB,W) always read the I/O pins independent of the pin's Input/Output modes. On Reset, all I/O ports are defined as input (inputs are at high-impedance) since the I/O control registers are all set.

5.1 PORTB

PORTB is a 6-bit I/O register. Only the low-order 6 bits are used (RB<5:0>). Bits 7 and 6 are unimplemented and read as '0's. Please note that RB3 is an input only pin. The Configuration Word can set several I/O's to alternate functions. When acting as alternate functions, the pins will read as '0' during a port read. Pins RB0, RB1, RB3 and RB4 can be configured with weak pull-ups and also for wake-up on change. The wake-up on change and weak pull-up functions are not pin selectable. If RB3/MCLR is configured as MCLR, weak pull-up is always on and wake-up on change for this pin is not enabled.

5.2 PORTC

PORTC is a 6-bit I/O register. Only the low-order 6 bits are used (RC<5:0>). Bits 7 and 6 are unimplemented and read as '0's.

5.3 TRIS Register

The Output Driver Control register is loaded with the contents of the W register by executing the TRIS f instruction. A '1' from a TRIS register bit puts the corresponding output driver in a High-Impedance mode. A '0' puts the contents of the output data latch on the selected pins, enabling the output buffer. The exceptions are RB3, which is input only and the T0CKI pin, which may be controlled by the OPTION register. See Register 3-2 and Register 3-3.

The TRIS register is "write-only" and is set (output drivers disabled) upon Reset.

TABLE 5-1: WEAK PULL-UP ENABLED PINS

Device	RB0 Weak Pull-up	RB1 Weak Pull-up	RB3 Weak Pull-up(1)	RB4 Weak Pull-up
MCV14A	Yes	Yes	Yes	Yes

Note 1: When MCLREN = 1, the weak pull-up on RB3/MCLR is always enabled.

REGISTER 5-1: PORTB: PORTB REGISTER

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	RB5	RB4	RB3	RB2	RB1	RB0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as '1'

bit 5-0 **RB<5:0>**: PORTB I/O Pin bits
1 = Port pin is >VIH min.
0 = Port pin is <VIL max.

REGISTER 5-2: PORTC: PORTC REGISTER

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	RC5	RC4	RC3	RC2	RC1	RC0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as ‘1’

bit 5-0 **RC<5:0>**: PORTC I/O Pin bits
1 = Port pin is >VIH min.
0 = Port pin is <VIL max.

5.4 I/O Interfacing

The equivalent circuit for an I/O port pin is shown in Figure 5-1. All port pins, except RB3 which is input only, may be used for both input and output operations. For input operations, these ports are non-latching. Any input must be present until read by an input instruction (e.g., `MOVF PORTB, W`). The outputs are latched and remain unchanged until the output latch is rewritten. To use a port pin as output, the corresponding direction control bit in TRIS must be cleared (= `0`). For use as an input, the corresponding TRIS bit must be set. Any I/O pin (except RB3) can be programmed individually as input or output.

FIGURE 5-1: MCV14A EQUIVALENT CIRCUIT FOR A SINGLE I/O PIN

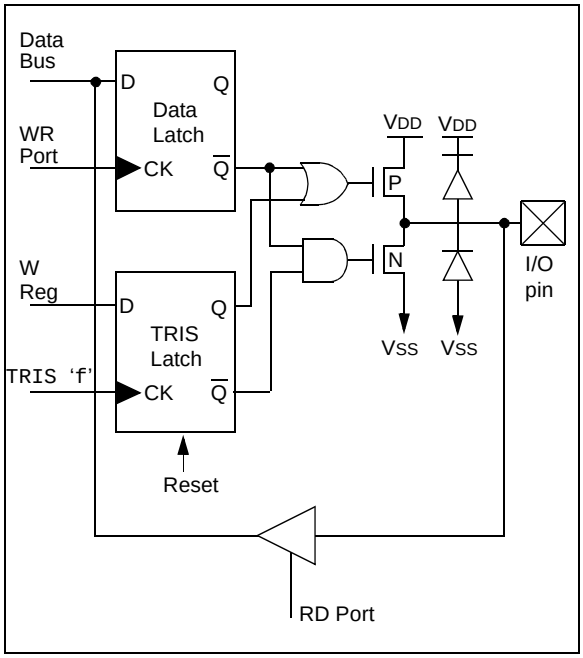


TABLE 5-2: SUMMARY OF PORT REGISTERS

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-On Reset	Value on All Other Resets
N/A	TRIS	—	—	I/O Control Register (PORTB, PORTC)						`--11 1111`	`--11 1111`
N/A	OPTION	$\overline{\text{RBWU}}$	$\overline{\text{RBPU}}$	TOCS	TOSE	PSA	PS2	PS1	PS0	`1111 1111`	`1111 1111`
03h	STATUS	RBWUF	CWUF	PA0	$\overline{\text{TO}}$	$\overline{\text{PD}}$	Z	DC	C	`0001 1xxx`	`q00q quuu`(1)
06h	PORTB	—	—	RB5	RB4	RB3	RB2	RB1	RB0	`--xx xxxx`	`--uu uuuu`
07h	PORTC	—	—	RC5	RC4	RC3	RC2	RC1	RC0	`--xx xxxx`	`--uu uuuu`

Legend: Shaded cells are not used by PORT registers, read as '`0`'. – = unimplemented, read as '`0`', x = unknown, u = unchanged,
q = depends on condition.

Note 1: If Reset was due to wake-up on pin change, then bit 7 = `1`. All other Resets will cause bit 7 = `0`.

TABLE 5-3: I/O PINS ORDER OF PRECEDENCE

Priority	RB0	RB1	RB2	RB3	RC0	RC1	RC2	RC4	RC5
1	AN0	AN1	AN2	RB3/$\overline{\text{MCLR}}$	C2IN+	C2IN-	CV_{REF}	C2OUT	T0CKI
2	C1IN+	C1IN-	C1OUT	—	TRISC	TRISC	TRISC	TRISC	TRISC
3	TRISB	TRISB	TRISB	—	—	—	—	—	—

5.5 I/O Programming Considerations

5.5.1 BIDIRECTIONAL I/O PORTS

Some instructions operate internally as read followed by write operations. The BCF and BSF instructions, for example, read the entire port into the CPU, execute the bit operation and rewrite the result. Caution must be used when these instructions are applied to a port where one or more pins are used as input/outputs. For example, a BSF operation on bit 5 of PORTB will cause all eight bits of PORTB to be read into the CPU, bit 5 to be set and the PORTB value to be written to the output latches. If another bit of PORTB is used as a bidirectional I/O pin (say bit 0) and it is defined as an input at this time, the input signal present on the pin itself would be read into the CPU and rewritten to the data latch of this particular pin, overwriting the previous content. As long as the pin stays in the Input mode, no problem occurs. However, if bit 0 is switched into Output mode later on, the content of the data latch may now be unknown.

Example 5-1 shows the effect of two sequential Read-Modify-Write instructions (e.g., BCF, BSF, etc.) on an I/O port.

A pin actively outputting a high or a low should not be driven from external devices at the same time in order to change the level on this pin ("wired OR", "wired AND"). The resulting high output currents may damage the chip.

EXAMPLE 5-1: READ-MODIFY-WRITE INSTRUCTIONS ON AN I/O PORT(e.g., MCV14A)

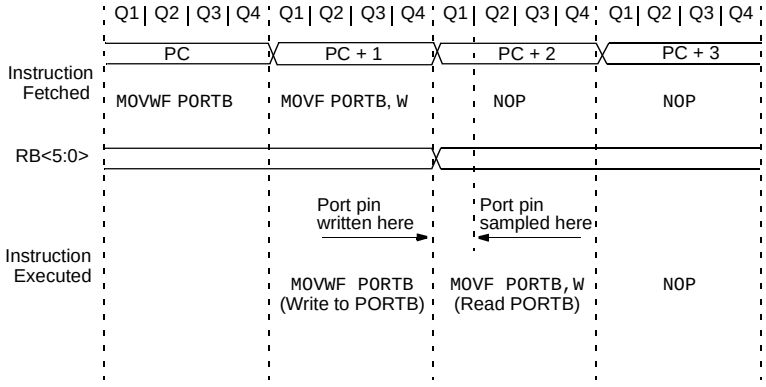
```
;Initial PORTB Settings
;PORTB<5:3> Inputs
;PORTB<2:0> Outputs
;
;                 PORTB latch  PORTB pins
;                 ----------   ----------
  BCF    PORTB, 5 ;--01 -ppp    --11 pppp
  BCF    PORTB, 4 ;--10 -ppp    --11 pppp
  MOVLW  007h;
  TRIS   PORTB    ;--10 -ppp    --11 pppp
;
```

Note 1: The user may have expected the pin values to be '--00 pppp'. The 2nd BCF caused RB5 to be latched as the pin value (High).

5.5.2 SUCCESSIVE OPERATIONS ON I/O PORTS

The actual write to an I/O port happens at the end of an instruction cycle, whereas for reading, the data must be valid at the beginning of the instruction cycle (Figure 5-2). Therefore, care must be exercised if a write followed by a read operation is carried out on the same I/O port. The sequence of instructions should allow the pin voltage to stabilize (load dependent) before the next instruction causes that file to be read into the CPU. Otherwise, the previous state of that pin may be read into the CPU rather than the new state. When in doubt, it is better to separate these instructions with a NOP or another instruction not accessing this I/O port.

FIGURE 5-2: SUCCESSIVE I/O OPERATION



This example shows a write to PORTB followed by a read from PORTB.

Data setup time = (0.25 T_{CY} – T_{PD})

where: T_{CY} = instruction cycle.

T_{PD} = propagation delay

Therefore, at higher clock frequencies, a write followed by a read may be problematic.

6.0 TIMER0 MODULE AND TMR0 REGISTER

The Timer0 module has the following features:

- 8-bit timer/counter register, TMR0
- Readable and writable
- 8-bit software programmable prescaler
- Internal or external clock select:
  - Edge select for external clock

Figure 6-1 is a simplified block diagram of the Timer0 module.

Timer mode is selected by clearing the T0CS bit (OPTION<5>). In Timer mode, the Timer0 module will increment every instruction cycle (without prescaler). If TMR0 register is written, the increment is inhibited for the following two cycles (Figure 6-2 and Figure 6-3). The user can work around this by writing an adjusted value to the TMR0 register.

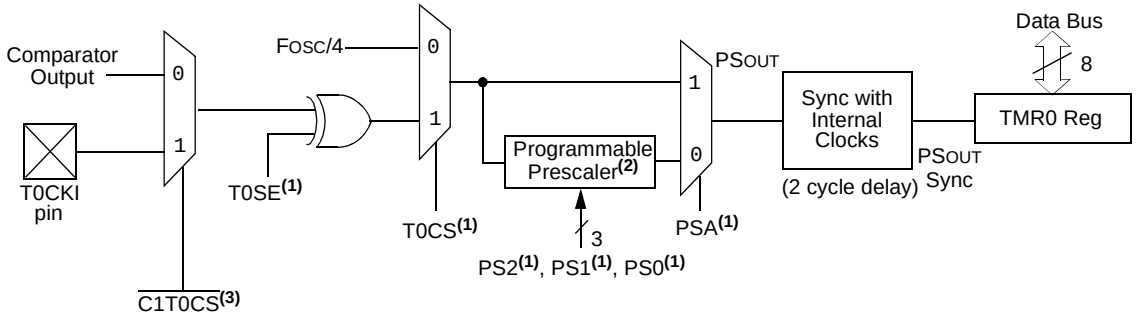
There are two types of Counter mode. The first Counter mode uses the T0CKI pin to increment Timer0. It is selected by setting the T0CS bit (OPTION<5>), setting the C1T0CS bit (CM1CON0<4>) and setting the C1OUTEN bit (CM1CON0<6>). In this mode, Timer0 will increment either on every rising or falling edge of pin T0CKI. The T0SE bit (OPTION<4>) determines the source edge. Clearing the T0SE bit selects the rising edge. Restrictions on the external clock input are discussed in detail in **Section 6.1 "Using Timer0 with an External Clock"**.

The second Counter mode uses the output of the comparator to increment Timer0. It can be entered in two different ways. The first way is selected by setting the T0CS bit (OPTION<5>), and clearing the C1T0CS bit (CM1CON0<4>) (C1OUTEN [CM1CON0<6>] does not affect this mode of operation). This enables an internal connection between the comparator and the Timer0.

The prescaler may be used by either the Timer0 module or the Watchdog Timer, but not both. The prescaler assignment is controlled in software by the control bit, PSA (OPTION<3>). Clearing the PSA bit will assign the prescaler to Timer0. The prescaler is not readable or writable. When the prescaler is assigned to the Timer0 module, prescale values of 1:2, 1:4,..., 1:256 are selectable. **Section 6.2 "Prescaler"** details the operation of the prescaler.

A summary of registers associated with the Timer0 module is found in Table 6-1.

FIGURE 6-1: TIMER0 BLOCK DIAGRAM



Note 1: Bits T0CS, T0SE, PSA, PS2, PS1 and PS0 are located in the OPTION register.
2: The prescaler is shared with the Watchdog Timer.
3: The C1T0CS bit is in the CM1CON0 register.

FIGURE 6-2: TIMER0 TIMING: INTERNAL CLOCK/NO PRESCALE

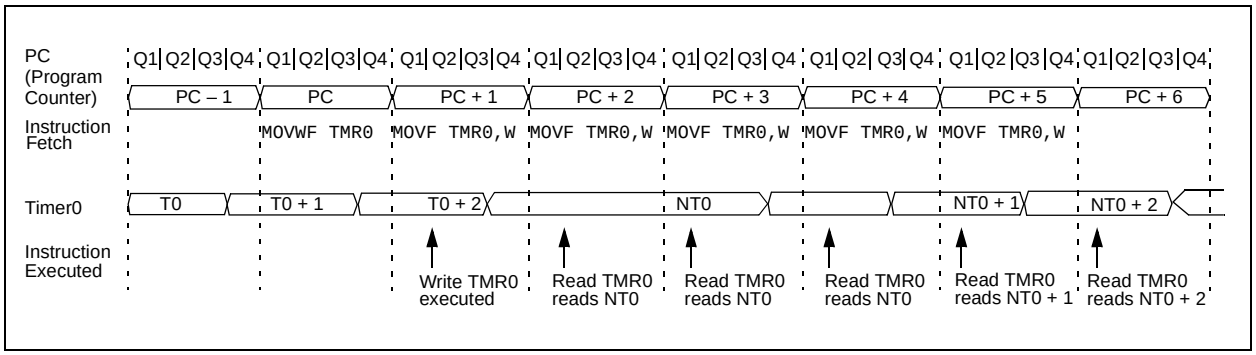


FIGURE 6-3: TIMER0 TIMING: INTERNAL CLOCK/PRESCALE 1:2

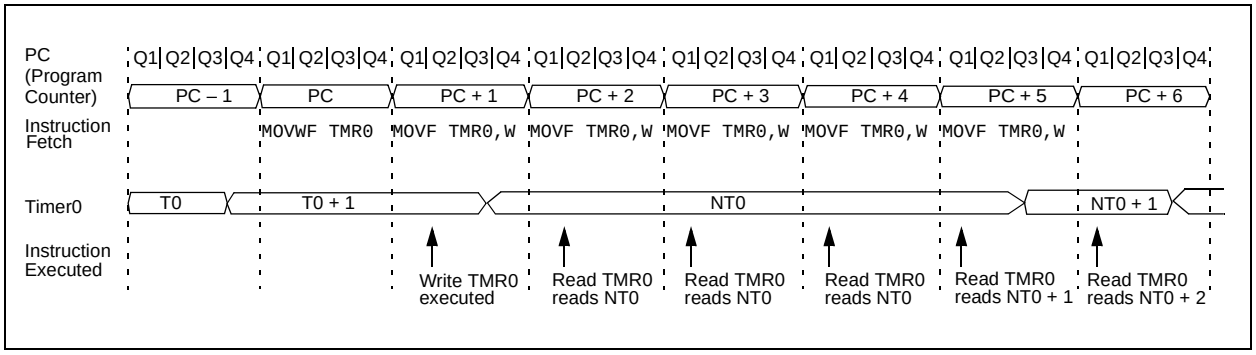


TABLE 6-1: REGISTERS ASSOCIATED WITH TIMER0

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-On Reset	Value on All Other Resets
01h	TMR0	Timer0 – 8-bit Real-Time Clock/Counter								xxxx xxxx	uuuu uuuu
08h	CM1CON0	C1OUT	$\overline{\text{C1OUTEN}}$	C1POL	$\overline{\text{C1T0CS}}$	C1ON	C1NREF	C1PREF	$\overline{\text{C1WU}}$	1111 1111	uuuu uuuu
0Bh	CM2CON0	C2OUT	$\overline{\text{C2OUTEN}}$	C2POL	C2PREF2	C2ON	C2NREF	C2PREF1	$\overline{\text{C2WU}}$	1111 1111	uuuu uuuu
N/A	OPTION	$\overline{\text{RBWU}}$	$\overline{\text{RBPU}}$	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
N/A	TRIS(1)	—	—	I/O Control Register (PORTB, PORTC)						--11 1111	--11 1111

Legend: Shaded cells are not used by Timer0. – = unimplemented, x = unknown, u = unchanged.
Note 1: The TRIS of the T0CKI pin is overridden when T0CS = 1.

6.1 Using Timer0 with an External Clock

When an external clock input is used for Timer0, it must meet certain requirements. The external clock requirement is due to internal phase clock (TOSC) synchronization. Also, there is a delay in the actual incrementing of Timer0 after synchronization.

6.1.1 EXTERNAL CLOCK SYNCHRONIZATION

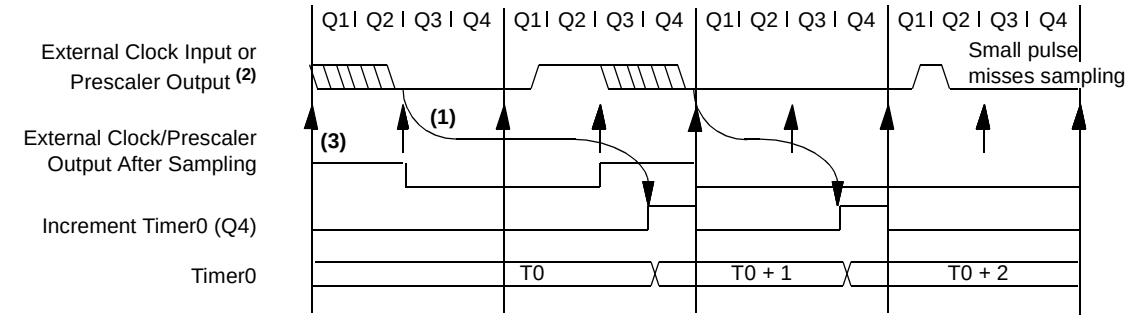
When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks (Figure 6-4). Therefore, it is necessary for T0CKI to be high for at least 2 TOSC (and a small RC delay of 2 Tt0H) and low for at least 2 TOSC (and a small RC delay of 2 Tt0H). Refer to the electrical specification of the desired device.

When a prescaler is used, the external clock input is divided by the asynchronous ripple counter-type prescaler, so that the prescaler output is symmetrical. For the external clock to meet the sampling requirement, the ripple counter must be taken into account. Therefore, it is necessary for T0CKI to have a period of at least 4 TOSC (and a small RC delay of 4 Tt0H) divided by the prescaler value. The only requirement on T0CKI high and low time is that they do not violate the minimum pulse width requirement of Tt0H. Refer to parameters 40, 41 and 42 in the electrical specification of the desired device.

6.1.2 TIMER0 INCREMENT DELAY

Since the prescaler output is synchronized with the internal clocks, there is a small delay from the time the external clock edge occurs to the time the Timer0 module is actually incremented. Figure 6-4 shows the delay from the external clock edge to the timer incrementing.

FIGURE 6-4: TIMER0 TIMING WITH EXTERNAL CLOCK



Note 1: Delay from clock input change to Timer0 increment is 3 TOSC to 7 TOSC. (Duration of Q = TOSC). Therefore, the error in measuring the interval between two edges on Timer0 input = ±4 TOSC max.

2: External clock if no prescaler selected; prescaler output otherwise.

3: The arrows indicate the points in time where sampling occurs.

6.2 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module or as a postscaler for the Watchdog Timer (WDT), respectively (see **Section 7.6 "Watchdog Timer (WDT)"**). For simplicity, this counter is being referred to as "prescaler" throughout this data sheet.

Note:	The prescaler may be used by either the Timer0 module or the WDT, but not both. Thus, a prescaler assignment for the Timer0 module means that there is no prescaler for the WDT and vice versa.

The PSA and PS<2:0> bits (OPTION<3:0>) determine prescaler assignment and prescale ratio.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (e.g., `CLRF 1`, `MOVWF 1`, `BSF 1, x`, etc.) will clear the prescaler. When assigned to WDT, a `CLRWDT` instruction will clear the prescaler along with the WDT. The prescaler is neither readable nor writable. On a Reset, the prescaler contains all '0's.

6.2.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control (i.e., it can be changed "on-the-fly" during program execution). To avoid an unintended device Reset, the following instruction sequence (Example 6-1) must be executed when changing the prescaler assignment from Timer0 to the WDT.

EXAMPLE 6-1: CHANGING PRESCALER (TIMER0 → WDT)

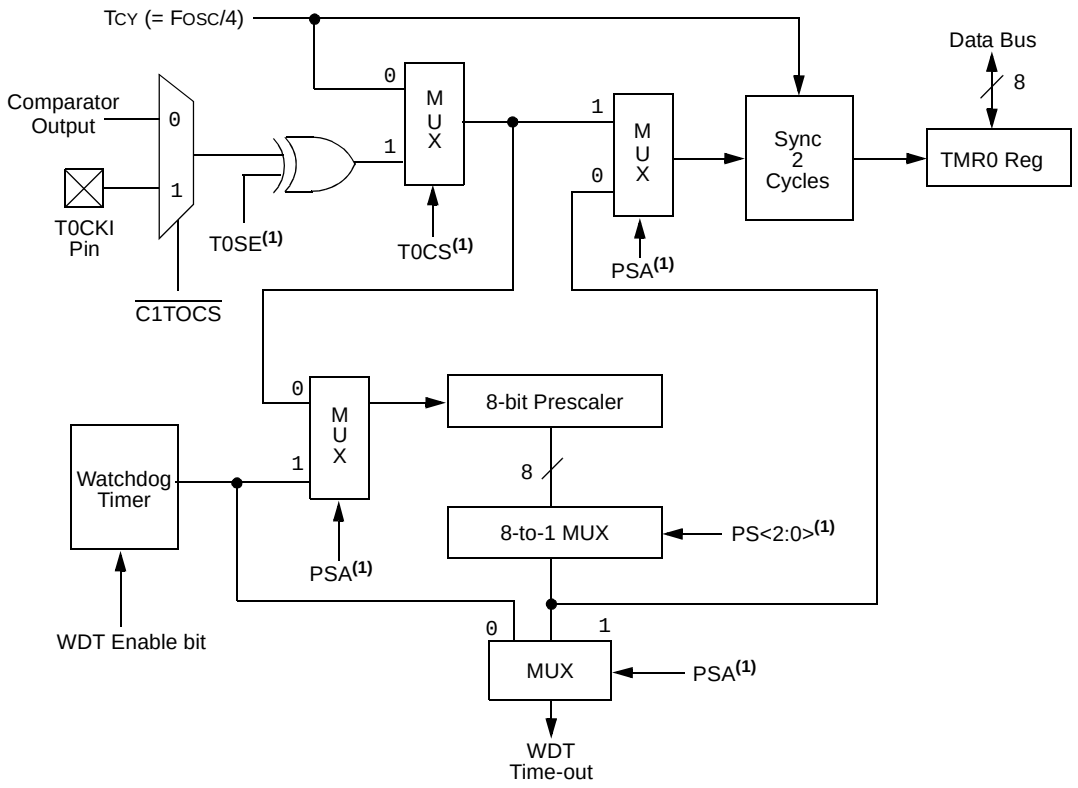
```
CLRWDT              ;Clear WDT
CLRF   TMR0         ;Clear TMR0 & Prescaler
MOVLW  '00xx1111'b  ;These 3 lines (5, 6, 7)
OPTION              ;are required only if
                    ;desired
CLRWDT              ;PS<2:0> are 000 or 001
MOVLW  '00xx1xxx'b  ;Set Postscaler to
OPTION              ;desired WDT rate
```

To change the prescaler from the WDT to the Timer0 module, use the sequence shown in Example 6-2. This sequence must be used even if the WDT is disabled. A `CLRWDT` instruction should be executed before switching the prescaler.

EXAMPLE 6-2: CHANGING PRESCALER (WDT → TIMER0)

```
CLRWDT              ;Clear WDT and
                    ;prescaler
MOVLW  'xxxx0xxx'   ;Select TMR0, new
                    ;prescale value and
                    ;clock source
OPTION
```


FIGURE 6-5: BLOCK DIAGRAM OF THE TIMER0/WDT PRESCALER



Note 1: T0CS, T0SE, PSA, PS<2:0> are bits in the OPTION register.

NOTES:

7.0 SPECIAL FEATURES OF THE CPU

What sets a microcontroller apart from other processors are special circuits that deal with the needs of real-time applications. The MCV14A microcontrollers have a host of such features intended to maximize system reliability, minimize cost through elimination of external components, provide power-saving operating modes and offer code protection. These features are:

- Oscillator Selection
- Reset:
  - Power-on Reset (POR)
  - Device Reset Timer (DRT)
  - Wake-up from Sleep on Pin Change
- Watchdog Timer (WDT)
- Sleep
- Code Protection
- ID Locations
- In-Circuit Serial Programming™
- Clock Out

The MCV14A device has a Watchdog Timer, which can be shut off only through Configuration bit WDTE. It runs off of its own RC oscillator for added reliability. If using HS, XT or LP selectable oscillator options, there is always an 18 ms (nominal) delay provided by the Device Reset Timer (DRT), intended to keep the chip in Reset until the crystal oscillator is stable. If using INTRC or EXTRC, there is a 1 ms delay only on V_{DD} power-up. With this timer on-chip, most applications need no external Reset circuitry.

The Sleep mode is designed to offer a very low current Power-Down mode. The user can wake-up from Sleep through a change on input pins or through a Watchdog Timer time-out. Several oscillator options are also made available to allow the part to fit the application, including an internal 4/8 MHz oscillator. The EXTRC oscillator option saves system cost while the LP crystal option saves power. A set of Configuration bits are used to select various options.

7.1 Configuration Bits

The MCV14A Configuration Words consist of 12 bits. Configuration bits can be programmed to select various device configurations. Three bits are for the selection of the oscillator type; one bit is the Watchdog Timer enable bit, one bit is the $\overline{MCLR}$ enable bit and one bit is for code protection (Register 7-1).

REGISTER 7-1: CONFIG: CONFIGURATION WORD REGISTER

$\overline{\text{CPDF}}$	IOSCFS	MCLRE	$\overline{\text{CP}}$	WDTE	FOSC2	FOSC1	FOSC0
bit 7							bit 0

bit 7 **$\overline{\text{CPDF}}$:** Code Protection bit – Flash Data Memory
1 = Code protection off
0 = Code protection on

bit 6 **IOSCFS:** Internal Oscillator Frequency Select bit
1 = 8 MHz INTOSC speed
0 = 4 MHz INTOSC speed

bit 5 **MCLRE:** Master Clear Enable bit
1 = RB3/$\overline{\text{MCLR}}$ pin functions as $\overline{\text{MCLR}}$
0 = RB3/$\overline{\text{MCLR}}$ pin functions as RB3, $\overline{\text{MCLR}}$ internally tied to V_{DD}

bit 4 **$\overline{\text{CP}}$:** Code Protection bit – User Program Memory
1 = Code protection off
0 = Code protection on

bit 3 **WDTE:** Watchdog Timer Enable bit
1 = WDT enabled
0 = WDT disabled

bit 2-0 **FOSC<2:0>:** Oscillator Selection bits
000 = LP oscillator and 18 ms DRT
001 = XT oscillator and 18 ms DRT
010 = HS oscillator and 18 ms DRT
011 = EC oscillator with RB4 function on RB4/OSC2/CLKOUT and 1 ms DRT(1)
100 = INTRC with RB4 function on RB4/OSC2/CLKOUT and 1 ms DRT(1)
101 = INTRC with CLKOUT function on RB4/OSC2/CLKOUT and 1 ms DRT(1)
110 = EXTRC with RB4 function on RB4/OSC2/CLKOUT and 1 ms DRT(1)
111 = EXTRC with CLKOUT function on RB4/OSC2/CLKOUT and 1 ms DRT(1)

Note 1: DRT length (18 ms or 1 ms) is a function of Clock mode selection. It is the responsibility of the application designer to ensure the use of either 18 ms (nominal) DRT or the 1 ms (nominal) DRT will result in acceptable operation. Refer to **Section 11.1 "DC Characteristics: MCV14A (Industrial)"** and **Section 11.2 "DC Characteristics: MCV14A"** for V_{DD} rise time and stability requirements for this mode of operation.

7.2 Oscillator Configurations

7.2.1 OSCILLATOR TYPES

The MCV14A device can be operated in up to six different Oscillator modes. The user can program up to three Configuration bits (FOSC<2:0>). To select one of these modes:

- LP: Low-Power Crystal
- XT: Crystal/Resonator
- HS: High-Speed Crystal/Resonator
- INTRC: Internal 4/8 MHz Oscillator
- EXTRC: External Resistor/Capacitor
- EC: External High-Speed Clock Input

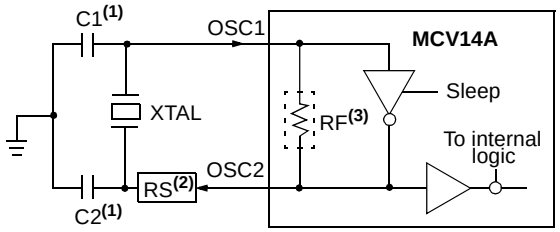
7.2.2 CRYSTAL OSCILLATOR/CERAMIC RESONATORS

In HS, XT or LP modes, a crystal or ceramic resonator is connected to the RB5/OSC1/CLKIN and RB4/OSC2/CLKOUT pins to establish oscillation (Figure 7-1). The MCV14A oscillator designs require the use of a parallel cut crystal. Use of a series cut crystal may give a frequency out of the crystal manufacturers specifications. When in HS, XT or LP modes, the device can have an external clock source drive the RB5/OSC1/CLKIN pin (Figure 7-2). This pin should be left open and unloaded. Also when using this mode, the external clock should observe the frequency limits for the Clock mode chosen (HS, XT or LP).

Note 1: This device has been designed to perform to the parameters of its data sheet. It has been tested to an electrical specification designed to determine its conformance with these parameters. Due to process differences in the manufacture of this device, this device may have different performance characteristics than its earlier version. These differences may cause this device to perform differently in your application than the earlier version of this device.

2: The user should verify that the device oscillator starts and performs as expected. Adjusting the loading capacitor values and/or the Oscillator mode may be required.

FIGURE 7-1: CRYSTAL OPERATION (OR CERAMIC RESONATOR) (HS, XT OR LP OSC CONFIGURATION)

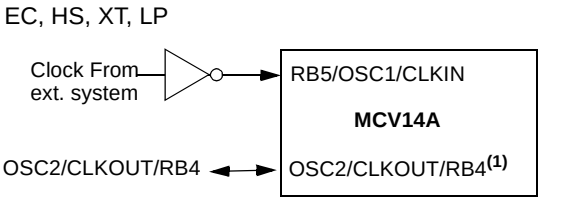


Note 1: See Capacitor Selection tables for recommended values of C1 and C2.

2: A series resistor (RS) may be required for AT strip cut crystals.

3: RF approx. value = 10 MΩ.

FIGURE 7-2: EXTERNAL CLOCK INPUT OPERATION (HS, XT, LP OR EC OSC CONFIGURATION)



Note 1: RB4 is available in EC mode only.

TABLE 7-1: CAPACITOR SELECTION FOR CERAMIC RESONATORS

Osc Type	Resonator Freq.	Cap. Range C1	Cap. Range C2
XT	4.0 MHz	30 pF	30 pF
HS	16 MHz	10-47 pF	10-47 pF

Note 1: These values are for design guidance only. Since each resonator has its own characteristics, the user should consult the resonator manufacturer for appropriate values of external components.

TABLE 7-2: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR(2)

Osc Type	Resonator Freq.	Cap. Range C1	Cap. Range C2
LP	32 kHz(1)	15 pF	15 pF
XT	200 kHz 1 MHz 4 MHz	47-68 pF 15 pF 15 pF	47-68 pF 15 pF 15 pF
HS	20 MHz	15-47 pF	15-47 pF

Note 1: For V_{DD} > 4.5V, C1 = C2 ≈ 30 pF is recommended.

2: These values are for design guidance only. Rs may be required to avoid overdriving crystals with low drive level specification. Since each crystal has its own characteristics, the user should consult the crystal manufacturer for appropriate values of external components.

7.2.3 EXTERNAL CRYSTAL OSCILLATOR CIRCUIT

Either a prepackaged oscillator or a simple oscillator circuit with TTL gates can be used as an external crystal oscillator circuit. Prepackaged oscillators provide a wide operating range and better stability. A well-designed crystal oscillator will provide good performance with TTL gates. Two types of crystal oscillator circuits can be used: one with parallel resonance, or one with series resonance.

Figure 7-3 shows implementation of a parallel resonant oscillator circuit. The circuit is designed to use the fundamental frequency of the crystal. The 74AS04 inverter performs the 180-degree phase shift that a parallel oscillator requires. The 4.7 kΩ resistor provides the negative feedback for stability. The 10 kΩ potentiometers bias the 74AS04 in the linear region. This circuit could be used for external oscillator designs.

FIGURE 7-3: EXTERNAL PARALLEL RESONANT CRYSTAL OSCILLATOR CIRCUIT

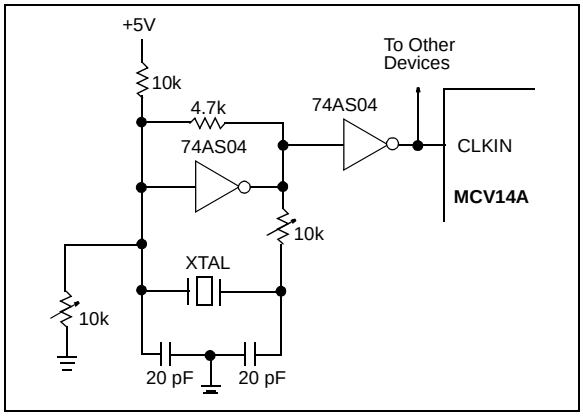
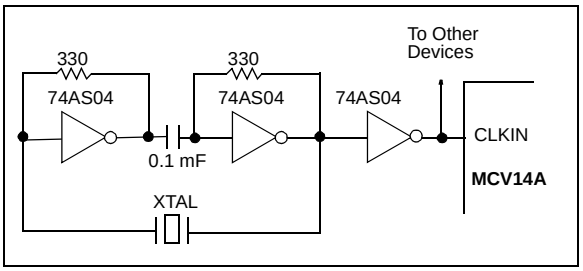


Figure 7-4 shows a series resonant oscillator circuit. This circuit is also designed to use the fundamental frequency of the crystal. The inverter performs a 180-degree phase shift in a series resonant oscillator circuit. The 330 Ω resistors provide the negative feedback to bias the inverters in their linear region.

FIGURE 7-4: EXTERNAL SERIES RESONANT CRYSTAL OSCILLATOR CIRCUIT



7.2.4 EXTERNAL RC OSCILLATOR

For timing insensitive applications, the RC device option offers additional cost savings. The RC oscillator frequency is a function of the supply voltage, the resistor (R_{EXT}) and capacitor (C_{EXT}) values, and the operating temperature. In addition to this, the oscillator frequency will vary from unit-to-unit due to normal process parameter variation. Furthermore, the difference in lead frame capacitance between package types will also affect the oscillation frequency, especially for low C_{EXT} values. The user also needs to take into account variation due to tolerance of external R and C components used.

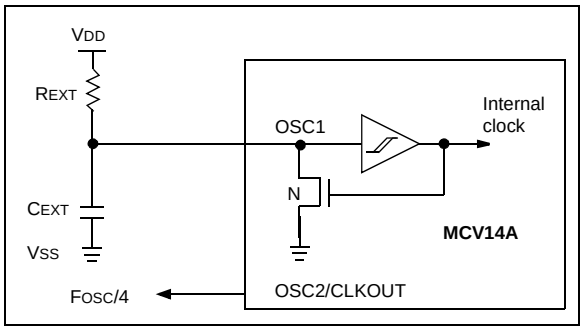
Figure 7-5 shows how the R/C combination is connected to the MCV14A device. For R_{EXT} values below 3.0 kΩ, the oscillator operation may become unstable, or stop completely. For very high R_{EXT} values (e.g., 1 MΩ), the oscillator becomes sensitive to noise, humidity and leakage. Thus, we recommend keeping R_{EXT} between 5.0 kΩ and 100 kΩ.

Although the oscillator will operate with no external capacitor (C_{EXT} = 0 pF), we recommend using values above 20 pF for noise and stability reasons. With no or small external capacitance, the oscillation frequency can vary dramatically due to changes in external capacitances, such as PCB trace capacitance or package lead frame capacitance.

Section 11.0 "Electrical Characteristics" shows RC frequency variation from part-to-part due to normal process variation. The variation is larger for larger values of R (since leakage current variation will affect RC frequency more for large R) and for smaller values of C (since variation of input capacitance will affect RC frequency more).

Also, see the Electrical Specifications section for variation of oscillator frequency due to VDD for given REXT/CEXT values, as well as frequency variation due to operating temperature for given R, C and VDD values.

FIGURE 7-5: EXTERNAL RC OSCILLATOR MODE



7.2.5 INTERNAL 4/8 MHz RC OSCILLATOR

The internal RC oscillator provides a fixed 4/8 MHz (nominal) system clock at VDD = 5V and 25°C, (see **Section 11.0 "Electrical Characteristics"** for information on variation over voltage and temperature).

In addition, a calibration instruction is programmed into the last address of memory, which contains the calibration value for the internal RC oscillator. This location is always non-code protected, regardless of the code-protect settings. This value is programmed as a `MOVLW XX` instruction where XX is the calibration value, and is placed at the Reset vector. This will load the W register with the calibration value upon Reset and the PC will then roll over to the users program at address 0x000. The user then has the option of writing the value to the OSCCAL Register (05h) or ignoring it.

OSCCAL, when written to with the calibration value, will "trim" the internal oscillator to remove process variation from the oscillator frequency.

Note:	Erasing the device will also erase the pre-programmed internal calibration value for the internal oscillator. The calibration value must be read prior to erasing the part so it can be reprogrammed correctly later.

For the MCV14A device, only bits <7:1> of OSCCAL are used for calibration. See Register 3-3 for more information.

Note:	The bit 0 of the OSCCAL register is unimplemented and should be written as '0' when modifying OSCCAL for compatibility with future devices.

7.3 Reset

The device differentiates between various kinds of Reset:

- Power-on Reset (POR)
- $\overline{\text{MCLR}}$ Reset during normal operation
- $\overline{\text{MCLR}}$ Reset during Sleep
- WDT Time-out Reset during normal operation
- WDT Time-out Reset during Sleep
- Wake-up from Sleep on pin change

Some registers are not reset in any way, they are unknown on POR and unchanged in any other Reset. Most other registers are reset to "Reset state" on Power-on Reset (POR), $\overline{\text{MCLR}}$, WDT or Wake-up on pin change Reset during normal operation. They are not affected by a WDT Reset during Sleep or $\overline{\text{MCLR}}$ Reset during Sleep, since these Resets are viewed as resumption of normal operation. The exceptions to this are $\overline{\text{TO}}$, $\overline{\text{PD}}$ and RBWUF bits. They are set or cleared differently in different Reset situations. These bits are used in software to determine the nature of Reset. See Table 7-3 for a full description of Reset states of all registers.

TABLE 7-3: RESET CONDITIONS FOR REGISTERS

Register	Address	Power-on Reset	$\overline{\text{MCLR}}$ Reset, WDT Time-out, Wake-up On Pin Change
W	–	qqqq qqq0(1)	qqqq qqq0(1)
INDF	00h	xxxx xxxx	uuuu uuuu
TMR0	01h	xxxx xxxx	uuuu uuuu
PCL	02h	1111 1111	1111 1111
STATUS	03h	0001 1xxx	qq0q quuu(2)
FSR	04h	100x xxxx	1uuu uuuu
OSCCAL	05h	1111 111-	uuuu uuu-
PORTB	06h	--xx xxxx	--uu uuuu
PORTC	07h	--xx xxxx	--uu uuuu
CMICON0	08h	q111 1111	quuu uuuu
ADCON0	09h	1111 1100	1111 1100
ADRES	0Ah	xxxx xxxx	uuuu uuuu
CM2CON0	0Bh	q111 1111	quuu uuuu
VRCON	0Ch	001-1111	uuu-uuuu
OPTION	–	1111 1111	1111 1111
TRISB	–	--11 1111	--11 1111
TRISC	–	--11 1111	--11 1111
EECON	21h/61h	---0 x000	---0 q000
EEDATA	25h/65h	xxxx xxxx	uuuu uuuu
EEADR	26h/66h	--xx xxxx	--uu uuuu

Legend: u = unchanged, x = unknown, – = unimplemented bit, read as '0', q = value depends on condition.

Note 1: Bits <7:1> of W register contain oscillator calibration values due to MOVLW XX instruction at top of memory.

2: See Table 7-4 for Reset value for specific conditions.

TABLE 7-4: RESET CONDITION FOR SPECIAL REGISTERS

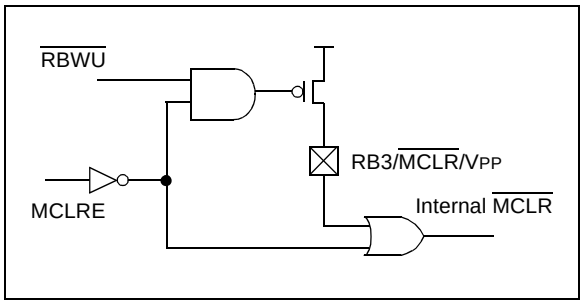
	STATUS Addr: 03h	PCL Addr: 02h
Power-on Reset	0001 1xxx	1111 1111
$\overline{\text{MCLR}}$ Reset during normal operation	000u uuuu	1111 1111
$\overline{\text{MCLR}}$ Reset during Sleep	0001 0uuu	1111 1111
WDT Reset during Sleep	0000 0uuu	1111 1111
WDT Reset normal operation	0000 uuuu	1111 1111
Wake-up from Sleep on pin change	1001 0uuu	1111 1111

Legend: u = unchanged, x = unknown, – = unimplemented bit, read as '0'.

7.3.1 $\overline{\text{MCLR}}$ ENABLE

This Configuration bit, when unprogrammed (left in the '1' state), enables the external $\overline{\text{MCLR}}$ function. When programmed, the $\overline{\text{MCLR}}$ function is tied to the internal V_{DD} and the pin is assigned to be a I/O. See Figure 7-6.

FIGURE 7-6: $\overline{\text{MCLR}}$ SELECT



7.4 Power-on Reset (POR)

The MCV14A device incorporates an on-chip Power-on Reset (POR) circuitry, which provides an internal chip Reset for most power-up situations.

The on-chip POR circuit holds the chip in Reset until V_{DD} has reached a high enough level for proper operation. To take advantage of the internal POR, program the RB3/$\overline{\text{MCLR}}$/V_{PP} pin as $\overline{\text{MCLR}}$ and tie through a resistor to V_{DD}, or program the pin as RB3. An internal weak pull-up resistor is implemented using a transistor (refer to Table 11-5 for the pull-up resistor ranges). This will eliminate external RC components usually needed to create a Power-on Reset. A maximum rise time for V_{DD} is specified. See **Section 11.0 "Electrical Characteristics"** for details.

When the device starts normal operation (exit the Reset condition), device operating parameters (voltage, frequency, temperature,...) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating parameters are met.

A simplified block diagram of the on-chip Power-on Reset circuit is shown in Figure 7-7.

The Power-on Reset circuit and the Device Reset Timer (see **Section 7.5 "Device Reset Timer (DRT)"**) circuit are closely related. On power-up, the Reset latch is set and the DRT is reset. The DRT timer begins counting once it detects $\overline{\text{MCLR}}$ to be high. After the time-out period, which is typically 18 ms or 1 ms, it will reset the Reset latch and thus end the on-chip Reset signal.

A power-up example where $\overline{\text{MCLR}}$ is held low is shown in Figure 7-8. V_{DD} is allowed to rise and stabilize before bringing $\overline{\text{MCLR}}$ high. The chip will actually come out of Reset T_{DRT} msec after $\overline{\text{MCLR}}$ goes high.

In Figure 7-9, the on-chip Power-on Reset feature is being used ($\overline{\text{MCLR}}$ and V_{DD} are tied together or the pin is programmed to be RB3. The V_{DD} is stable before the start-up timer times out and there is no problem in getting a proper Reset. However, Figure 7-10 depicts a problem situation where V_{DD} rises too slowly. The time between when the DRT senses that $\overline{\text{MCLR}}$ is high and when $\overline{\text{MCLR}}$ and V_{DD} actually reach their full value, is too long. In this situation, when the start-up timer times out, V_{DD} has not reached the V_{DD} (min) value and the chip may not function correctly. For such situations, we recommend that external RC circuits be used to achieve longer POR delay times (Figure 7-9).

Note:	When the device starts normal operation (exit the Reset condition), device operating parameters (voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

For additional information, refer to Application Notes AN522, *"Power-Up Considerations"* (DS00522) and AN607, *"Power-up Trouble Shooting"* (DS00607).

FIGURE 7-7: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT

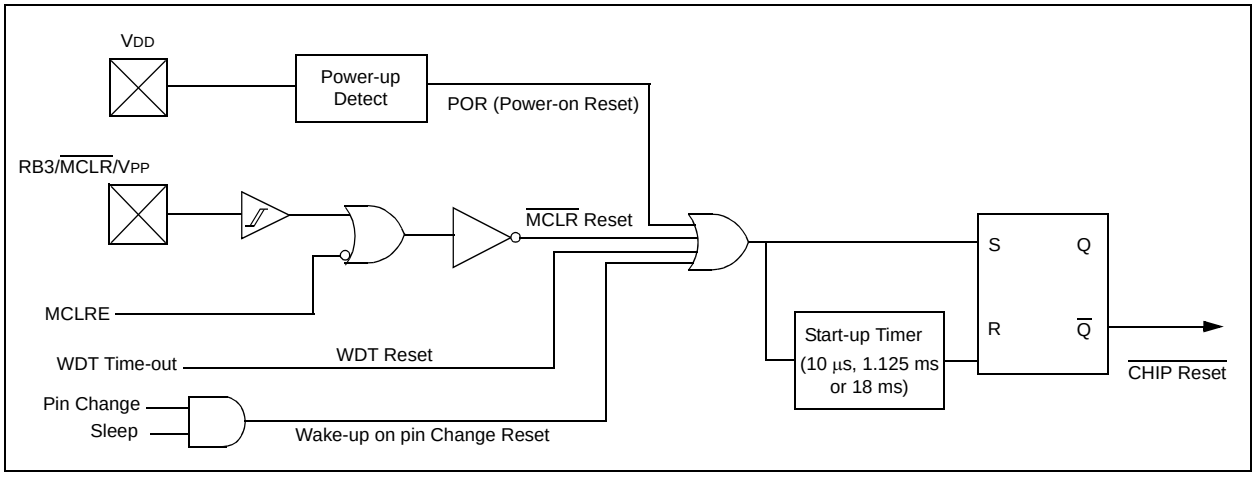


FIGURE 7-8: TIME-OUT SEQUENCE ON POWER-UP (MCLR PULLED LOW)

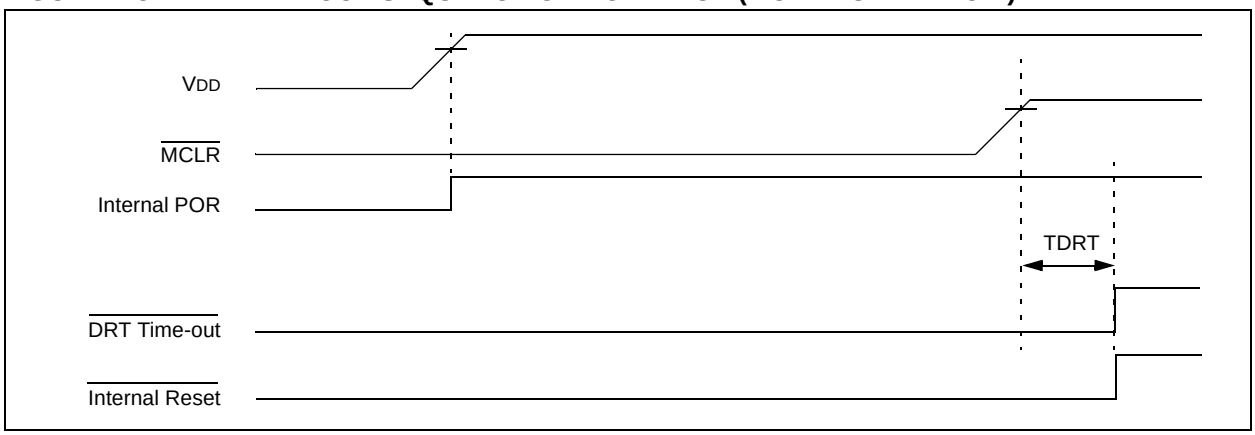


FIGURE 7-9: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD): FAST VDD RISE TIME

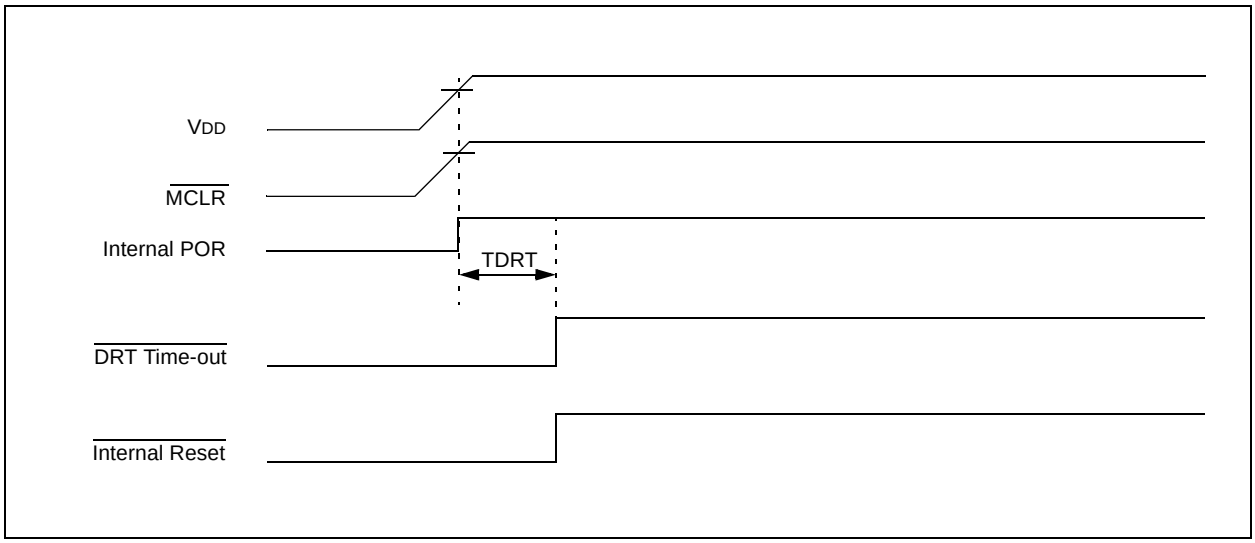
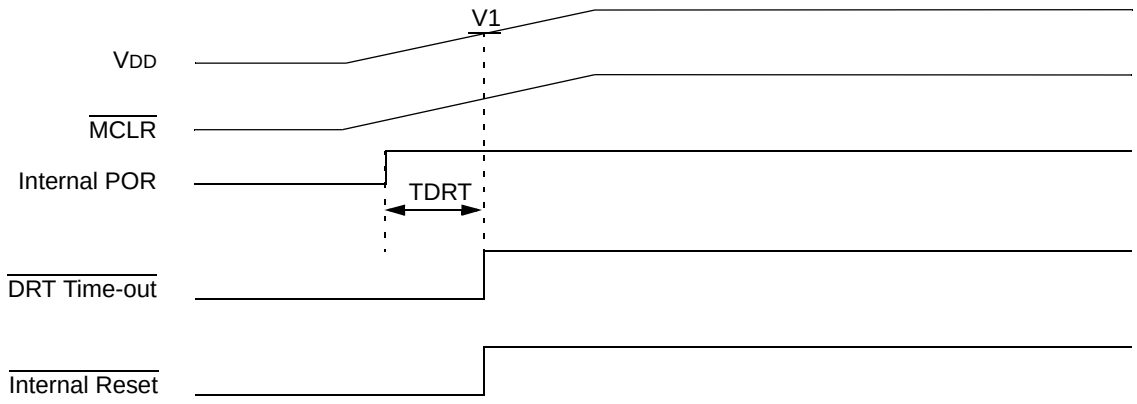


FIGURE 7-10: TIME-OUT SEQUENCE ON POWER-UP ($\overline{MCLR}$ TIED TO V_{DD}): SLOW V_{DD} RISE TIME



Note: When V_{DD} rises slowly, the T_{DRT} time-out expires long before V_{DD} has reached its final value. In this example, the chip will reset properly if, and only if, V1 ≥ V_{DD} min.

7.5 Device Reset Timer (DRT)

On the MCV14A device, the DRT runs any time the device is powered up. DRT runs from Reset and varies based on oscillator selection and Reset type (see Table 7-5).

The DRT operates on an internal RC oscillator. The processor is kept in Reset as long as the DRT is active. The DRT delay allows VDD to rise above VDD min. and for the oscillator to stabilize.

Oscillator circuits based on crystals or ceramic resonators require a certain time after power-up to establish a stable oscillation. The on-chip DRT keeps the device in a Reset condition after MCLR has reached a logic high (VIH MCLR) level. Programming RB3/MCLR/VPP as MCLR and using an external RC network connected to the MCLR input is not required in most cases. This allows savings in cost-sensitive and/or space restricted applications, as well as allowing the use of the RB3/MCLR/VPP pin as a general purpose input.

The Device Reset Time delays will vary from chip-to-chip due to VDD, temperature and process variation. See AC parameters for details.

The DRT will also be triggered upon a Watchdog Timer time-out from Sleep. This is particularly important for applications using the WDT to wake from Sleep mode automatically.

Reset sources are POR, MCLR, WDT time-out and wake-up on pin change. See **Section 7.8.2 "Wake-up from Sleep", Notes 1, 2** and **3**.

7.6 Watchdog Timer (WDT)

The Watchdog Timer (WDT) is a free running on-chip RC oscillator, which does not require any external components. This RC oscillator is separate from the external RC oscillator of the RB5/OSC1/CLKIN pin and the internal 4/8 MHz oscillator. This means that the WDT will run even if the main processor clock has been stopped, for example, by execution of a SLEEP instruction. During normal operation or Sleep, a WDT Reset or wake-up Reset, generates a device Reset.

The TO bit (STATUS<4>) will be cleared upon a Watchdog Timer Reset.

The WDT can be permanently disabled by programming the configuration WDTE as a '0' (see **Section 7.1 "Configuration Bits"**). Refer to the MCV14A Programming Specifications to determine how to access the Configuration Word.

TABLE 7-5: TYPICAL DRT PERIODS

Oscillator Configuration	POR Reset	Subsequent Resets
HS, XT, LP	18 ms	18 ms
EC	1.125 ms	10 μs
INTOSC, EXTRC	1.125 ms	10 μs

7.6.1 WDT PERIOD

The WDT has a nominal time-out period of 18 ms, (with no prescaler). If a longer time-out period is desired, a prescaler with a division ratio of up to 1:128 can be assigned to the WDT (under software control) by writing to the OPTION register. Thus, a time-out period of a nominal 2.3 seconds can be realized. These periods vary with temperature, VDD and part-to-part process variations (see DC specs).

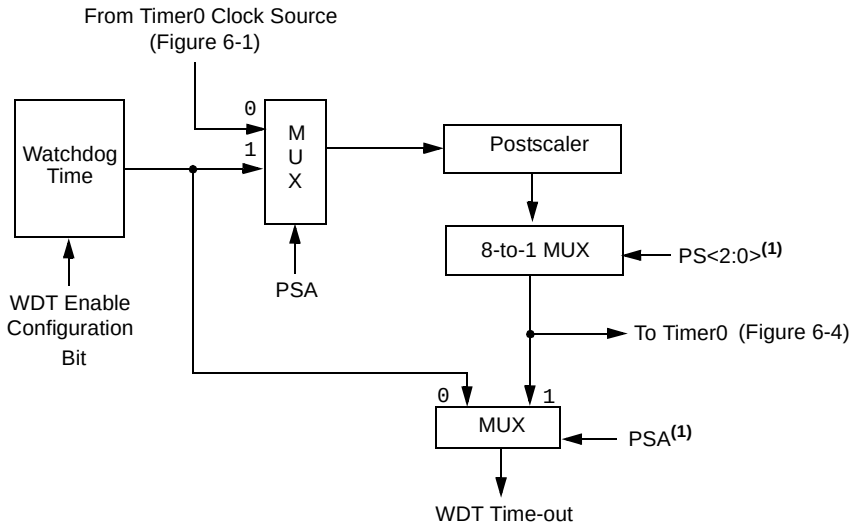
Under worst-case conditions (VDD = Min., Temperature = Max., max. WDT prescaler), it may take several seconds before a WDT time-out occurs.

7.6.2 WDT PROGRAMMING CONSIDERATIONS

The CLRWDT instruction clears the WDT and the postscaler, if assigned to the WDT, and prevents it from timing out and generating a device Reset.

The SLEEP instruction resets the WDT and the postscaler, if assigned to the WDT. This gives the maximum Sleep time before a WDT wake-up Reset.

FIGURE 7-11: WATCHDOG TIMER BLOCK DIAGRAM



Note 1: PSA, PS<2:0> are bits in the OPTION register.

TABLE 7-6: SUMMARY OF REGISTERS ASSOCIATED WITH THE WATCHDOG TIMER

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-On Reset	Value on All Other Resets
N/A	OPTION	$\overline{\text{RBWU}}$	$\overline{\text{RBPU}}$	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111

Legend: Shaded boxes = Not used by Watchdog Timer.

7.7 Time-out Sequence, Power-down and Wake-up from Sleep Status Bits (TO, PD, RBWUF)

The $\overline{TO}$, $\overline{PD}$ and RBWUF bits in the STATUS register can be tested to determine if a Reset condition has been caused by a power-up condition, a $\overline{MCLR}$ or Watchdog Timer (WDT) Reset.

TABLE 7-7: $\overline{TO}$/$\overline{PD}$/RBWUF STATUS AFTER RESET

RBWUF	$\overline{TO}$	$\overline{PD}$	Reset Caused By
0	0	0	WDT wake-up from Sleep
0	0	u	WDT time-out (not from Sleep)
0	1	0	$\overline{MCLR}$ wake-up from Sleep
0	1	1	Power-up
0	u	u	$\overline{MCLR}$ not during Sleep
1	1	0	Wake-up from Sleep on pin change

Legend: u = unchanged

Note 1: The $\overline{TO}$, $\overline{PD}$ and RBWUF bits maintain their status (u) until a Reset occurs. A low-pulse on the $\overline{MCLR}$ input does not change the $\overline{TO}$, $\overline{PD}$ and RBWUF Status bits.

7.8 Power-down Mode (Sleep)

A device may be powered down (Sleep) and later powered up (wake-up from Sleep).

7.8.1 SLEEP

The Power-Down mode is entered by executing a SLEEP instruction.

If enabled, the Watchdog Timer will be cleared but keeps running, the $\overline{TO}$ bit (STATUS<4>) is set, the $\overline{PD}$ bit (STATUS<3>) is cleared and the oscillator driver is turned off. The I/O ports maintain the status they had before the SLEEP instruction was executed (driving high, driving low or high-impedance).

Note:	A Reset generated by a WDT time-out does not drive the $\overline{MCLR}$ pin low.

For lowest current consumption while powered down, the T0CKI input should be at V_{DD} or V_{SS} and the RB3/$\overline{MCLR}$/V_{PP} pin must be at a logic high level if $\overline{MCLR}$ is enabled.

7.8.2 WAKE-UP FROM SLEEP

The device can wake-up from Sleep through one of the following events:

1. An external Reset input on RB3/$\overline{MCLR}$/V_{PP} pin, when configured as $\overline{MCLR}$.
2. A Watchdog Timer Time-out Reset (if WDT was enabled).
3. A change on input pin RB0, RB1, RB3 or RB4 when wake-up on change is enabled.

These events cause a device Reset. The $\overline{TO}$, $\overline{PD}$ and RBWUF bits can be used to determine the cause of device Reset. The $\overline{TO}$ bit is cleared if a WDT time-out occurred (and caused wake-up). The $\overline{PD}$ bit, which is set on power-up, is cleared when SLEEP is invoked. The RBWUF bit indicates a change in state while in Sleep at pins RB0, RB1, RB3 or RB4 (since the last file or bit operation on RB port).

Note:	**Caution:** Right before entering Sleep, read the input pins. When in Sleep, wake-up occurs when the values at the pins change from the state they were in at the last reading. If a wake-up on change occurs and the pins are not read before re-entering Sleep, a wake-up will occur immediately even if no pins change while in Sleep mode.

The WDT is cleared when the device wakes from Sleep, regardless of the wake-up source.

7.9 Program Verification/Code Protection

If the code protection bit has not been programmed, the on-chip program memory can be read out for verification purposes.

The first 64 locations and the last location (OSCCAL) can be read, regardless of the code protection bit setting.

The last memory location can be read regardless of the code protection bit setting on the MCV14A device.

7.10 ID Locations

Four memory locations are designated as ID locations where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution, but are readable and writable during Program/Verify.

Use only the lower 4 bits of the ID locations and always program the upper 8 bits as '0's.

7.11 In-Circuit Serial Programming™

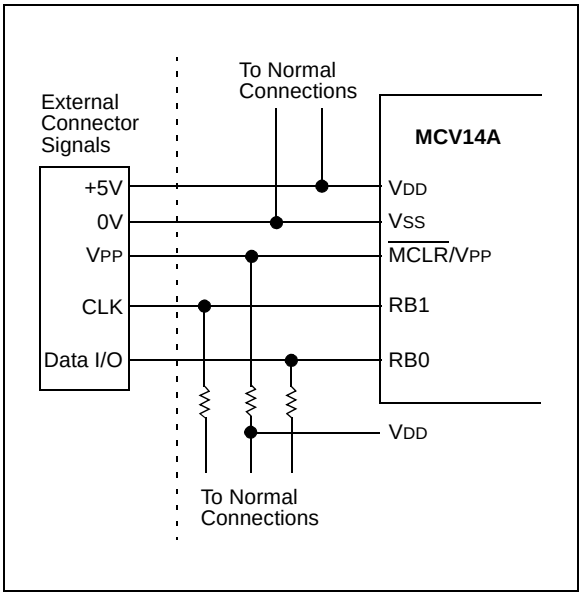
The MCV14A microcontroller can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware, or a custom firmware, to be programmed.

The devices are placed into a Program/Verify mode by holding the RB1 and RB0 pins low while raising the $\overline{MCLR}$ (V_{PP}) pin from V_{IL} to V_{IHH}. RB1 becomes the programming clock and B0 becomes the programming data. Both RB1 and RB0 are Schmitt Trigger inputs in this mode.

After Reset, a 6-bit command is then supplied to the device. Depending on the command, 14 bits of program data are then supplied to or from the device, depending if the command was a Load or a Read.

A typical In-Circuit Serial Programming connection is shown in Figure 7-12.

FIGURE 7-12: TYPICAL IN-CIRCUIT SERIAL PROGRAMMING CONNECTION



8.0 ANALOG-TO-DIGITAL (A/D) CONVERTER

The A/D Converter allows conversion of an analog signal into an 8-bit digital signal.

8.1 Clock Divisors

The ADC has 4 clock source settings ADCS<1:0>. There are 3 divisor values 16, 8 and 4. The fourth setting is INTOSC with a divisor of 4. These settings will allow a proper conversion when using an external oscillator at speeds from 20 MHz to 350 kHz. Using an external oscillator at a frequency below 350 kHz requires the ADC oscillator setting to be INTOSC/4 (ADCS<1:0> = 11) for valid ADC results.

The ADC requires 13 TAD periods to complete a conversion. The divisor values do not affect the number of TAD periods required to perform a conversion. The divisor values determine the length of the TAD period.

When the ADCS<1:0> bits are changed while an ADC conversion is in process, the new ADC clock source will not be selected until the next conversion is started. This clock source selection will be lost when the device enters Sleep.

Note:	The ADC clock is derived from the instruction clock. The ADCS divisors are then applied to create the ADC clock

8.1.1 VOLTAGE REFERENCE

There is no external voltage reference for the ADC. The ADC reference voltage will always be VDD.

8.1.2 ANALOG MODE SELECTION

The ANS<1:0> bits are used to configure pins for analog input. Upon any Reset, ANS<1:0> defaults to 11. This configures pins AN0, AN1 and AN2 as analog inputs. The comparator output, C1OUT, will override AN2 as an input if the comparator output is enabled. Pins configured as analog inputs are not available for digital output. Users should not change the ANS bits while a conversion is in process. ANS bits are active regardless of the condition of ADON.

8.1.3 ADC CHANNEL SELECTION

The CHS bits are used to select the analog channel to be sampled by the ADC. The CHS<1:0> bits can be changed at any time without adversely effecting a conversion. To acquire an analog signal the CHS<1:0> selection must match one of the pin(s) selected by the ANS<1:0> bits. When the ADC is on (ADON = 1) and a channel is selected that is also being used by the comparator, then both the comparator and the ADC will see the analog voltage on the pin.

Note:	It is the users responsibility to ensure that use of the ADC and comparator simultaneously on the same pin, does not adversely affect the signal being monitored or adversely effect device operation.

When the CHS<1:0> bits are changed during an ADC conversion, the new channel will not be selected until the current conversion is completed. This allows the current conversion to complete with valid results. All channel selection information will be lost when the device enters Sleep.

TABLE 8-1: CHANNEL SELECT (ADCS) BITS AFTER AN EVENT

Event	ADCS<1:0>
$\overline{\text{MCLR}}$	11
Conversion completed	CS<1:0>
Conversion terminated	CS<1:0>
Power-on	11
Wake from Sleep	11

8.1.4 THE GO/$\overline{\text{DONE}}$ BIT

The GO/$\overline{\text{DONE}}$ bit is used to determine the status of a conversion, to start a conversion and to manually halt a conversion in process. Setting the GO/$\overline{\text{DONE}}$ bit starts a conversion. When the conversion is complete, the ADC module clears the GO/$\overline{\text{DONE}}$ bit. A conversion can be terminated by manually clearing the GO/$\overline{\text{DONE}}$ bit while a conversion is in process. Manual termination of a conversion may result in a partially converted result in ADRES.

The GO/$\overline{\text{DONE}}$ bit is cleared when the device enters Sleep, stopping the current conversion. The ADC does not have a dedicated oscillator, it runs off of the instruction clock. Therefore, no conversion can occur in sleep.

The GO/$\overline{\text{DONE}}$ bit cannot be set when ADON is clear.

8.1.5 SLEEP

This ADC does not have a dedicated ADC clock, and therefore, no conversion in Sleep is possible. If a conversion is underway and a Sleep command is executed, the GO/$\overline{\text{DONE}}$ and ADON bit will be cleared. This will stop any conversion in process and power-down the ADC module to conserve power. Due to the nature of the conversion process, the ADRES may contain a partial conversion. At least 1 bit must have been converted prior to Sleep to have partial conversion data in ADRES. The ADCS and CHS bits are reset to their default condition; ANS<1:0> = 11 and CHS<1:0> = 11.

- For accurate conversions, TAD must meet the following:
- 500 ns < TAD < 50 µs
- TAD = 1/(FOSC/divisor)

Shaded areas indicate TAD out of range for accurate conversions. If analog input is desired at these frequencies, use INTOSC/8 for the ADC clock source.

TABLE 8-2: TAD FOR ADCS SETTINGS WITH VARIOUS OSCILLATORS

Source	ADCS <1:0>	Divisor	20 MHz	16 MHz	8 MHz	4 MHz	1 MHz	500 kHz	350 kHz	200 kHz	100 kHz	32 kHz
INTOSC	11	4	—	—	.5 µs	1 µs	—	—	—	—	—	—
FOSC	10	4	.2 µs	.25 µs	.5 µs	1 µs	4 µs	8 µs	11 µs	20 µs	40 µs	125 µs
FOSC	01	8	.4 µs	.5 µs	1 µs	2 µs	8 µs	16 µs	23 µs	40 µs	80 µs	250 µs
FOSC	00	16	.8 µs	1 µs	2 µs	4 µs	16 µs	32 µs	46 µs	80 µs	160 µs	500 µs

TABLE 8-3: EFFECTS OF SLEEP ON ADCON0

	ANS1	ANS0	ADCS1	ADCS0	CHS1	CHS0	GO/$\overline{\text{DONE}}$	ADON
Entering Sleep	Unchanged	Unchanged	1	1	1	1	0	0
Wake or Reset	1	1	1	1	1	1	0	0

8.1.6 ANALOG CONVERSION RESULT REGISTER

The ADRES register contains the results of the last conversion. These results are present during the sampling period of the next analog conversion process. After the sampling period is over, ADRES is cleared (= 0). A 'leading one' is then right shifted into the ADRES to serve as an internal conversion complete bit. As each bit weight, starting with the MSB, is converted, the leading one is shifted right and the converted bit is stuffed into ADRES. After a total of 9 right shifts of the 'leading one' have taken place, the conversion is complete; the 'leading one' has been shifted out and the GO/$\overline{\text{DONE}}$ bit is cleared.

If the GO/$\overline{\text{DONE}}$ bit is cleared in software during a conversion, the conversion stops. The data in ADRES is the partial conversion result. This data is valid for the bit weights that have been converted. The position of the 'leading one' determines the number of bits that have been converted. The bits that were not converted before the GO/$\overline{\text{DONE}}$ was cleared are unrecoverable.

REGISTER 8-1: ADCON0: A/D CONTROL REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-0	R/W-0
ANS1	ANS0	ADCS1	ADCS0	CHS1	CHS0	GO/$\overline{\text{DONE}}$	ADON
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **ANS<1:0>:** ADC Analog Input Pin Select bits[1), (2), (5)]
- 00 = No pins configured for analog input
- 01 = AN2 configured as an analog input
- 10 = AN2 and AN0 configured as analog inputs
- 11 = AN2, AN1 and AN0 configured as analog inputs

bit 5-4 **ADCS<1:0>:** ADC Conversion Clock Select bits
- 00 = FOSC/16
- 01 = FOSC/8
- 10 = FOSC/4
- 11 = INTOSC/4

bit 3-2 **CHS<1:0>:** ADC Channel Select bits[3, 5]
- 00 = Channel AN0
- 01 = Channel AN1
- 10 = Channel AN2
- 11 = 0.6V absolute voltage reference

bit 1 **GO/$\overline{\text{DONE}}$:** ADC Conversion Status bit[4]
- 1 = ADC conversion in progress. Setting this bit starts an ADC conversion cycle. This bit is automatically cleared by hardware when the ADC is done converting.
- 0 = ADC conversion completed/not in progress. Manually clearing this bit while a conversion is in process terminates the current conversion.

bit 0 **ADON:** ADC Enable bit
- 1 = ADC module is operating
- 0 = ADC module is shut-off and consumes no power

Note 1: When the ANS bits are set, the channels selected will automatically be forced into Analog mode, regardless of the pin function previously defined. The only exception to this is the comparator, where the analog input to the comparator and the ADC will be active at the same time. It is the users responsibility to ensure that the ADC loading on the comparator input does not affect their application.

2: The ANS<1:0> bits are active regardless of the condition of ADON.

3: CHS<1:0> bits default to 11 after any Reset.

4: If the ADON bit is clear, the GO/$\overline{\text{DONE}}$ bit cannot be set.

5: C1OUT, when enabled, overrides AN2.

REGISTER 8-2: ADRES: ADDRESS REGISTER

R/W-X	R/W-X	R/W-X	R/W-X	R/W-X	R/W-X	R/W-X	R/W-X
ADRES7	ADRES6	ADRES5	ADRES4	ADRES3	ADRES2	ADRES1	ADRES0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

EXAMPLE 8-1: PERFORMING AN ANALOG-TO-DIGITAL CONVERSION

```
;Sample code operates out of BANK0
        MOVLW 0xF1     ;configure A/D
        MOVWF ADCON0
        BSF ADCON0, 1  ;start conversion
loop0   BTFSC ADCON0, 1;wait for 'DONE'
        GOTO loop0
        MOVF ADRES, W  ;read result
        MOVWF result0  ;save result

        BSF ADCON0, 2  ;setup for read of
                       ;channel 1
        BSF ADCON0, 1  ;start conversion
loop1   BTFSC ADCON0, 1;wait for 'DONE'
        GOTO loop1
        MOVF ADRES, W  ;read result
        MOVWF result1  ;save result

        BSF ADCON0, 3  ;setup for read of
        BCF ADCON0, 2  ;channel 2
        BSF ADCON0, 1  ;start conversion
loop2   BTFSC ADCON0, 1;wait for 'DONE'
        GOTO loop2
        MOVF ADRES, W  ;read result
        MOVWF result2  ;save result
```

EXAMPLE 8-2: CHANNEL SELECTION CHANGE DURING CONVERSION

```
        MOVLW 0xF1     ;configure A/D
        MOVWF ADCON0
        BSF ADCON0, 1  ;start conversion
        BSF ADCON0, 2  ;setup for read of
                       ;channel 1
loop0   BTFSC ADCON0, 1;wait for 'DONE'
        GOTO loop0
        MOVF ADRES, W  ;read result
        MOVWF result0  ;save result

        BSF ADCON0, 1  ;start conversion
        BSF ADCON0, 3  ;setup for read of
        BCF ADCON0, 2  ;channel 2
loop1   BTFSC ADCON0, 1;wait for 'DONE'
        GOTO loop1
        MOVF ADRES, W  ;read result
        MOVWF result1  ;save result

        BSF ADCON0, 1  ;start conversion
loop2   BTFSC ADCON0, 1;wait for 'DONE'
        GOTO loop2
        MOVF ADRES, W  ;read result
        MOVWF result2  ;save result
        CLRF ADCON0    ;optional: returns
        ;pins to Digital mode and turns off
        ;the ADC module
```


9.0 COMPARATOR(S)

This device contains two comparators and a comparator voltage reference.

REGISTER 9-1: CM1CON0: COMPARATOR C1 CONTROL REGISTER

R-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
C1OUT	$\overline{\text{C1OUTEN}}$	C1POL	$\overline{\text{C1T0CS}}$	C1ON	C1NREF	C1PREF	$\overline{\text{C1WU}}$
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **C1OUT:** Comparator Output bit
1 = $V_{IN+} > V_{IN-}$
0 = $V_{IN+} < V_{IN-}$

bit 6 **$\overline{\text{C1OUTEN}}$:** Comparator Output Enable bit(1), (2)
1 = Output of comparator is NOT placed on the C1OUT pin
0 = Output of comparator is placed in the C1OUT pin

bit 5 **C1POL:** Comparator Output Polarity bit(2)
1 = Output of comparator is not inverted
0 = Output of comparator is inverted

bit 4 **$\overline{\text{C1T0CS}}$:** Comparator TMR0 Clock Source bit(2)
1 = TMR0 clock source selected by T0CS control bit
0 = Comparator output used as TMR0 clock source

bit 3 **C1ON:** Comparator Enable bit
1 = Comparator is on
0 = Comparator is off

bit 2 **C1NREF:** Comparator Negative Reference Select bit(2)
1 = C1IN- pin
0 = 0.6V V_{REF}

bit 1 **C1PREF:** Comparator Positive Reference Select bit(2)
1 = C1IN+ pin
0 = C1IN- pin

bit 0 **$\overline{\text{C1WU}}$:** Comparator Wake-up On Change Enable bit(2)
1 = Wake-up On Comparator Change is disabled
0 = Wake-up On Comparator Change is enabled

Note 1: Overrides T0CS bit for TRIS control of RB2.

2: When comparator is turned on, these control bits assert themselves. Otherwise, the other registers have precedence.

REGISTER 9-2: CM2CON0: COMPARATOR C2 CONTROL REGISTER

R-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
C2OUT	$\overline{\text{C2OUTEN}}$	C2POL	C2PREF2	C2ON	C2NREF	C2PREF1	$\overline{\text{C2WU}}$
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **C2OUT:** Comparator Output bit
1 = $V_{IN+} > V_{IN-}$
0 = $V_{IN+} < V_{IN-}$

bit 6 **$\overline{\text{C2OUTEN}}$:** Comparator Output Enable bit(1), (2)
1 = Output of comparator is NOT placed on the C2OUT pin
0 = Output of comparator is placed in the C2OUT pin

bit 5 **C2POL:** Comparator Output Polarity bit(2)
1 = Output of comparator not inverted
0 = Output of comparator inverted

bit 4 **C2PREF2:** Comparator Positive Reference Select bit(2)
1 = C1IN+ pin
0 = C2IN- pin

bit 3 **C2ON:** Comparator Enable bit
1 = Comparator is on
0 = Comparator is off

bit 2 **C2NREF:** Comparator Negative Reference Select bit(2)
1 = C2IN- pin
0 = CV_{REF}

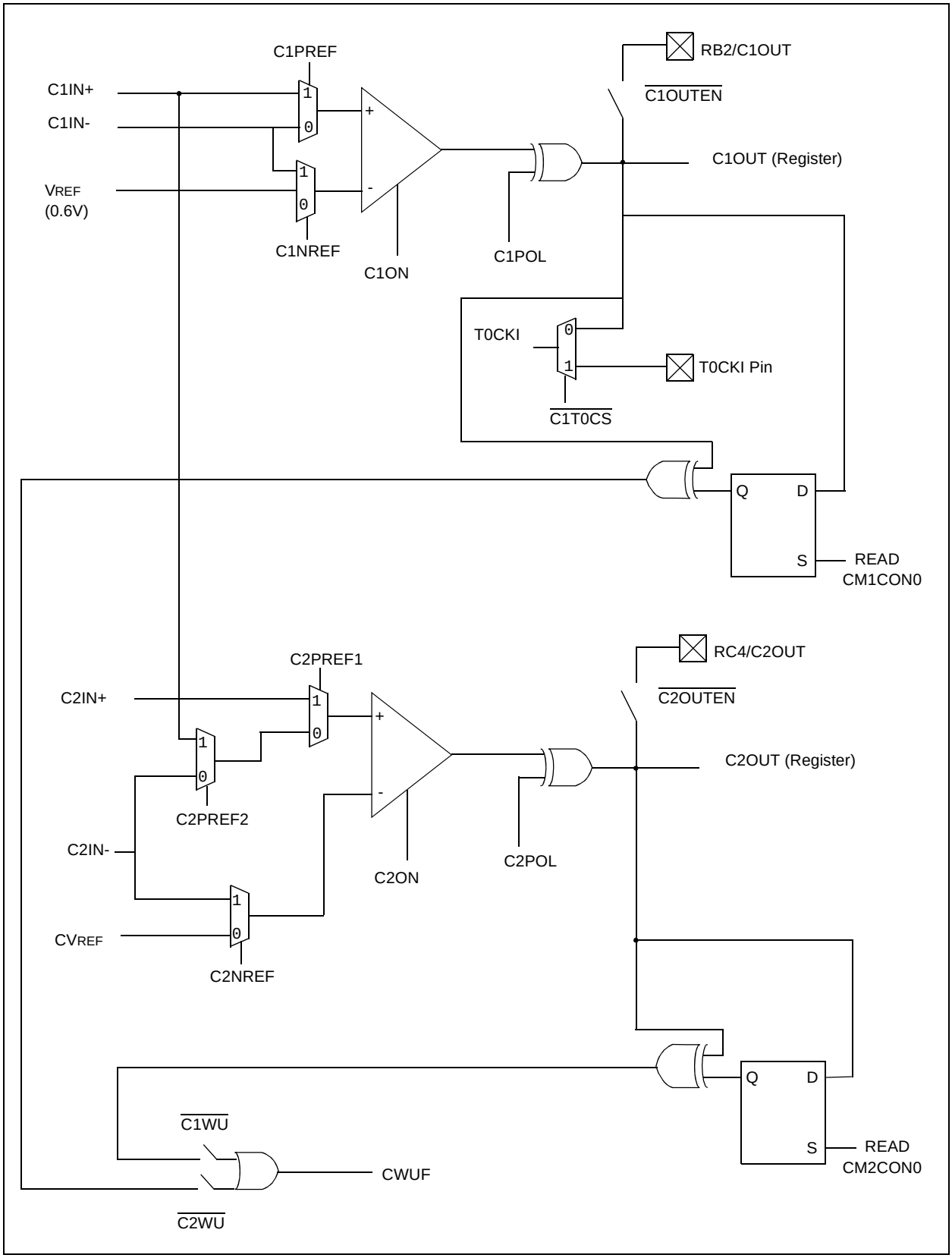
bit 1 **C2PREF1:** Comparator Positive Reference Select bit(2)
1 = C2IN+ pin
0 = C2PREF2 controls analog input selection

bit 0 **$\overline{\text{C2WU}}$:** Comparator Wake-up on Change Enable bit(2)
1 = Wake-up on Comparator change is disabled
0 = Wake-up on Comparator change is enabled.

Note 1: Overrides TOCS bit for TRIS control of RC4.

2: When comparator is turned on, these control bits assert themselves. Otherwise, the other registers have precedence.

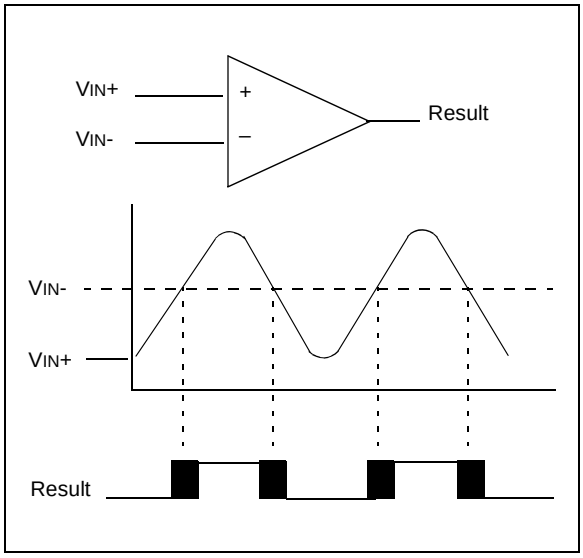
FIGURE 9-1: COMPARATORS BLOCK DIAGRAM



9.1 Comparator Operation

A single comparator is shown in Figure 9-2 along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. The shaded area of the output of the comparator in Figure 9-2 represent the uncertainty due to input offsets and response time. See Table 11-2 for Common Mode Voltage.

FIGURE 9-2: SINGLE COMPARATOR



9.2 Comparator Reference

An internal reference signal may be used depending on the comparator operating mode. The analog signal that is present at VIN- is compared to the signal at VIN+, and the digital output of the comparator is adjusted accordingly (Figure 9-2). Please see **Section 10.0 "Comparator Voltage Reference Module"** for internal reference specifications.

9.3 Comparator Response Time

Response time is the minimum time after selecting a new reference voltage or input source before the comparator output is to have a valid level. If the comparator inputs are changed, a delay must be used to allow the comparator to settle to its new state. Please see Table 11-3 for comparator response time specifications.

9.4 Comparator Output

The comparator output is read through the CM1CON0 or CM2CON0 register. This bit is read-only. The comparator output may also be used externally, see Figure 9-2.

Note:	Analog levels on any pin that is defined as a digital input may cause the input buffer to consume more current than is specified.

9.5 Comparator Wake-up Flag

The Comparator Wake-up Flag is set whenever all of the following conditions are met:

- $\overline{\text{C1WU}}$ = 0 (CM1CON0<0>) or $\overline{\text{C2WU}}$ = 0 (CM2CON0<0>)
- CM1CON0 or CM2CON0 has been read to latch the last known state of the C1OUT and C2OUT bit (`MOVF CM1CON0, W`)
- Device is in Sleep
- The output of a comparator has changed state

The wake-up flag may be cleared in software or by another device Reset.

9.6 Comparator Operation During Sleep

When the comparator is enabled it is active. To minimize power consumption while in Sleep mode, turn off the comparator before entering Sleep.

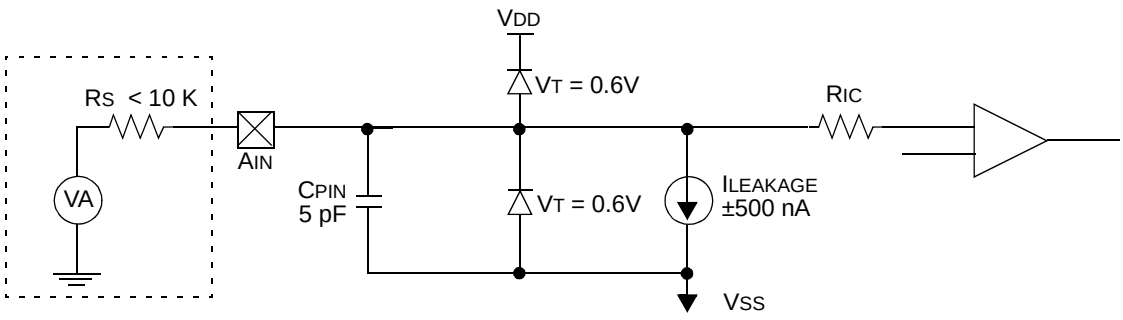
9.7 Effects of Reset

A Power-on Reset (POR) forces the CM2CON0 register to its Reset state. This forces the Comparator input pins to analog Reset mode. Device current is minimized when analog inputs are present at Reset time.

9.8 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 9-3. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and VSS. The analog input, therefore, must be between VSS and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur. A maximum source impedance of 10 kΩ is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.

FIGURE 9-3: ANALOG INPUT MODE



Legend:
CPIN = Input Capacitance
VT = Threshold Voltage
ILEAKAGE = Leakage Current at the Pin
RIC = Interconnect Resistance
RS = Source Impedance
VA = Analog Voltage

TABLE 9-1: REGISTERS ASSOCIATED WITH COMPARATOR MODULE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on All Other Resets
STATUS	RBWUF	CWUF	PA0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	qq0q quuu
CM1CON0	C1OUT	$\overline{C1OUTEN}$	C1POL	$\overline{C1T0CS}$	C1ON	C1NREF	C1PREF	$\overline{C1WU}$	1111 1111	uuuu uuuu
CM2CON0	C2OUT	$\overline{C2OUTEN}$	C2POL	C2PREF 2	C2ON	C2NREF	C2PREF 1	$\overline{C2WU}$	1111 1111	uuuu uuuu
TRIS	—	—	I/O Control Register (PORTB, PORTC)						--11 1111	--11 1111

Legend: x = Unknown, u = Unchanged, – = Unimplemented, read as '0', q = Depends on condition.

NOTES:

10.0 COMPARATOR VOLTAGE REFERENCE MODULE

The Comparator Voltage Reference module also allows the selection of an internally generated voltage reference for one of the C2 comparator inputs. The VRCON register (Register 10-1) controls the Voltage Reference module shown in Figure 10-1.

10.1 Configuring The Voltage Reference

The voltage reference can output 32 voltage levels; 16 in a high range and 16 in a low range.

Equation 10-1 determines the output voltages:

EQUATION 10-1:

VRR = 1 (low range): $CV_{REF} = (VR<3:0>/24) \times V_{DD}$

VRR = 0 (high range):

$$CV_{REF} = (V_{DD}/4) + (VR<3:0> \times V_{DD}/32)$$

10.2 Voltage Reference Accuracy/Error

The full range of V_{SS} to V_{DD} cannot be realized due to construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 10-1) keep CV_{REF} from approaching V_{SS} or V_{DD}. The exception is when the module is disabled by clearing the VREN bit (VRCON<7>). When disabled, the reference voltage is V_{SS} when VR<3:0> is '0000' and the VRR (VRCON<5>) bit is set. This allows the comparator to detect a zero-crossing and not consume the CV_{REF} module current.

The voltage reference is V_{DD} derived and, therefore, the CV_{REF} output changes with fluctuations in V_{DD}. The tested absolute accuracy of the comparator voltage reference can be found in **Section 11.2 "DC Characteristics: MCV14A"**.

REGISTER 10-1: VRCON: VOLTAGE REFERENCE CONTROL REGISTER

R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
VREN	VROE	VRR	—	VR3	VR2	VR1	VR0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **VREN:** CV_{REF} Enable bit
1 = CV_{REF} is powered on
0 = CV_{REF} is powered down, no current is drawn

bit 6 **VROE:** CV_{REF} Output Enable bit(1)
1 = CV_{REF} output is enabled
0 = CV_{REF} output is disabled

bit 5 **VRR:** CV_{REF} Range Selection bit
1 = Low range
0 = High range

bit 4 **Unimplemented:** Read as '0'

bit 3-0 **VR<3:0>** CV_{REF} Value Selection bit
When V_{RR} = 1: $CV_{REF}= (VR<3:0>/24)*V_{DD}$
When V_{RR} = 0: $CV_{REF}= V_{DD}/4+(VR<3:0>/32)*V_{DD}$

Note 1: When this bit is set, the TRIS for the CV_{REF} pin is overridden and the analog voltage is placed on the CV_{REF} pin.

2: CV_{REF} controls for ratio metric reference applies to Comparator 2.

FIGURE 10-1: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM

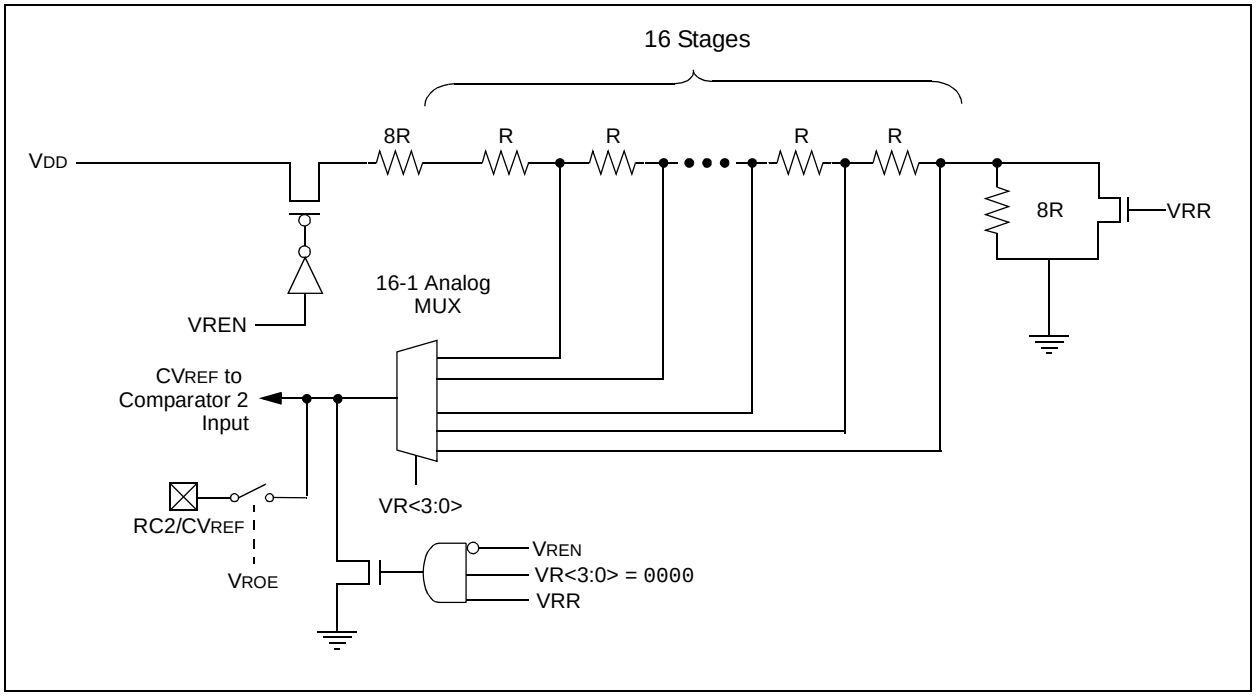


TABLE 10-1: REGISTERS ASSOCIATED WITH COMPARATOR VOLTAGE REFERENCE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on all other Resets
VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	000- 0000	000- 0000
CM1CON0	C1OUT	$\overline{\text{C1OUTEN}}$	C1POL	$\overline{\text{C1T0CS}}$	C1ON	C1NREF	C1PREF	$\overline{\text{C1WU}}$	1111 1111	uuuu uuuu
CM2CON0	C2OUT	$\overline{\text{C2OUTEN}}$	C2POL	C2PREF2	C2ON	C2NREF	C2PREF1	$\overline{\text{C2WU}}$	1111 1111	uuuu uuuu

Legend: x = unknown, u = unchanged, – = unimplemented, read as '0'.

11.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings(†)

Ambient temperature under bias ..... -40°C to +85°C
Storage temperature ..... -65°C to +150°C
Voltage on V_{DD} with respect to V_{SS} ..... 0 to +6.5V
Voltage on $\overline{MCLR}$ with respect to V_{SS} ..... 0 to +13.5V
Voltage on all other pins with respect to V_{SS} ..... -0.3V to (V_{DD} + 0.3V)
Total power dissipation(1) ..... 700 mW
Max. current out of V_{SS} pin ..... 200 mA
Max. current into V_{DD} pin ..... 150 mA
Input clamp current, I_{IK} (V_I < 0 or V_I > V_{DD}) ..... ±20 mA
Output clamp current, I_{OK} (V_O < 0 or V_O > V_{DD}) ..... ±20 mA
Max. output current sunk by any I/O pin ..... 25 mA
Max. output current sourced by any I/O pin ..... 25 mA
Max. output current sourced by I/O port ..... 75 mA
Max. output current sunk by I/O port ..... 75 mA

Note 1: Power dissipation is calculated as follows: $P_{DIS} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum (V_{OL} \times I_{OL})$

†NOTICE: Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

FIGURE 11-1: MCV14A VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA ≤ +85°C

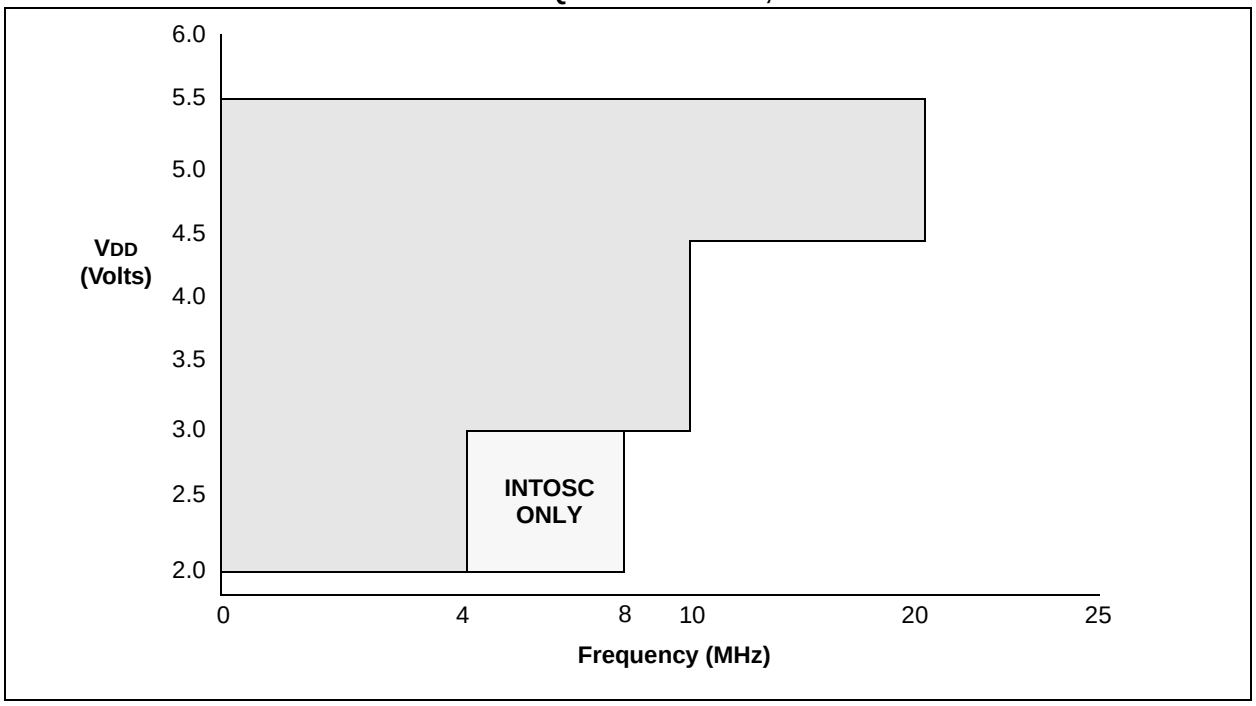
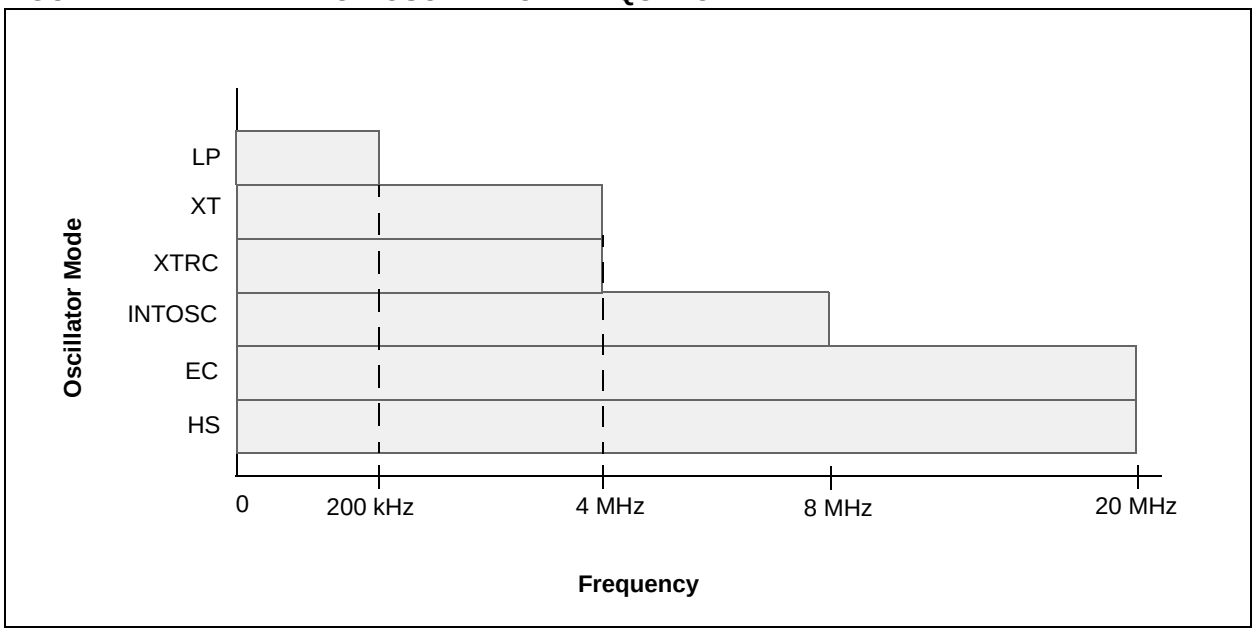


FIGURE 11-2: MAXIMUM OSCILLATOR FREQUENCY TABLE



11.1 DC Characteristics: MCV14A (Industrial)

DC Characteristics			Standard Operating Conditions (unless otherwise specified) Operating Temperature 40°C ≤ TA ≤ +85°C (industrial)				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
D001	VDD	**Supply Voltage**	2.0		5.5	V	See **Figure 11-1**
D002	VDR	**RAM Data Retention Voltage(2)**	—	1.5*	—	V	Device in Sleep mode
D003	VPOR	**VDD Start Voltage** to ensure Power-on Reset	—	Vss	—	V	See **Section 7.4 "Power-on Reset (POR)"** for details
D004	SVDD	**VDD Rise Rate** to ensure Power-on Reset	0.05*	—	—	V/ms	See **Section 7.4 "Power-on Reset (POR)"** for details
D010	IDD	**Supply Current(3,4)**	— —	175 400	250 700	μA mA	FOSC = 4 MHz, VDD = 2.0V FOSC = 4 MHz, VDD = 5.0V
			— —	250 0.75	450 1.2	μA mA	FOSC = 8 MHz, VDD = 2.0V FOSC = 8 MHz, VDD = 5.0V
			—	1.8	2.5	mA	FOSC = 20 MHz, VDD = 5.0V
			— —	11 38	22 55	μA μA	FOSC = 32 kHz, VDD = 2.0V FOSC = 32 kHz, VDD = 5.0V
D020	IPD	**Power-down Current(5)**	— —	0.1 0.35	1.2 2.2	μA μA	VDD = 2.0V VDD = 5.0V
D022	IWDT	**WDT Current(5)**	— —	1.0 7.0	3.0 16.0	μA μA	VDD = 2.0V VDD = 5.0V
D023	ICMP	**Comparator Current(5)**	— —	15 60	26 76	μA μA	VDD = 2.0V (per comparator) VDD = 5.0V (per comparator)
D022	IVREF	**VREF Current(5)**	— —	30 75	75 135	μA μA	VDD = 2.0V (high range) VDD = 5.0V (high range)
D023	IFVR	**Internal 0.6V Fixed Voltage Reference Current(5)**	— —	100 175	120 205	μA μA	VDD = 2.0V (reference and 1 comparator enabled) VDD = 5.0V (reference and 1 comparator enabled)
D024	ΔIAD	**A/D Conversion Current**	—	120	150	μA	2.0V
			—	200	250	μA	5.0V

* These parameters are characterized but not tested.

Note 1: Data in the Typical ("Typ") column is based on characterization results at 25°C. This data is for design guidance only and is not tested.

2: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.

3: The supply current is mainly a function of the operating voltage and frequency. Other factors such as bus loading, oscillator type, bus rate, internal code execution pattern and temperature also have an impact on the current consumption.

4: The test conditions for all IDD measurements in Active Operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VSS, T0CKI = VDD, $\overline{\text{MCLR}}$ = VDD; WDT enabled/disabled as specified.

5: For standby current measurements, the conditions are the same as IDD, except that the device is in Sleep mode. If a module current is listed, the current is for that specific module enabled and the device in Sleep.

6: Does not include current through REXT. The current through the resistor can be estimated by the formula: I = VDD/2REXT (mA) with REXT in kΩ.

11.2 DC Characteristics: MCV14A

TABLE 11-1: DC CHARACTERISTICS: MCV14A (Industrial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise specified) Operating temperature -40°C ≤ TA ≤ +85°C (industrial) Operating voltage VDD range as described in DC spec.				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
	VIL	**Input Low Voltage**					
		I/O ports					
D030		with TTL buffer	Vss	—	0.8V	V	For all 4.5 ≤ VDD ≤ 5.5V
D030A			Vss	—	0.15 VDD	V	Otherwise
D031		with Schmitt Trigger buffer	Vss	—	0.15 VDD	V	
D032		$\overline{MCLR}$, T0CKI	Vss	—	0.15 VDD	V	
D033		OSC1 (EXTRC mode), EC(1)	Vss	—	0.15 VDD	V	
D033		OSC1 (HS mode)	Vss	—	0.3 VDD	V	
D033		OSC1 (XT and LP modes)	Vss	—	0.3	V	
	VIH	**Input High Voltage**					
		I/O ports		—			
D040		with TTL buffer	2.0	—	VDD	V	4.5 ≤ VDD ≤ 5.5V
D040A			0.25VDD + 0.8V	—	VDD	V	Otherwise
D041		with Schmitt Trigger buffer	0.85VDD	—	VDD	V	For entire VDD range
D042		MCLR, T0CKI	0.85VDD	—	VDD	V	
D042A		OSC1 (EXTRC mode), EC(1)	0.85VDD	—	VDD	V	
D042A		OSC1 (HS mode)	0.7VDD	—	VDD	V	
D043		OSC1 (XT and LP modes)	1.6	—	VDD	V	
D070	IPUR	**PORTB weak pull-up current(4)**	50	250	400	µA	VDD = 5V, VPIN = VSS
	IIL	**Input Leakage Current(2)**					
D060		I/O ports	—	—	±1	µA	Vss ≤ VPIN ≤ VDD, Pin at high-impedance
D061		RB3/$\overline{MCLR}$(3)	—	±0.7	±5	µA	Vss ≤ VPIN ≤ VDD
D063		OSC1	—	—	±5	µA	Vss ≤ VPIN ≤ VDD, XT, HS and LP osc configuration
		Output Low Voltage					
D080		I/O ports	—	—	0.6	V	IOL = 8.5 mA, VDD = 4.5V, –40°C to +85°C
D083		CLKOUT	—	—	0.6	V	IOL = 1.6 mA, VDD = 4.5V, –40°C to +85°C
		Output High Voltage					
D090		I/O ports(2)	VDD – 0.7	—	—	V	IOH = -3.0 mA, VDD = 4.5V, –40°C to +85°C
D092		CLKOUT	VDD – 0.7	—	—	V	IOH = -1.3 mA, VDD = 4.5V, –40°C to +85°C
		Capacitive Loading Specs on Output Pins					
D100		OSC2 pin	—	—	15	pF	In XT, HS and LP modes when external clock is used to drive OSC1.
D101		All I/O pins and OSC2	—	—	50	pF	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In EXTRC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended that the MCV14A be driven with external clock in RC mode.

2: Negative current is defined as coming out of the pin.

3: This spec. applies to RB3/$\overline{MCLR}$ configured as RB3 with internal pull-up disabled.

4: This spec applies to all weak pull-up devices, including the weak pull-up found on RB3/$\overline{MCLR}$. The current value listed will be the same whether or not the pin is configured as RB3 with pull-up enabled or as $\overline{MCLR}$.

TABLE 11-2: COMPARATOR SPECIFICATIONS.

Comparator Specifications	Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C to 85°C					
Characteristics	**Sym**	**Min**	**Typ**	**Max**	**Units**	**Comments**
Internal Voltage Reference	VIVRF	0.50	0.60	0.70	V	
Input offset voltage	VOS	—	± 5.0	± 10	mV	
Input common mode voltage*	VCM	0	—	VDD – 1.5	V	
CMRR*	CMRR	55	—	—	db	
Response Time(1)*	TRT	—	150	400	ns	
Comparator Mode Change to Output Valid*	TMC2COV	—	—	10	μs	

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at (VDD – 1.5)/2 while the other input transitions from VSS to VDD – 1.5V.

TABLE 11-3: COMPARATOR VOLTAGE REFERENCE (VREF) SPECIFICATIONS

Sym	Characteristics	Min	Typ	Max	Units	Comments
CVRES	Resolution	— —	VDD/24* VDD/32	— —	LSb LSb	Low Range (VRR = 1) High Range (VRR = 0)
	Absolute Accuracy(2)	— —	— —	±1/2* ±1/2*	LSb LSb	Low Range (VRR = 1) High Range (VRR = 0)
	Unit Resistor Value (R)	— —	2K*	—	Ω	
	Settling Time(1)	—	—	10*	μs	

* These parameters are characterized but not tested.

Note 1: Settling time measured while VRR = 1 and VR<3:0> transitions from 0000 to 1111.

2: Do not use reference externally when VDD < 2.7V. Under this condition, reference should only be used with comparator Voltage Common mode observed.

TABLE 11-4: A/D CONVERTER CHARACTERISTICS:

A/D Converter Specifications			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
A01	NR	Resolution	—	—	8	bit	
A03	EINL	Integral Error	—	—	±1.5	LSb	VDD = 5.0V
A04	EDNL	Differential Error	—	—	-1≤ EDNL ≤1.7	LSb	VDD = 5.0V
A06	EOFF	Offset Error	—	—	±1.5	LSb	VDD = 5.0V
A07	EGN	Gain Error	-0.7	—	+2.2	LSb	VDD = 5.0V
A10	—	Monotonicity	—	guaranteed(1)	—	—	VSS ≤ VAIN ≤ VDD
A25	VAIN	Analog Input Voltage	VSS	—	VDD	V	
A30	ZAIN	Recommended Impedance of Analog Voltage Source	—	—	10	KΩ	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The A/D conversion result never decreases with an increase in the input voltage.

TABLE 11-5: PULL-UP RESISTOR RANGES

VDD (Volts)	Temperature (°C)	Min	Typ	Max	Units
RB0/RB1					
2.0	-40	73K	105K	186K	Ω
	25	73K	113K	187K	Ω
	85	82K	123K	190K	Ω
	125	86K	132k	190K	Ω
5.5	-40	15K	21K	33K	Ω
	25	15K	22K	34K	Ω
	85	19K	26k	35K	Ω
	125	23K	29K	35K	Ω
RB3					
2.0	-40	63K	81K	96K	Ω
	25	77K	93K	116K	Ω
	85	82K	96k	116K	Ω
	125	86K	100K	119K	Ω
5.5	-40	16K	20k	22K	Ω
	25	16K	21K	23K	Ω
	85	24K	25k	28K	Ω
	125	26K	27K	29K	Ω

11.3 Timing Parameter Symbology and Load Conditions

The timing parameter symbols have been created following one of the following formats:

1. TppS2ppS
2. TppS

T			
F	Frequency	T	Time

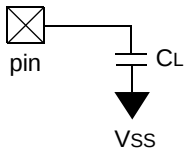
Lowercase subscripts (pp) and their meanings:

pp			
2	to	mc	$\overline{\text{MCLR}}$
ck	CLKOUT	osc	Oscillator
cy	Cycle time	os	OSC1
drt	Device Reset Timer	t0	T0CKI
io	I/O port	wdt	Watchdog Timer

Uppercase letters and their meanings:

S			
F	Fall	P	Period
H	High	R	Rise
I	Invalid (high-impedance)	V	Valid
L	Low	Z	High-impedance

FIGURE 11-3: LOAD CONDITIONS



Legend:

C_L = 50 pF for all pins except OSC2

15 pF for OSC2 in XT, HS or LP modes when external clock is used to drive OSC1

FIGURE 11-4: EXTERNAL CLOCK TIMING

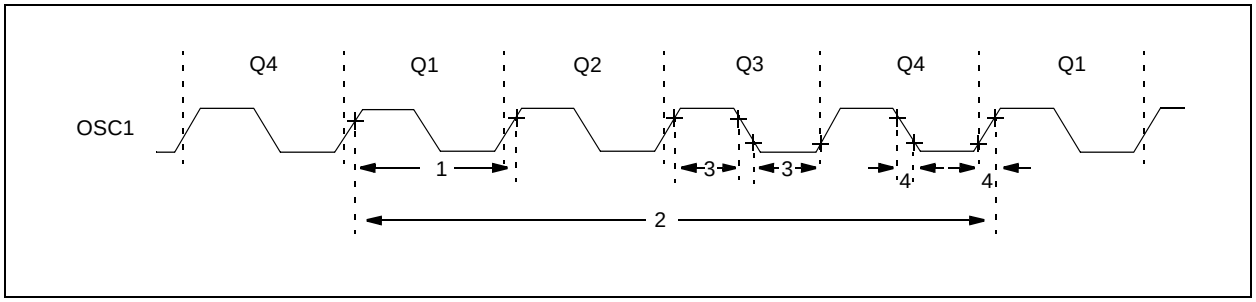


TABLE 11-6: EXTERNAL CLOCK TIMING REQUIREMENTS

AC CHARACTERISTICS			**Standard Operating Conditions (unless otherwise specified)** Operating Temperature -40°C ≤ TA ≤ +85°C (industrial), Operating Voltage VDD range is described in **Section 11.1 "DC Characteristics: MCV14A (Industrial)"**				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
1A	FOSC	External CLKIN Frequency(2)	DC	—	4	MHz	XT Oscillator mode
			DC	—	20	MHz	HS Oscillator mode
			DC	—	200	kHz	LP Oscillator mode
		Oscillator Frequency(2)	—	—	4	MHz	EXTRC Oscillator mode
			0.1	—	4	MHz	XT Oscillator mode
			4	—	20	MHz	HS Oscillator mode
			—	—	200	kHz	LP Oscillator mode
1	TOSC	External CLKIN Period(2)	250	—	—	ns	XT Oscillator mode
			50	—	—	ns	HS Oscillator mode
			5	—	—	µs	LP Oscillator mode
		Oscillator Period(2)	250	—	—	ns	EXTRC Oscillator mode
			250	—	10,000	ns	XT Oscillator mode
			50	—	250	ns	HS Oscillator mode
			5	—	—	µs	LP Oscillator mode
2	TCY	Instruction Cycle Time	200	4/FOSC	—	ns	
3	TosL, TosH	Clock in (OSC1) Low or High Time	50*	—	—	ns	XT Oscillator
			2*	—	—	µs	LP Oscillator
			10*	—	—	ns	HS Oscillator
4	TosR, TosF	Clock in (OSC1) Rise or Fall Time	—	—	25*	ns	XT Oscillator
			—	—	50*	ns	LP Oscillator
			—	—	15*	ns	HS Oscillator

* These parameters are characterized but not tested.

Note 1: Data in the Typical ("Typ") column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

2: All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

TABLE 11-7: CALIBRATED INTERNAL RC FREQUENCIES

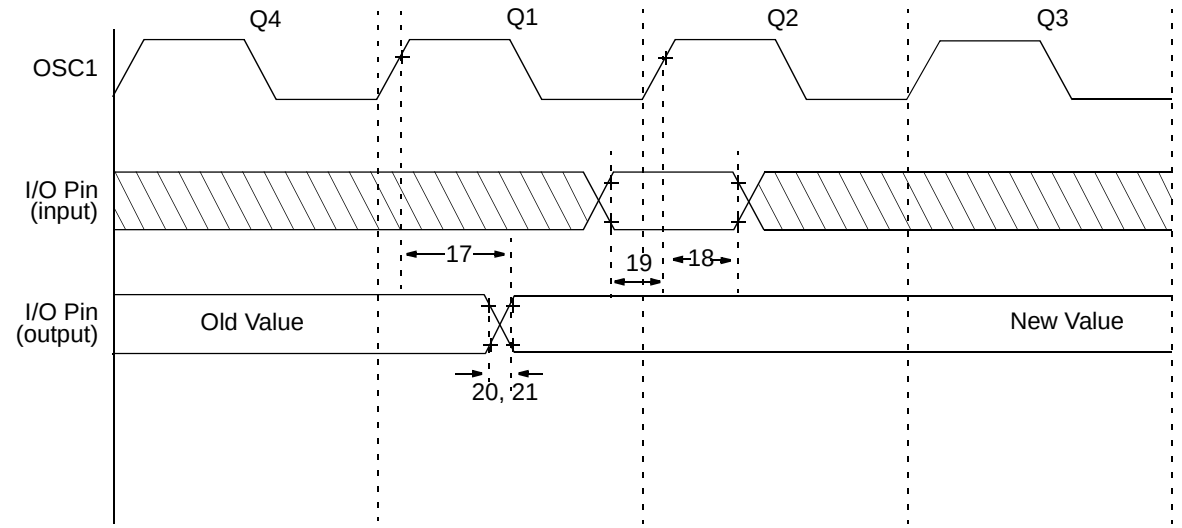
AC CHARACTERISTICS			Standard Operating Conditions (unless otherwise specified) Operating Temperature -40°C ≤ TA ≤ +85°C (industrial), Operating Voltage VDD range is described in Section 11.1 "DC Characteristics: MCV14A (Industrial)"					
Param No.	**Sym**	**Characteristic**	**Freq Tolerance**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
F10	FOSC	Internal Calibrated INTOSC Frequency(1)	± 1%	7.92	8.00	8.08	MHz	3.5V, +25°C
			± 5%	7.60	8.00	8.40	MHz	2.0V ≤ VDD ≤ 5.5V -40°C ≤ TA ≤ +85°C (Ind.)

* These parameters are characterized but not tested.

† Data in the Typical ("Typ") column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 uF and 0.01 uF values in parallel are recommended.

FIGURE 11-5: I/O TIMING



Note: All tests must be done with specified capacitive loads (see data sheet) 50 pF on I/O pins and CLKOUT.

TABLE 11-8: TIMING REQUIREMENTS

AC CHARACTERISTICS		**Standard Operating Conditions (unless otherwise specified)** Operating Temperature -40°C ≤ TA ≤ +85°C (industrial) Operating Voltage VDD range is described in **Section 11.1 "DC Characteristics: MCV14A (Industrial)"**				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**
17	TOSH2IOV	OSC1↑ (Q1 cycle) to Port Out Valid(2), (3)	—	—	100*	ns
18	TOSH2IOI	OSC1↑ (Q2 cycle) to Port Input Invalid (I/O in hold time)(2)	—	—	—	ns
19	TIOV2OSH	Port Input Valid to OSC1↑ (I/O in setup time)	—	—	—	ns
20	TIOR	Port Output Rise Time(3)	—	10	50**	ns
21	TIOF	Port Output Fall Time(3)	—	10	58**	ns

Legend: TBD = To Be Determined.

* These parameters are characterized but not tested.

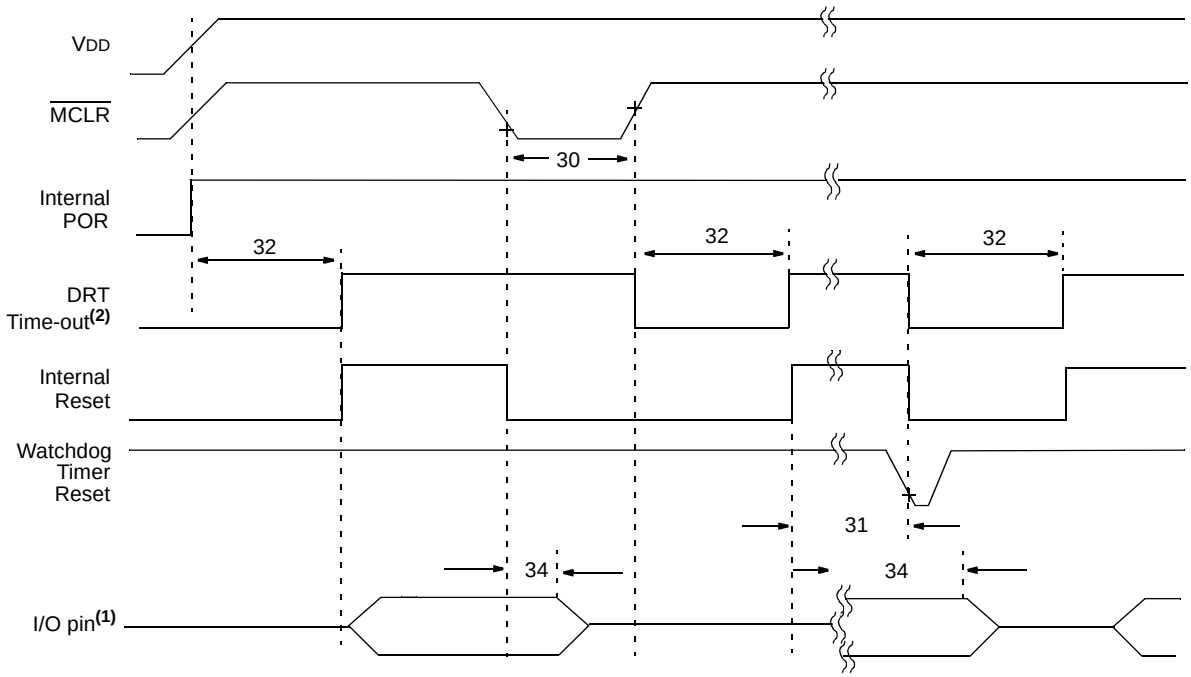
** These parameters are design targets and are not tested.

Note 1: Data in the Typical ("Typ") column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

2: Measurements are taken in EXTRC mode.

3: See Figure 11-3 for loading conditions.

FIGURE 11-6: RESET, WATCHDOG TIMER AND DEVICE RESET TIMER TIMING



Note 1: I/O pins must be taken out of High-Impedance mode by enabling the output drivers in software.
2: Runs in MCLR or WDT Reset only in XT, LP and HS modes.

TABLE 11-9: RESET, WATCHDOG TIMER AND DEVICE RESET TIMER

AC CHARACTERISTICS			**Standard Operating Conditions (unless otherwise specified)** Operating Temperature -40°C ≤ TA ≤ +85°C (industrial) Operating Voltage VDD range is described in **Section 11.1 "DC Characteristics: MCV14A (Industrial)"**				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
30	TMCL	MCLR Pulse Width (low)	2000*	—	—	ns	VDD = 5.0V
31	TWDT	Watchdog Timer Time-out Period (no prescaler)	9*	18*	30*	ms	VDD = 5.0V (Industrial)
32	TDRT	Device Reset Timer Period	9*	18*	30*	ms	VDD = 5.0V (Industrial)
34	TIOZ	I/O High-impedance from MCLR low	—	—	2000*	ns	

* These parameters are characterized but not tested.

Note 1: Data in the Typical ("Typ") column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 11-7: TIMER0 CLOCK TIMINGS

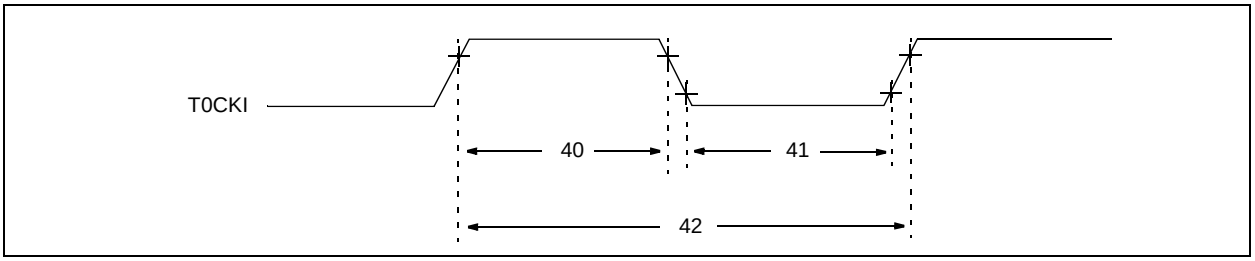


TABLE 11-10: TIMER0 CLOCK REQUIREMENT

AC CHARACTERISTICS				**Standard Operating Conditions (unless otherwise specified)** Operating Temperature -40°C ≤ TA ≤ +85°C (industrial) Operating Voltage VDD range is described in **Section 11.1 "DC Characteristics: MCV14A (Industrial)"**				
Param No.	**Sym**	**Characteristic**		**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
40	Tt0H	T0CKI High Pulse Width	No Prescaler	0.5 TCY + 20*	—	—	ns	
			With Prescaler	10*	—	—	ns	
41	Tt0L	T0CKI Low Pulse Width	No Prescaler	0.5 TCY + 20*	—	—	ns	
			With Prescaler	10*	—	—	ns	
42	Tt0P	T0CKI Period		20 or TCY + 40* N	—	—	ns	Whichever is greater. N = Prescale Value (1, 2, 4,..., 256)

* These parameters are characterized but not tested.

Note 1: Data in the Typical ("Typ") column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 11-11: FLASH DATA MEMORY WRITE/ERASE TIME

AC CHARACTERISTICS			**Standard Operating Conditions (unless otherwise specified)** Operating Temperature -40°C ≤ TA ≤ +85°C (industrial) Operating Voltage VDD range is described in **Section 11.1 "DC Characteristics: MCV14A (Industrial)"**				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ(1)**	**Max**	**Units**	**Conditions**
43	TDW	Flash Data Memory Write Cycle Time	2	3.5	5	ms	
44	TDE	Flash Data Memory Erase Cycle Time	2	3.5	5	ms	

* These parameters are characterized but not tested.

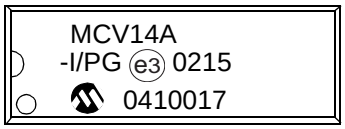
Note 1: Data in the Typical ("Typ") column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

12.0 PACKAGING INFORMATION

12.1 Package Marking Information

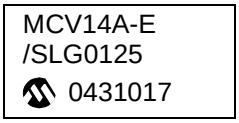
14-Lead PDIP (300 mil)

Example



14-Lead SOIC (3.90 mm)

Example



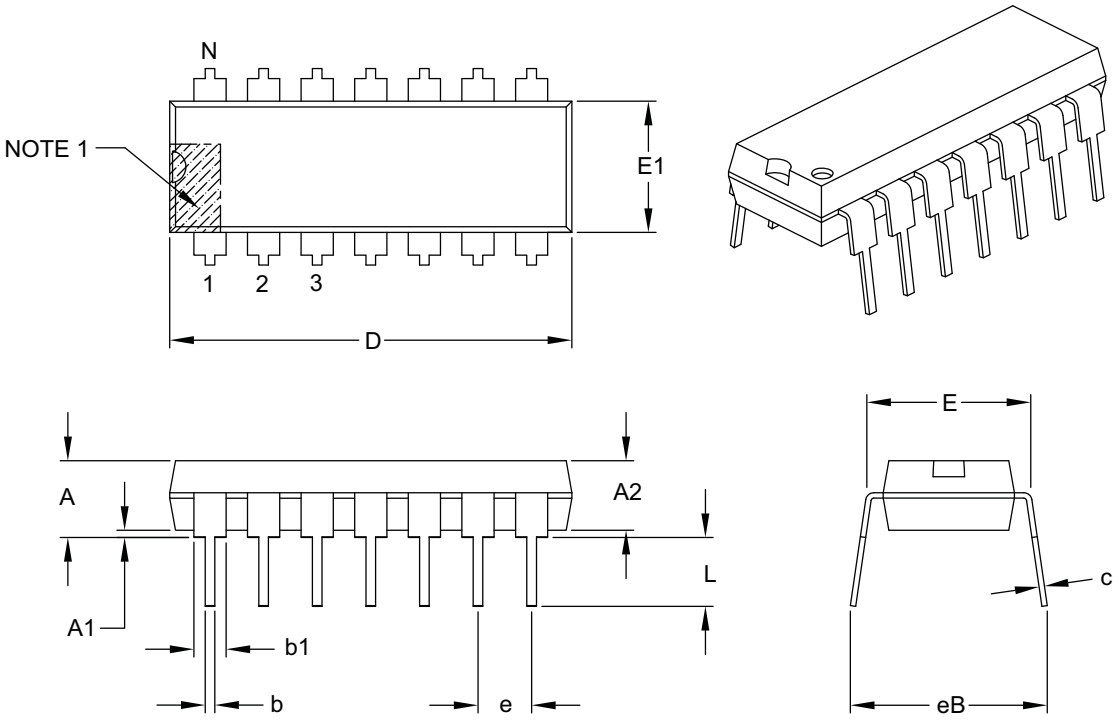
Legend:	XX...X	Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week ‘01’)
	NNN	Alphanumeric traceability code
	(e3)	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator ((e3)) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line thus limiting the number of available characters for customer specific information.

***** Standard MCV device marking consists of Microchip part number, year code, week code, and traceability code. For MCV device marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.

14-Lead Plastic Dual In-Line (P) – 300 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		14	
Pitch	e		.100 BSC	
Top to Seating Plane	A	–	–	.210
Molded Package Thickness	A2	.115	.130	.195
Base to Seating Plane	A1	.015	–	–
Shoulder to Shoulder Width	E	.290	.310	.325
Molded Package Width	E1	.240	.250	.280
Overall Length	D	.735	.750	.775
Tip to Seating Plane	L	.115	.130	.150
Lead Thickness	c	.008	.010	.015
Upper Lead Width	b1	.045	.060	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	–	–	.430

Notes:

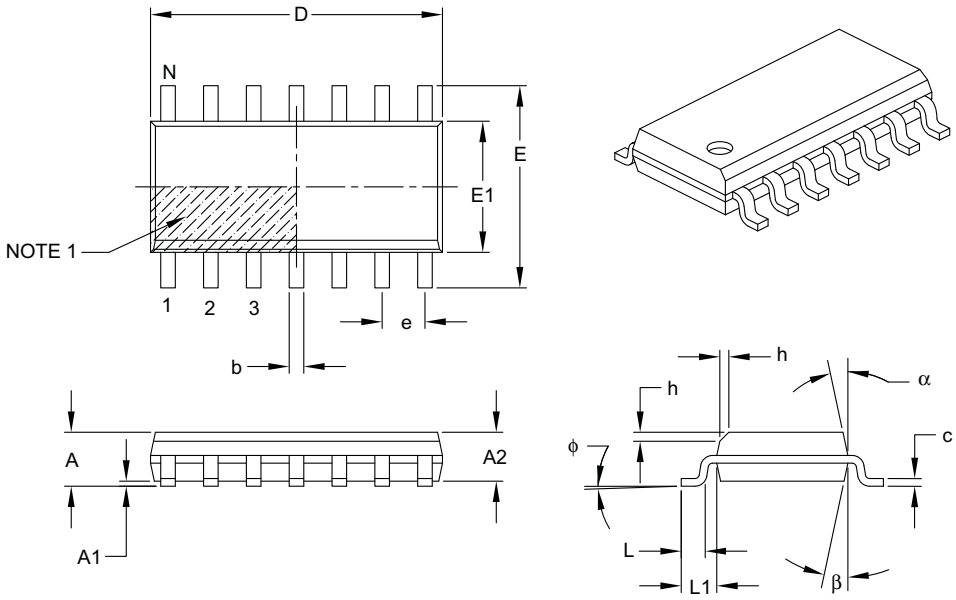
1. Pin 1 visual index feature may vary, but must be located with the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
4. Dimensioning and tolerancing per ASME Y14.5M.

   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-005B

14-Lead Plastic Small Outline (SL) – Narrow, 3.90 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		14	
Pitch	e		1.27 BSC	
Overall Height	A	–	–	1.75
Molded Package Thickness	A2	1.25	–	–
Standoff §	A1	0.10	–	0.25
Overall Width	E		6.00 BSC	
Molded Package Width	E1		3.90 BSC	
Overall Length	D		8.65 BSC	
Chamfer (optional)	h	0.25	–	0.50
Foot Length	L	0.40	–	1.27
Footprint	L1		1.04 REF	
Foot Angle	φ	0°	–	8°
Lead Thickness	c	0.17	–	0.25
Lead Width	b	0.31	–	0.51
Mold Draft Angle Top	α	5°	–	15°
Mold Draft Angle Bottom	β	5°	–	15°

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15 mm per side.
4. Dimensioning and tolerancing per ASME Y14.5M.

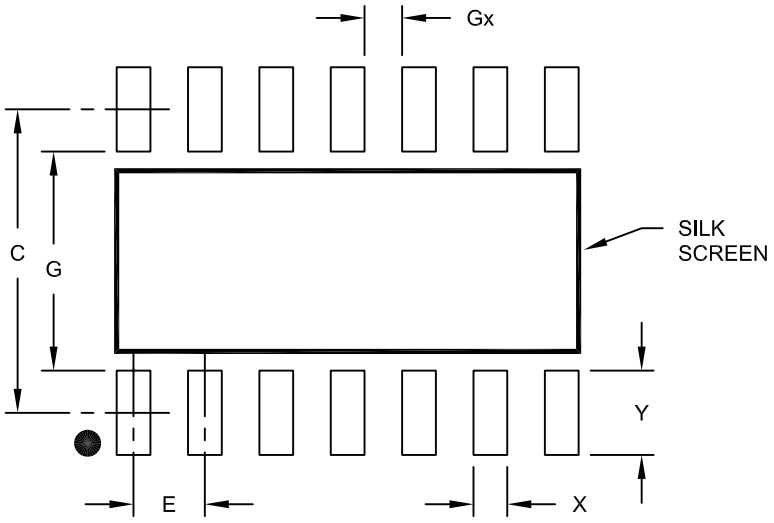
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

   REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-065B

14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		1.27 BSC	
Contact Pad Spacing	C		5.40	
Contact Pad Width	X			0.60
Contact Pad Length	Y			1.50
Distance Between Pads	Gx	0.67		
Distance Between Pads	G	3.90		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2065A

APPENDIX A: REVISION HISTORY

Revision A (November 2007)

Original release of this document.

Revision B (June 2009)

Revised Table 7-3: Reset Conditions for Registers; 11.1 DC Characteristics; Table 11-2: Comparator Specifications; Table 11-4: A/D Converter Characteristics; Table 11-11: Flash Data Memory Write/Erase Time.

Revision C (September 2009)

Revised Table 11-4: Deleted the “No missing codes to 8 bits” condition for Param. No. A04; Modified Note 1; Added SOIC (SL) Land Pattern Package.

NOTES:

INDEX

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	X	/XX	XXX
Device	Temperature Range	Package	Pattern

Device:	MCV14A MCV14AT(1)
Temperature Range:	I = -40°C to +85°C (Industrial)
Package:	P = Plastic (PDIP) SL = 14L Small Outline, 3.90 mm (SOIC)
Pattern:	Special Requirements

Examples:

a) MCV14A-I/SN = Industrial Temp., SOIC package

b) MCV14A-I/P = Industrial Temp., PDIP package

Note 1: T = in tape and reel SOIC package only

NOTES:

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